

NOTES FOR LECTURES 6B & 7A: CONTROLLER CANONICAL FORM & OBSERVABILITYJAIJEET RORCHANDHURUCONTROLLER CANONICAL FORM (CCF)

→ Recall a previous example: $A = \begin{bmatrix} 0 & 1 \\ a_1 & a_2 \end{bmatrix}$

→ its characteristic polynomial $c(\lambda) \triangleq \det(A - \lambda I)$ had a nice form:

$$\rightarrow -\lambda(a_2 - \lambda) - a_1 \equiv \boxed{\lambda^2 - a_2\lambda - a_1}$$

→ A rather elegant observation someone made long ago is this:

→ suppose you have a size n matrix in the same form:

$$A = \begin{bmatrix} 0 & 1 & & & & \\ 0 & 0 & 1 & & & \\ & 0 & 0 & 1 & & \\ & & \ddots & \ddots & \ddots & \\ & & & 1 & & \\ a_1 & a_2 & a_3 & \dots & \dots & a_n \end{bmatrix}; \text{ and } \vec{b} = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{bmatrix}$$

TOGETHER CALLED CONTROLLER CANONICAL FORM (CCF)

$$\rightarrow \text{then, } c(\lambda) \triangleq \det(A - \lambda I) = \lambda^n - a_n\lambda^{n-1} - a_{n-1}\lambda^{n-2} - \dots - a_3\lambda^2 - a_2\lambda - a_1$$

→ proof: easily (though laboriously) obtained by applying the formula for determinants involving expansion using minors to $(A - \lambda I)$. Try it yourself if you have some time on your hands and you like getting to the bottom of things.

→ Now, suppose you have a scalar input to your system ($u(t)$) and the system is: $\frac{d\vec{x}}{dt} = A\vec{x} + \vec{b}u(t)$

this (the original system without feedback) is called the OPEN LOOP system

→ Now, if you apply feedback $u(t) \mapsto u(t) - \vec{b}^T \vec{x}(t)$, then the "closed loop system" becomes:

$$\dot{\vec{x}} = (A - \vec{b} \vec{k}^T) \vec{x} + \vec{b} u(t),$$

$$\text{with } (A - \vec{b} \vec{k}^T) = \begin{bmatrix} 0 & 1 & & & \\ 0 & 0 & 1 & & \\ & & 0 & 1 & \\ & & & \ddots & \ddots \\ a_{1-k_1} & a_{2-k_2} & a_{3-k_3} & \cdots & a_{n-k_n} \end{bmatrix}$$

$$\rightarrow \text{with characteristic polynomial: } C_f(\lambda) \stackrel{\Delta}{=} \det(A - \vec{b} \vec{k}^T - \lambda I) \\ = \lambda^n - (a_{n-k_n}) \lambda^{n-1} - (a_{n-1-k_{n-1}}) \lambda^{n-2} - \cdots - (a_{3-k_3}) \lambda^2 - (a_{2-k_2}) \lambda - (a_{1-k_1})$$

→ Now, our goal is to choose k_1, \dots, k_n (the feedback) to place the eigenvalues of $(A - \vec{b} \vec{k}^T)$, i.e., the roots of $C_f(\lambda)$, wherever we want.

→ Suppose we want the roots to be $\underbrace{\lambda_1, \lambda_2, \dots, \lambda_n}$

↳ remember that if complex, the conjugate must also be present.

→ i.e., the desired char. poly. is:

$$(\lambda - \lambda_1)(\lambda - \lambda_2)(\lambda - \lambda_3) \cdots (\lambda - \lambda_n)$$

$$= \lambda^n - (\underbrace{\lambda_1 + \lambda_2 + \cdots + \lambda_n}_{b_n}) \lambda^{n-1} + \left(\lambda_1 \lambda_2 + \lambda_1 \lambda_3 + \cdots + \lambda_1 \lambda_n \right) \lambda^{n-2} + \cdots + (-1)^n \lambda_1 \lambda_2 \lambda_3 \cdots \lambda_n$$

$$+ \underbrace{\lambda_2 \lambda_3 + \lambda_2 \lambda_4 + \cdots + \lambda_2 \lambda_n}_{b_2}$$

$$+ \underbrace{\lambda_3 \lambda_4 + \cdots + \lambda_3 \lambda_n}_{b_3}$$

$$+ \underbrace{\lambda_n \lambda_1 \lambda_2 \cdots \lambda_{n-1}}_{b_{n-1}}$$

→ i.e., b_1, \dots, b_n can be calculated from $\lambda_1, \dots, \lambda_n$, though the process may be extremely tedious.

→ now all you have to do to devise the feedback is equate the coefficients of the 2 char. poly. expressions:

$$a_n - k_n = -b_n$$

$$a_{n-1} - k_{n-1} = -b_{n-1}$$

:

$$a_1 - k_1 = -b_1$$

→ I.e., if (A, \vec{b}) is in controllability canonical form, then it is always possible to devise feedback to place the eigenvalues wherever you like.

→ Size 3 example:

$$\rightarrow A = \begin{bmatrix} 0 & 0 \\ 0 & 0 & 1 \\ 1 & 2 & 3 \end{bmatrix}; \quad \vec{b} = \begin{bmatrix} 0 \\ 0 \\ 1 \end{bmatrix}$$

↓

$$\rightarrow A - \vec{b} \vec{k}^T = \begin{bmatrix} 0 & 0 \\ 0 & 0 & 1 \\ 1-k_1 & 2-k_2 & 3-k_3 \end{bmatrix}$$

$$\rightarrow \text{char. poly.: } \lambda^3 - (3-k_3)\lambda^2 - (2-k_2)\lambda - (1-k_1)$$

→ Suppose we want the 3 new eigenvalues to be:

$$\Rightarrow \lambda_1 = 0, \lambda_2 = 0, \lambda_3 = 0$$

$$\Rightarrow (\lambda - \lambda_1)(\lambda - \lambda_2)(\lambda - \lambda_3) = \lambda^3$$

→ Therefore we should set: $k_3 = 3, k_2 = 2, k_1 = 1$

→ Suppose we want $\lambda_1 = -1$, $\lambda_2 = -2$, $\lambda_3 = -3$

$$\Rightarrow (\lambda - \lambda_1)(\lambda - \lambda_2)(\lambda - \lambda_3) = \lambda^3 - \lambda(\lambda_1 + \lambda_2 + \lambda_3) + \lambda(\lambda_1\lambda_2 + \lambda_2\lambda_3 + \lambda_1\lambda_3)$$

$$= \lambda^3 + 6\lambda^2 + \lambda(\underbrace{2+6+3}_{11}) + 6$$

$$\Rightarrow -(3 - k_3) = -6 \Rightarrow k_3 = -3$$

$$\Rightarrow -(2 - k_2) = -11 \Rightarrow k_2 = -9$$

$$\Rightarrow -(1 - k_1) = -6 \Rightarrow k_1 = -5$$

→ But can any system be put in controller canonical form?

→ A: yes, if it is controllable!

→ Here is how you put any controllable system in c.c.f. :

1. Form the controllability matrix:

$$- R_n \stackrel{\triangle}{=} \left[\vec{b}, \vec{Ab}, \vec{A}^2\vec{b}, \dots, \vec{A}^{n-1}\vec{b} \right] \rightarrow \text{a square matrix}$$

- if the system is controllable, then R_n is full rank, i.e., invertible.

2. Calculate R_n^{-1}

3. Take the last row of R_n^{-1} — call this last row \vec{q}^T (\vec{q} is a col. vector; \vec{q}^T a row vector)

4. Form the following matrix, row by row:

$$T \stackrel{\triangle}{=} \begin{Bmatrix} \leftarrow \vec{q}^T \rightarrow \\ \leftarrow \vec{q}^T \vec{A} \rightarrow \\ \leftarrow \vec{q}^T \vec{A}^2 \rightarrow \\ \vdots \\ \leftarrow \vec{q}^T \vec{A}^{n-1} \rightarrow \end{Bmatrix} \begin{array}{l} \leftarrow \text{this is an } n \times n \text{ matrix} \\ \leftarrow \text{it will be } \underline{\text{full rank and invertible}} \end{array}$$

5. Define $\vec{z} = T\vec{x} \Leftrightarrow \vec{x} = T^{-1}\vec{z}$ (this is called a basis transformation from \vec{x} to \vec{z} , using T)

6. Using the definition of \vec{z} in $\frac{d\vec{x}}{dt} = \vec{A}\vec{x} + \vec{b}u(t)$, you get, equivalently,

$$\frac{d\vec{z}}{dt} = \underbrace{T\vec{A}T^{-1}\vec{z}}_{\hat{\vec{A}}} + \underbrace{T\vec{b}}_{\hat{\vec{b}}} u(t)$$

7. then:

$\rightarrow \hat{A} \triangleq TAT^{-1}$, with $\hat{b} \triangleq T\vec{b}$ will be in c.c.f!

→ Proof of the above procedure and claims:

→ from definition, $R_n^{-1}R_n = I$, or

$$\rightarrow \left[\begin{array}{c|c|c|c|c} \hline & \cdots & & & \\ \hline & \vdots & & & \\ \hline & \cdots & & & \\ \hline \leftarrow \vec{q}_V^T \rightarrow & \vec{b} & A\vec{b} & A^2\vec{b} & \cdots A^{n-1}\vec{b} \\ \hline \end{array} \right] \left[\begin{array}{c|c|c|c|c} \hline & | & | & | & | \\ \hline | & | & | & | & | \\ \hline | & | & | & | & | \\ \hline \end{array} \right] = \left[\begin{array}{ccccc} 1 & 0 & & 0 & 0 \\ 0 & 1 & & \ddots & \vdots \\ \vdots & \ddots & \ddots & \ddots & 0 \\ 0 & 0 & & 0 & 1 \\ \hline \end{array} \right]$$

$$\rightarrow \text{Hence: } \vec{q}_V^T \left[\begin{array}{c|c|c|c|c} \hline & | & | & | & | \\ \hline | & | & | & | & | \\ \hline | & | & | & | & | \\ \hline \end{array} \right] = [0, 0, \dots, 0, 1]$$

$$\Rightarrow \vec{q}_V^T \vec{b} = 0, \vec{q}_V^T A\vec{b} = 0, \dots, \vec{q}_V^T A^{n-2}\vec{b} = 0$$

$$\text{and } \vec{q}_V^T A^{n-1}\vec{b} = 1$$

$$\rightarrow \text{also from definition, } T \triangleq \left[\begin{array}{c|c|c|c|c} \hline \leftarrow \vec{q}_V^T \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A^2 \rightarrow & & & & \\ \hline \vdots & & & & \\ \hline \leftarrow \vec{q}_V^T A^{n-1} \rightarrow & & & & \\ \hline \end{array} \right] \Rightarrow \left[\begin{array}{c} \vec{q}_V^T \vec{b} \\ \vec{q}_V^T A\vec{b} \\ \vec{q}_V^T A^2\vec{b} \\ \vdots \\ \vec{q}_V^T A^{n-1}\vec{b} \end{array} \right] = \overset{\vec{b}}{\underset{T\vec{b}}{\sim}} = \left[\begin{array}{c} 0 \\ 0 \\ \vdots \\ 0 \\ 1 \end{array} \right]$$

→ Consider TA :

$$\rightarrow TA : \left[\begin{array}{c|c|c|c|c} \hline \leftarrow \vec{q}_V^T \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A^2 \rightarrow & & & & \\ \hline \vdots & & & & \\ \hline \leftarrow \vec{q}_V^T A^{n-1} \rightarrow & & & & \\ \hline \end{array} \right] A = \left[\begin{array}{c|c|c|c|c} \hline \leftarrow \vec{q}_V^T A \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A^2 \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A^3 \rightarrow & & & & \\ \hline \vdots & & & & \\ \hline \leftarrow \vec{q}_V^T A^{n-1} \rightarrow & & & & \\ \hline \leftarrow \vec{q}_V^T A^n \rightarrow & & & & \\ \hline \end{array} \right]$$

→ notice that the top $(n-s)$ rows of T are just the last $(n-1)$ rows of T ,
i.e., the latter are shifted up by 1.

→ The last entry, $\tilde{q}^T A^n$, can be expressed as a linear combination of the others, using the Cayley-Hamilton Theorem

→ C-H. Thm: A satisfies its own characteristic polynomial.

→ i.e., if $C(\lambda) = \lambda^n + a_{n-1}\lambda^{n-1} + \dots + a_1$ is the char. poly. of A ,

then $A^n + a_{n-1}A^{n-1} + \dots + a_1 I = 0$, or

$$A^n = -a_n A^{n-1} - a_{n-1} A^{n-2} - \dots - a_1 I$$

$$\rightarrow \tilde{q}^T A^n = -a_n \tilde{q}^T A^{n-1} - a_{n-1} \tilde{q}^T A^{n-2} - \dots - a_1 \tilde{q}^T A - a_1 \tilde{q}^T$$

→ These two observations can be encapsulated in matrix form as:

$$\rightarrow TA = \begin{bmatrix} \leftarrow \tilde{q}^T A \rightarrow \\ \leftarrow \tilde{q}^T A^2 \rightarrow \\ \leftarrow \tilde{q}^T A^3 \rightarrow \\ \vdots \\ \leftarrow \tilde{q}^T A^{n-1} \rightarrow \\ \leftarrow \tilde{q}^T A^n \rightarrow \end{bmatrix} = \begin{bmatrix} 0 & 1 & & & & \\ 0 & 0 & 1 & & & \\ \vdots & \vdots & \ddots & \ddots & & \\ & & & 0 & 1 & \\ & & & & 0 & 1 \\ -a_1 & -a_2 & \dots & a_{n-1} & a_n \end{bmatrix} \begin{bmatrix} \leftarrow \tilde{q}^T \rightarrow \\ \leftarrow \tilde{q}^T A \rightarrow \\ \leftarrow \tilde{q}^T A^2 \rightarrow \\ \vdots \\ \leftarrow \tilde{q}^T A^{n-1} \rightarrow \\ \leftarrow \tilde{q}^T A^n \rightarrow \end{bmatrix}$$

call this \hat{A}

$$\rightarrow \text{thus we have } TA = \hat{A} T \text{ or } \hat{A} = T A T^{-1}$$

→ and (\hat{A}, \hat{b}) are in c.c.f.

→ A note about a fallacious proof:

→ in some places, a procedure is proposed that:

1. starts from some (\hat{A}, \hat{b}) in c.c.f. (with a_1, \dots, a_n arbitrary)

2. builds $\hat{R}_n \triangleq [\hat{A}^{n-1}\hat{b}, \hat{A}^{n-2}\hat{b}, \dots, \hat{b}]$ and shows that it is lower triangular, with
Is on the diagonal,

3. defines $R_n \triangleq [A^{n-1}\bar{b}, A^{n-2}\bar{b}, \dots, \bar{b}]$

4. defines $T = \hat{R}_n R_n^{-1}$

and 5. claims that $\hat{A} = T A T^{-1}$

because step 5 is NOT TRUE IN GENERAL.

→ This procedure DOES NOT PROVE that controllability of $(A, b) \Rightarrow (T A T^{-1}, T \bar{b})$ is in c.c.f.

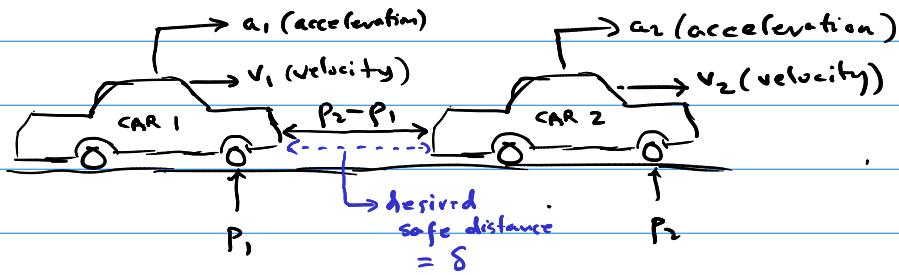
- Why are the above results useful?
- Because they tell us that if a system is controllable, we can always devise feedback to place its eigenvalues wherever we want.
- but we don't necessarily have to put it in c.c.f first to do eigenvalue placement.

- we can do it directly - ie, writing out $C(s) \triangleq \det(A - \tilde{b}k^T - \lambda I)$

and matching its coefficients w those of $(\lambda - \lambda_1) \dots (\lambda - \lambda_n)$

↳ leads to a linear system of eqns in the unknowns k_1, \dots, k_n ,
 which can be solved ^(eg.) numerically.

Example: co-operative adaptive cruise control



CAR 1

$$v_1 = \frac{dp_1}{dt}$$

$$a_1 = \frac{d^2p_1}{dt^2}$$

CAR 2

$$v_2 = \frac{dp_2}{dt}$$

$$a_2 = \frac{d^2p_2}{dt^2}$$

$$\frac{d}{dt}(p_2 - p_1) = v_2 - v_1,$$

call this $x_1 + s$, or $x_1 = p_2 - p_1 - s$

$$\frac{d}{dt}(v_2 - v_1) = a_2 - a_1,$$

x_2

will be 0 if cars at desired safe distance

$$\frac{d\vec{x}}{dt} = \underbrace{\begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix}}_A \vec{x} + \underbrace{\begin{bmatrix} 0 \\ 1 \end{bmatrix}}_b u(t)$$

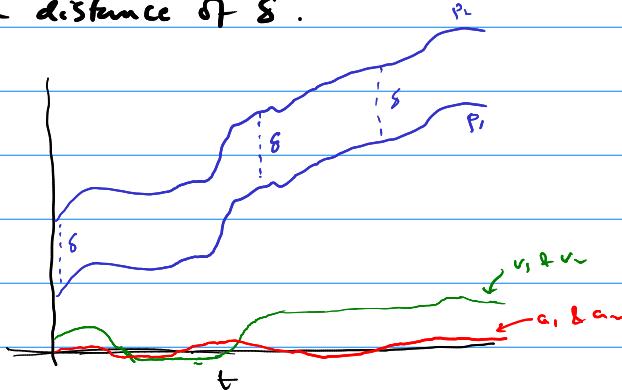
— Suppose, at $t=0$: $x_1=0 \Rightarrow p_2 - p_1 = \delta$

$$x_2 = 0 \Rightarrow v_2 = v_1 \leftarrow \text{same velocities}$$

→ and car 1's driver (somehow) perfectly matches car 2's acceleration : i.e., $a_1(t) \equiv a_2(t) \Rightarrow u(t) = 0$.

→ then $\frac{d\vec{x}}{dt} = 0$, and nothing changes.

→ they both go at the same velocities, keeping a distance of δ .



— Now, suppose there is a small acceleration error made by the driver of car 1: $a_1(t) = a_2(t) + \Delta a(t) \rightarrow u(t) = \Delta a(t)$

→ solving the system: we have

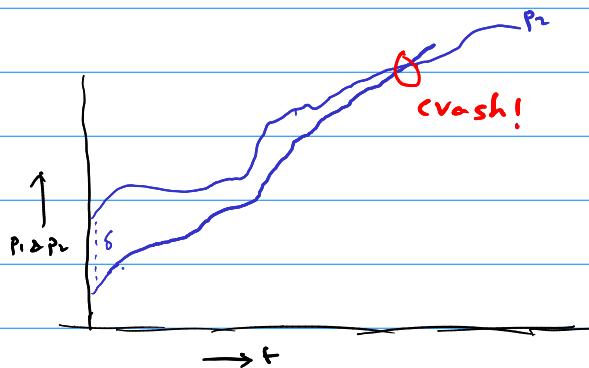
$$\rightarrow \Delta v(t) = v_2(t) - v_1(t) = \int_0^t \Delta a(z) dz$$

$$\rightarrow \Delta p(t) = p_2(t) - p_1(t) - \delta = \int_0^t \Delta v(z) dz$$

→ example: $\Delta a = \text{small constant } \epsilon$

$$\Rightarrow \Delta v(t) = \epsilon t, \text{ and } \Delta p(t) = \frac{\epsilon t^2}{2}$$

→ keeps increasing without bounds; cars will hit each other if $\epsilon > 0$



— How can feedback help?

$$\rightarrow A \mapsto \begin{bmatrix} 0 & 1 \\ -k_1 & -k_2 \end{bmatrix}; c(\lambda) = \lambda(\lambda + k_2) + k_1 = 0 \Rightarrow \lambda^2 + k_2\lambda + k_1 = 0$$

position feedback velocity feedback

Doppler radar provides both.

$$\lambda_{1,2} = -\frac{k_2}{2} \pm \frac{1}{2}\sqrt{k_2^2 - 4k_1}$$

→ make $k_2 > 0, k_1 > 0$

→ $\lambda_{1,2}$ both will have -ve real parts

$$\begin{bmatrix} 0 & 1 \\ -k_1 & -k_2 \end{bmatrix} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \lambda \begin{bmatrix} x_1 \\ x_2 \end{bmatrix}$$

$$x_2 = \lambda x_1$$

$$-(k_1 x_1 + k_2 x_2) = \lambda x_2 \quad \rightarrow 0$$

$$\dot{x} = Ax + Bu \quad \Rightarrow \quad -k_1 x_1 - k_2 \lambda x_1 = \lambda^2 x_1 \Rightarrow (\lambda^2 + k_2 \lambda + k_1) x_1 = 0$$

$$\dot{\vec{x}} = A\vec{x} + \vec{bu}$$

$$\dot{\vec{x}} = \vec{P} \vec{A} \vec{P}^{-1} \vec{x} + \vec{P} \vec{b}$$

$$\vec{x} = P^{-1} \vec{x} \Leftrightarrow \vec{x} = \vec{P} \vec{x} \quad \checkmark$$

$$P = \begin{bmatrix} 1 & 1 \\ \lambda_1 & \lambda_2 \end{bmatrix}; P^{-1} = \frac{1}{\lambda_2 - \lambda_1} \begin{bmatrix} \lambda_2 & -1 \\ -\lambda_1 & 1 \end{bmatrix}$$

say $x_1 = 1$

$$\begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ \lambda_1 & \lambda_2 \end{bmatrix} \begin{bmatrix} \int_0^t e^{\lambda_1(t-z)} \Delta a(z) dz \\ \lambda_2 \int_0^t e^{\lambda_2(t-z)} \Delta a(z) dz \end{bmatrix}$$

if $\Delta a(t) \equiv \epsilon$

$$\begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} = \begin{bmatrix} 1 & 1 \\ \lambda_1 & \lambda_2 \end{bmatrix} \begin{bmatrix} \frac{\epsilon}{\lambda_1} [1 - e^{\lambda_1 t}] \\ \epsilon \in [1 - e^{\lambda_2 t}] \end{bmatrix}$$

→ always in the range $[0, \epsilon/\lambda_1]$

$$\int_0^t e^{\lambda_1(t-z)} dz = \frac{1}{\lambda_1} [1 - e^{\lambda_1 t}]$$

→ always in the range $[0, \epsilon]$

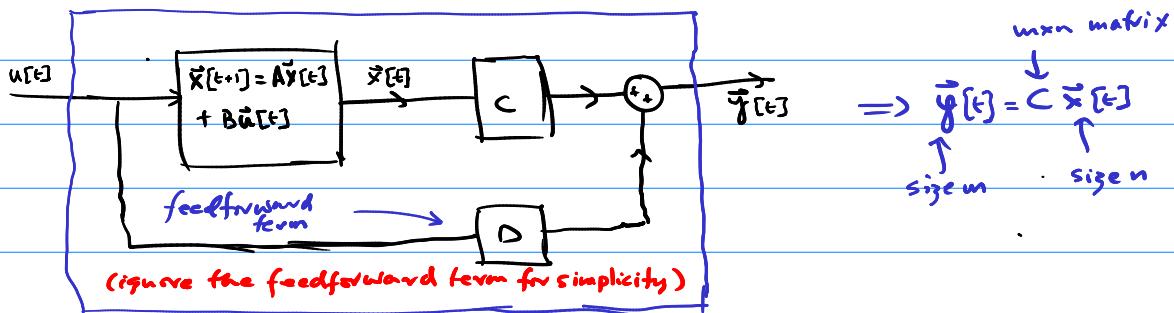
$$\text{hence } x_1(t) = \frac{\epsilon}{\lambda_1} [1 - e^{\lambda_1 t}] + \epsilon \in [1 - e^{\lambda_1 t}, \frac{\epsilon}{\lambda_1}]$$

↑ make λ_1 large
this $\rightarrow 1$

OBSERVABILITY AND OBSERVERS

- So far, we have mostly been treating $\vec{x}(t)$ as the output $\vec{y}(t)$
- but in most practical situations, the outputs are different from the state.
- recall our output equation: $\vec{y}[t] = C \vec{x}[t] + D \vec{u}[t]$

(BACK TO DISCRETE SYSTEMS)



- Q: from measurements of the output (and knowing the input)
can you infer the state? (especially if $m < n$)
- If yes, the system (A, B, C) is called OBSERVABLE.

→ Say, $m=1$ (just one output), while $\vec{x} \in \mathbb{R}^n$, $n > 1$

$$y[t] = \underbrace{\vec{c}^\top}_{[c_1, c_2, \dots, c_n]} \vec{x}[t] \rightarrow 1 \times n$$

$$y[0] = \vec{c}^\top \vec{x}[0]$$

$$y[1] = \vec{c}^\top \vec{x}[1] = \vec{c}^\top (A \vec{x}[0] + B \vec{u}[0])$$

$$y[2] = \vec{c}^\top \vec{x}[2] = \vec{c}^\top (A^2 \vec{x}[0] + A B \vec{u}[0] + B^2 \vec{u}[1])$$

$$y[i] = \vec{c}^\top \vec{x}[i] = \vec{c}^\top (A^i \vec{x}[0] + \sum_{j=1}^{i-1} (A^{i-j} B \vec{u}[j-1]))$$

$$y[n] = \vec{c}^\top \vec{x}[n] = \vec{c}^\top (A^n \vec{x}[0] + \sum_{j=1}^n (A^{n-j} B \vec{u}[j-1]))$$

→ Now, say $u[t] \equiv 0$ — i.e., no input.

→ might you still be able to recover $\vec{x}[0], \vec{x}[1], \text{etc.}$ from $y[t]$?

→ boils down to recovery of just $\vec{x}[0]$, from which

$\vec{x}[1], \vec{x}[2], \text{etc.}$
can be recovered

$$\begin{aligned} y[0] &= \vec{c}^T \vec{x}[0] \\ y[1] &= \vec{c}^T \vec{x}[1] = \vec{c}^T A \vec{x}[0] \\ y[2] &= \vec{c}^T \vec{x}[2] = \vec{c}^T A^2 \vec{x}[0] \end{aligned}$$

$$\rightarrow \begin{bmatrix} y[0] \\ y[1] \\ \vdots \\ y[n] \end{bmatrix} = \begin{bmatrix} \leftarrow \vec{c}^T \rightarrow \\ \leftarrow \vec{c}^T A \rightarrow \\ \leftarrow \vec{c}^T A^2 \rightarrow \\ \vdots \\ \leftarrow \vec{c}^T A^{n-1} \rightarrow \end{bmatrix} \vec{x}[0]$$

↙ observability matrix O

$$\begin{aligned} y[i] &= \vec{c}^T \vec{x}[i] = \vec{c}^T A^i \vec{x}[0] \\ y[n] &= \vec{c}^T \vec{x}[n] = \vec{c}^T A^n \vec{x}[0] \end{aligned}$$

→ if O is full rank (invertible), then yes, otherwise no.

$$\vec{x}[0] = O^{-1} \begin{bmatrix} y[0] \\ \vdots \\ y[n] \end{bmatrix}$$

→ more generally, if $m > 1$ (several outputs), $\vec{y} = C \vec{x}$

$$O = km \times n$$

← must be full rank, i.e., rank = n

always

→ Observability boils down to finding just the IC, which we don't know

→ we know $A, B, C, u[t]$ and $y[t]$

↙ CLARIFY THIS POINT IN THE BEGINNING.

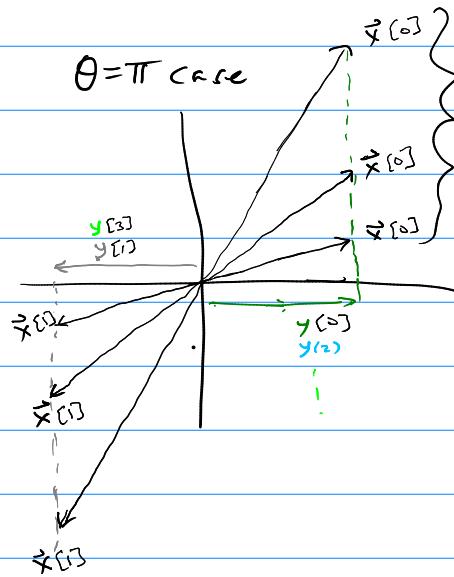
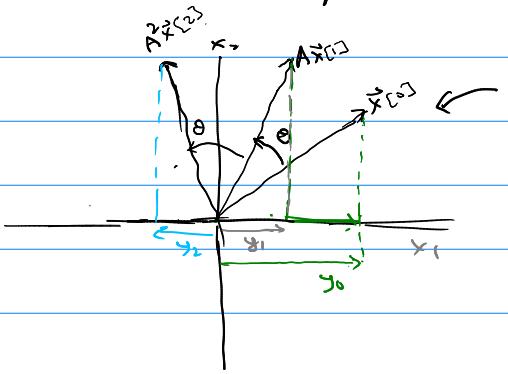
this is a rotation matrix

$$\text{EXAMPLE : } \begin{bmatrix} x_1[t+1] \\ x_2[t+1] \end{bmatrix} = \underbrace{\begin{bmatrix} \cos(\theta) & -\sin(\theta) \\ \sin(\theta) & \cos(\theta) \end{bmatrix}}_A \begin{bmatrix} x_1[t] \\ x_2[t] \end{bmatrix}$$

$$y[t] = x_1[t] = \underbrace{\begin{bmatrix} 1 & 0 \\ 0 & 1 \end{bmatrix}}_{\in \mathbb{C}^T} \vec{x}[t]$$

$$\begin{bmatrix} \leftarrow C \rightarrow \\ \leftarrow C^T A \rightarrow \end{bmatrix} = \begin{bmatrix} 1 & 0 \\ \cos(\theta) & -\sin(\theta) \end{bmatrix} : \begin{array}{l} \text{rank} = 2 \text{ if } \sin(\theta) \neq 0 \\ \text{rank} = 1 \text{ if } \theta = 0, \pi, 2\pi, \dots \end{array}$$

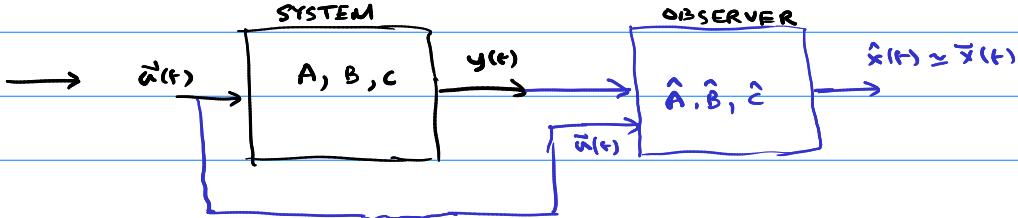
Graphical interpretation



Any of these could
(be $\vec{x}(0)$), matching
measured y_0 and
the rotation matrix

OBSERVERS

→ A NEW SYSTEM (that you set up) TO ESTIMATE $\vec{x}(t)$ from the output



→ Form of observer: $\hat{x}[t+1] = A\hat{x}[t] + B\bar{u}[t] + L(C\hat{x}[t] - \bar{y}[t])$

$$L(C\hat{x}[t] - \bar{y}[t])$$

if $\hat{x}[t] = \bar{x}[t]$, then
this term will be 0

and the observer
will be just the original
system.

— How good is the observer at doing its job (estimating $\vec{x}(t)$ well)?

— Define the error $\vec{e}(t) \triangleq \hat{x}[t] - \vec{x}[t]$

— then $\vec{e}[t+1] = \hat{x}[t+1] - \vec{x}[t+1]$

$$= A(\hat{x}[t] - \vec{x}[t]) + LC(\hat{x}[t] - \vec{x}[t])$$

$$\Rightarrow \vec{e}[t+1] = A\vec{e}[t] + LC\vec{e}[t] = (A + LC)\vec{e}[t]$$

→ we would like $\vec{e}[t]$ to go to 0

⇒ choose L to make the eigenvalues of $(A + LC)$ stable!

→ exactly like feedback for controllability!

→ recall: $(A + BK)$ for controllability.

→ you can re-use all that algebra and results: $\begin{matrix} \text{treat like } -B \\ \downarrow \quad \downarrow \quad \downarrow \\ \text{e.v.s of } (A + LC) = \text{e.v.s of } (A + LC)^T = A^T + C^T L^T \end{matrix}$

→ i.e., can always place eigenvalues if $(A^T, -C^T)$ controllable

⇒ $[(A^T)^{-1}, (A^T)^{-2}, \dots, A^T C^T, C^T]$ has full rank

$\Rightarrow [(\mathbf{A}^T)^{k-1}, (\mathbf{A}^T)^{k-2}, \dots, \mathbf{A}^T \mathbf{C}^T, \mathbf{C}^T]^T$ has full rank

$$\hookrightarrow = \begin{bmatrix} - & \mathbf{C} \mathbf{A}^{k-1} & - \\ - & \mathbf{C} \mathbf{A}^{k-2} & - \\ \vdots & \vdots & \vdots \\ - & \mathbf{C} \mathbf{A} & - \\ - & \mathbf{C} & - \end{bmatrix} \quad \text{Simplify the Observability criterion for the system}$$

Example: same rotation matrix example

$$- \begin{bmatrix} x_1[t+1] \\ x_2[t+1] \end{bmatrix} = \underbrace{\begin{bmatrix} \cos(\theta) & -\sin(\theta) \\ \sin(\theta) & \cos(\theta) \end{bmatrix}}_A \begin{bmatrix} x_1[t] \\ x_2[t] \end{bmatrix}$$

$$y[t] = x_1[t] = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_{\tilde{c}^T} \vec{x}[t]$$

$$- \text{try } \theta = \pi/2 \text{ as a special case: } A = \begin{bmatrix} 0 & -1 \\ 1 & 0 \end{bmatrix}, \quad \tilde{c}^T = [1 \ 0]$$

$$\rightarrow A^T = \begin{bmatrix} 0 & 1 \\ -1 & 0 \end{bmatrix}, \quad (\tilde{c}^T)^T = \tilde{c} = \begin{bmatrix} 1 \\ 0 \end{bmatrix}$$

$$\rightarrow \text{e.v.s of } A^T: \quad \lambda_1 + 1 = 0 \Rightarrow \lambda_{1,2} = \pm j \quad \text{magnitude} = 1, \text{ BIBO unstable} \quad (\text{discrete-time})$$

$$\rightarrow \text{feedback: } \tilde{L} = \begin{bmatrix} l_1 \\ l_2 \end{bmatrix} \Rightarrow \tilde{c} \tilde{L}^T = \begin{bmatrix} l_1 & l_2 \\ 0 & 0 \end{bmatrix}$$

$$\Rightarrow A^T + \tilde{c} \tilde{L}^T = \begin{bmatrix} l_1 & 1+l_2 \\ -1 & 0 \end{bmatrix}$$

$$\text{eigenvalues: } (l_1 - \lambda)(-\lambda) + (1+l_2) = 0$$

$$\Rightarrow \lambda^2 - l_1 \lambda + (1+l_2) = 0$$

$$\Rightarrow \lambda_{1,2} = \frac{l_1 \pm \sqrt{l_1^2 - 4(1+l_2)}}{2}$$

$$\Rightarrow \underline{\lambda_1 + \lambda_2 = l_1} \quad \text{and} \quad \lambda_1 - \lambda_2 = \pm \sqrt{l_1^2 - 4(1+l_2)}$$

$$\Rightarrow (\lambda_1 - \lambda_2)^2 = l_1^2 - 4(1+l_2)$$

$$= (\lambda_1 - \lambda_2)^2 - (\lambda_1 + \lambda_2)^2 = -4(1+l_2)$$

$$\Rightarrow -4\lambda_1 \lambda_2 = -4(1+l_2)$$

$$\Rightarrow \lambda_2 = \lambda_1, \lambda_2 - 1$$

→ choose any 2 stable λ's (respecting complex conjugacy)
and set λ₁ & λ₂

→ what if θ = π (unobservable case)?

$$\rightarrow A = \begin{bmatrix} -1 & 0 \\ 0 & -1 \end{bmatrix}; A^T = A$$

$$A^T + \vec{C} \vec{R}^T = \begin{bmatrix} -1+\lambda_1 & \lambda_2 \\ 0 & -1 \end{bmatrix}$$

$$\det: -(λ+1)(-1+λ_1, -λ) = 0$$

$$\Rightarrow \boxed{\lambda_1 = -1}, \lambda_2 = \lambda_1 - 1$$

↳ BIBO unstable and cannot be changed by choosing \vec{C} .

→ USING OBSERVERS FOR MORE ACCURATE POSITIONING & NAVIGATION

→ recall the accelerating car

discrete-time example from the

$$\begin{bmatrix} x((t+1)T) \\ v((t+1)T) \end{bmatrix} = \begin{bmatrix} 1 & T \\ 0 & 1 \end{bmatrix} \begin{bmatrix} x(tT) \\ v(tT) \end{bmatrix} + \begin{bmatrix} \frac{T^2}{2} \\ T \end{bmatrix} a(t)$$

lecture on controllability:

— Terminology: $x[t] \stackrel{\text{discrete}}{\equiv} x(tT); v[t] \stackrel{\text{continuous}}{\equiv} v(tT); a[t] \equiv a(tT)$

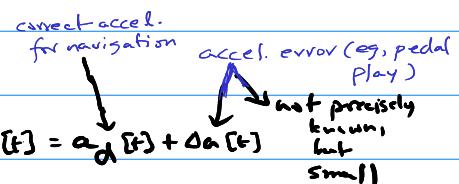
→ eigenvalues: $(-λ)^2 = Δ \Rightarrow λ_{1,2} = ±1 \rightarrow \text{BIBO UNSTABLE}$

↳ TYPICAL FOR CARS, PLANES, ETC.

→ CONSIDER the scenario:

→ the car is NOT feedback stabilized

— there is (inevitable) error in the accel: $a[t] = a_d[t] + Δa[t]$



→ we estimate its state using an observer:

$$\rightarrow \hat{x}[t+1] = A \hat{x}[t] + \vec{b} a_d[t] \rightarrow \text{same as system, no } \vec{b} \vec{C}^T \text{ term}$$

(we don't know a[t])

→ the estimation error follows:

$$\vec{e}[t+1] = A \vec{e}[t] + \vec{b} Δa[t] \leftarrow \text{unstable} \Rightarrow \vec{e}(t)$$

grows larger and larger even if Δa(t) remains small.

→ but, if we are able to measure the position of the actual car, say using GPS measurements:

$$\vec{y}[t] = \underbrace{\begin{bmatrix} 1 & 0 \end{bmatrix}}_{\vec{c}^T} \vec{x}[t] + \Delta p[t]$$

GPS measurement error,
also not precisely known, but
small.

→ Our observer becomes:

$$\vec{x}(\vec{c}^T \vec{x} - y[t])$$

$$\begin{aligned}\hat{x}[t+1] &= A \hat{x}[t] + \vec{b} \alpha_d[t] + \vec{k} (\vec{c}^T \hat{x} - y[t]) \\ &= A \hat{x}[t] + \vec{b} \alpha_d[t] + \vec{k} \vec{c}^T \hat{x}[t] - \vec{k} \vec{c}^T \vec{x}[t] \\ &\quad - \vec{k} \Delta p[t] \\ &= (A + \vec{k} \vec{c}^T) \hat{x}[t] + \vec{k} \vec{c}^T (\hat{x}[t] - \vec{x}[t]) \\ &\quad + \vec{b} \alpha_d[t] - \vec{k} \Delta p[t]\end{aligned}$$

Hence the error $\vec{e}[t] \triangleq \hat{x}[t] - \vec{x}[t]$ obeys

$$\vec{e}[t+1] = A \vec{e}[t] + \vec{k} \vec{c}^T \vec{e}[t] + \underbrace{(\vec{b} \alpha_d[t] - \vec{k} \Delta p[t])}_{\substack{\text{due to errors,} \\ \text{but small}}}$$

$$\Rightarrow \vec{e}[t+1] = (A + \vec{k} \vec{c}^T) \vec{e}[t] + \underbrace{(\vec{b} \alpha_d[t] - \vec{k} \Delta p[t])}_{\substack{\text{always small}}}$$

→ Now, if (A, \vec{c}) are observable, you can choose \vec{k} to stabilize the observer

→ which means that the estimation error can always be kept bounded and small, even taking into account small errors in GPS & acceleration

↳ this is what all navigational systems today do!

→ Actually, they use an additional improvement, whereby \vec{k} changes w/ t (i.e., becomes $\vec{k}[t]$)

→ the updates are used to minimize the error due to GPS/pedal errors, etc., etc., even further

↳ this is the famous KALMAN FILTER, the gold standard for accurate navigational estimation

— WHAT IT ACHIEVES:

- open-loop position/velocity estimation → blows up errors :: unstable
- estimation simply by GPS also has errors
- using a well-designed observer, you can make much better estimates than either of these 2 alone. ← quite remarkable
- and then, you can use this good estimate \hat{x} to feed back to the car (via controllability) to make its response stable.
↳ THIS IS WHAT IS GOING ON IN AUTONOMOUS VEHICLES

— PICTURES AND HISTORY OF KALMAN.

(TESLA PICTURE)

→ ALSO: FEEDBACK STABILIZED LEVITRON DEMOS (earlier)