EE16B, Spring 2018 UC Berkeley EECS

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Lectures 6B & 7A: Overview Slides

Controller Canonical Form Observability

• Recall prior example: $\vec{x}[t+1] = \begin{vmatrix} 0 & 1 \\ a_1 & a_2 \end{vmatrix} \vec{x}[t] + \begin{vmatrix} 0 \\ 1 \end{vmatrix} u[t]$

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- Generalization: Controller Canonical Form (CCF)

$$\bullet \ A = \begin{bmatrix} 0 & 1 & 0 & \cdots & 0 & 0 \\ 0 & 0 & 1 & \cdots & 0 & 0 \\ \vdots & & \ddots & \ddots & & \vdots \\ \vdots & & & \ddots & \ddots & \vdots \\ 0 & 0 & \cdots & \cdots & 0 & 1 \\ a_1 & a_2 & a_3 & \cdots & a_{n-1} & a_n \end{bmatrix}, \quad \vec{b} = \begin{bmatrix} 0 \\ 0 \\ \vdots \\ \vdots \\ 0 \\ 1 \end{bmatrix}$$

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- char poly: $\lambda^n a_n \lambda^{n-1} a_{n-1} \lambda^{n-2} \dots a_2 \lambda a_1$
 - not difficult to show this (though a bit tedious)
 - apply determinant formula using minors to the last row

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its roots are the eigenvalues that determine stability

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• equate coefficients against $\lambda^n - (a_n - k_n)\lambda^{n-1} - (a_{n-1} - k_{n-1})\lambda^{n-2}$ $a_n - k_n = -\gamma_n$ $-\cdots - (a_2 - k_2)\lambda - (a_1 - k_1)$

$$a_{n-1} - k_{n-1} = -\gamma_{n-1}$$

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$$\begin{array}{c} a_{n} - k_{n} = -\gamma_{n} \\ a_{n-1} - k_{n-1} = -\gamma_{n-1} \\ \vdots \\ a_{1} - k_{1} = -\gamma_{1} \end{array} \Rightarrow \begin{cases} k_{n} = \gamma_{n} - a_{n} \\ k_{n-1} = \gamma_{n-1} - a_{n-1} \\ \vdots \\ k_{1} = \gamma_{1} - a_{1} \end{cases}$$

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$$\Rightarrow \begin{cases} k_n = \gamma_n - a_n \\ k_{n-1} = \gamma_{n-1} - a_{n-1} \end{cases}$$

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We just showed: if a system is in CCF, feedback can move its eigenvalues to any desired locations

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$$-(3 - k_3) = 6$$

$$-(2 - k_2) = 11 \Rightarrow \begin{cases} k_3 = 9 \\ k_2 = 13 \\ k_1 = 7 \end{cases}$$

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 - 3. Compute its inverse: R_n^{-1}

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 - **2.** Form its controllability matrix: $R_n \triangleq \left[\vec{b}, A\vec{b}, A^2\vec{b}, \cdots, A^{n-1}\vec{b} \right]$
 - 3. Compute its inverse: R_n^{-1}
 - **4.** Grab the <u>last row</u> of R_n^{-1} : call it \vec{q}^T

- But CCF seems a very special/restrictive form ...
 - ... key question: what systems are in CCF?
- A: any controllable system can be converted to CCF!
- Here's how you do it:
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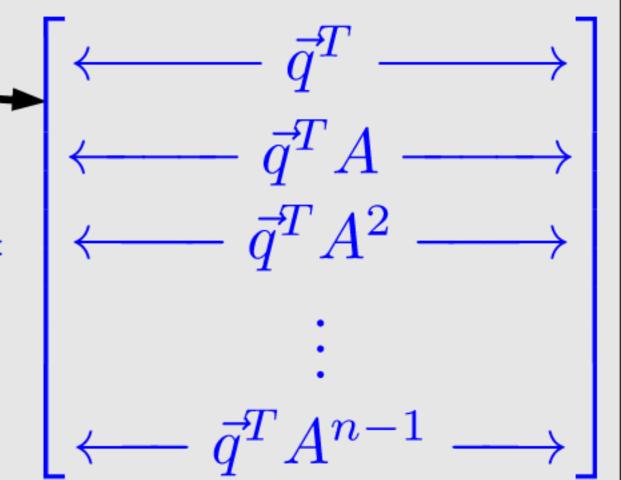
•
$$R_n^{-1} = \begin{bmatrix} \hline \\ \hline \\ \vdots \\ \hline \\ \hline \\ \hline \\ \hline \\ \hline \\ \hline \\ \end{matrix}$$
 ; (\vec{q} is a col. vector; \vec{q}^T is a row vector)



5. Form the basis transformation matrix $T \triangleq \begin{bmatrix} \longleftarrow & \vec{q}^T & \longrightarrow \\ \longleftarrow & \vec{q}^T A & \longrightarrow \\ \longleftarrow & \vec{q}^T A^2 & \longrightarrow \\ & \vdots & \\ \longleftarrow & \vec{q}^T A^{n-1} & \longrightarrow \end{bmatrix}$

T will be full rank, hence non-singular and invertible

- 5. Form the basis transformation matrix $T \triangleq$
- 6. Define $\vec{z}(t) = T\vec{x}(t) \Leftrightarrow \vec{x}(t) = T^{-1}\vec{z}(t)$

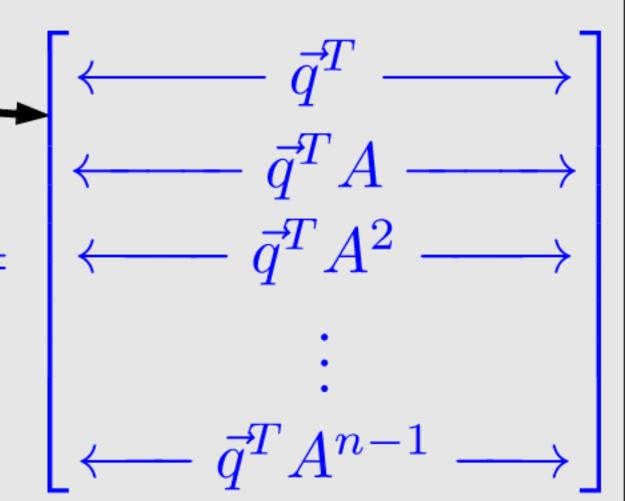


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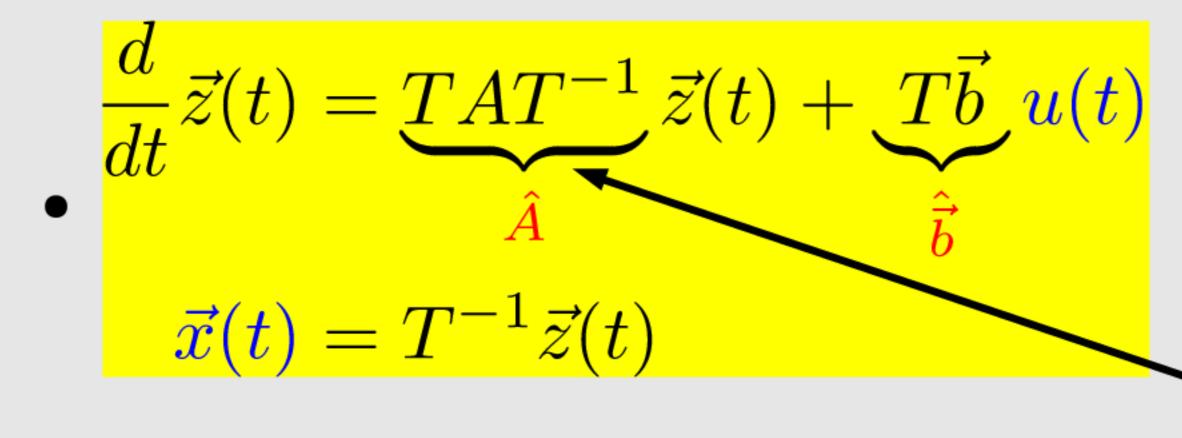
$$\frac{d}{dt}\vec{z}(t) = \underbrace{TAT^{-1}}_{\hat{A}}\vec{z}(t) + \underbrace{T\vec{b}}_{\hat{b}}u(t)$$

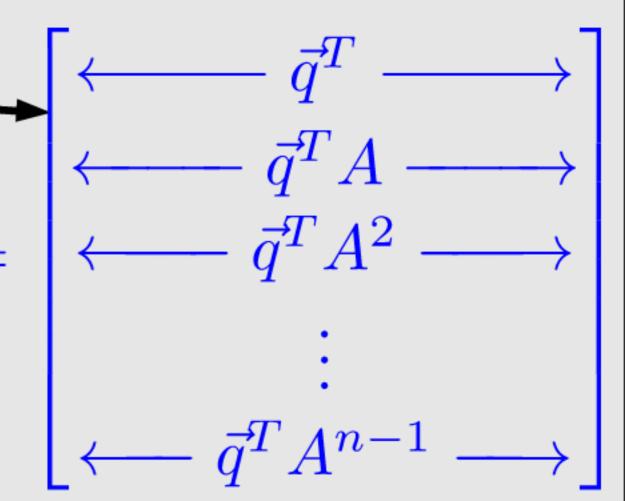
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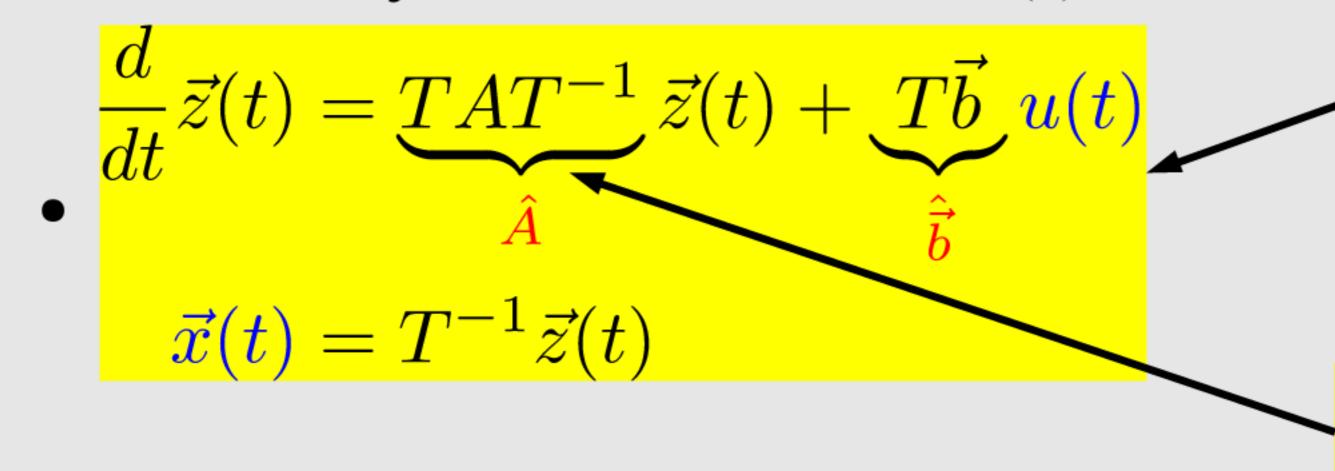




similarity transformation

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foriginal system: $u(t) \mapsto \vec{x}(t)$ is the same

equivalent to the

sımılarıty transformation

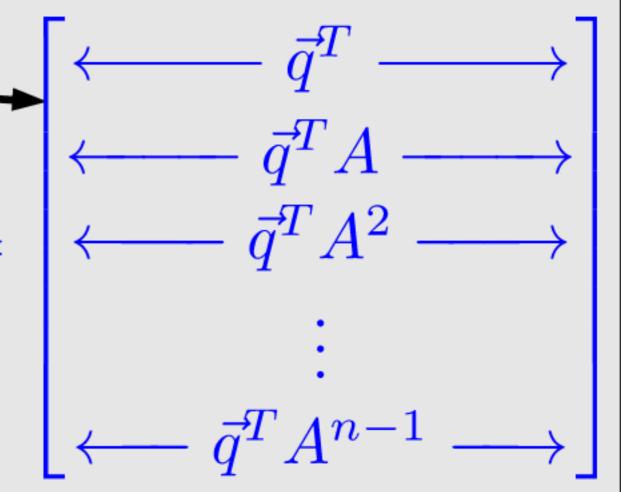
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8. (\hat{A}, \vec{b}) will be in CCF!



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similarity transformation

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 similarity

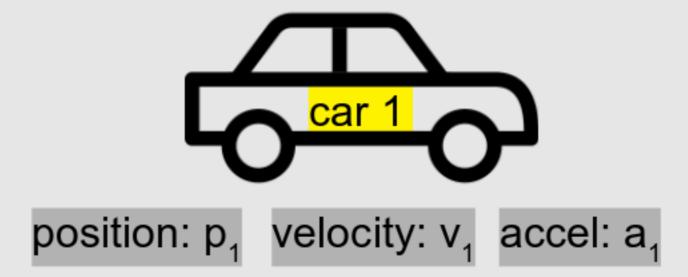
- 8. (\hat{A}, \vec{b}) will be in CCF!
- Proof: see the handwritten notes

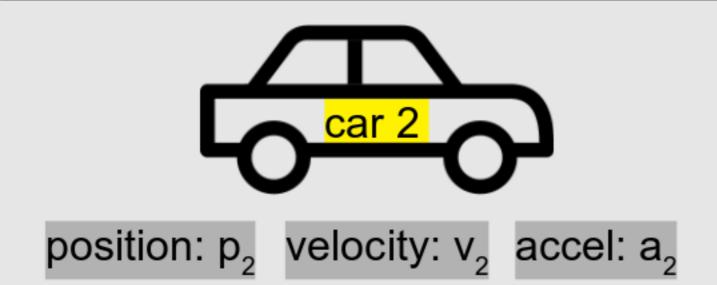
transformation

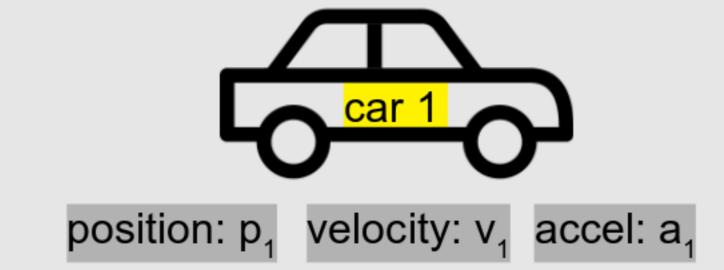


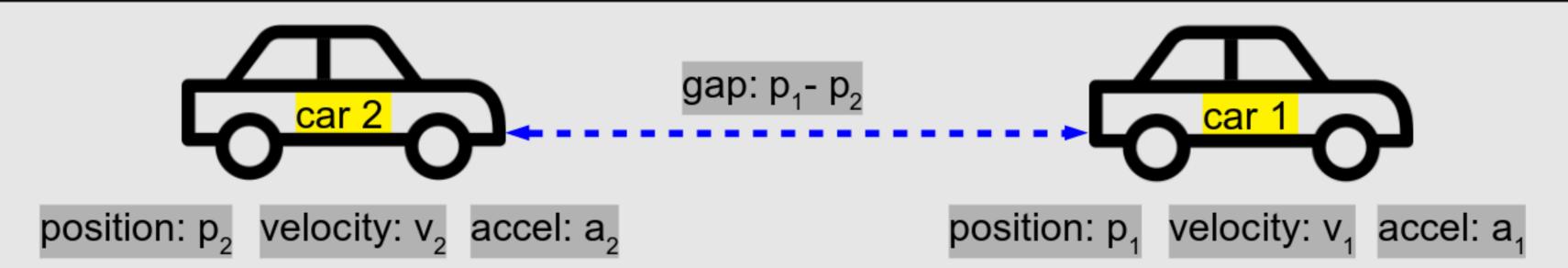


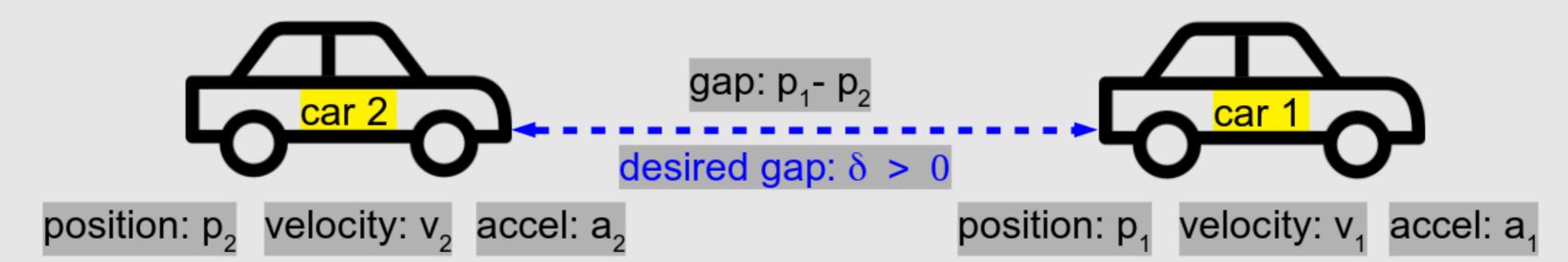


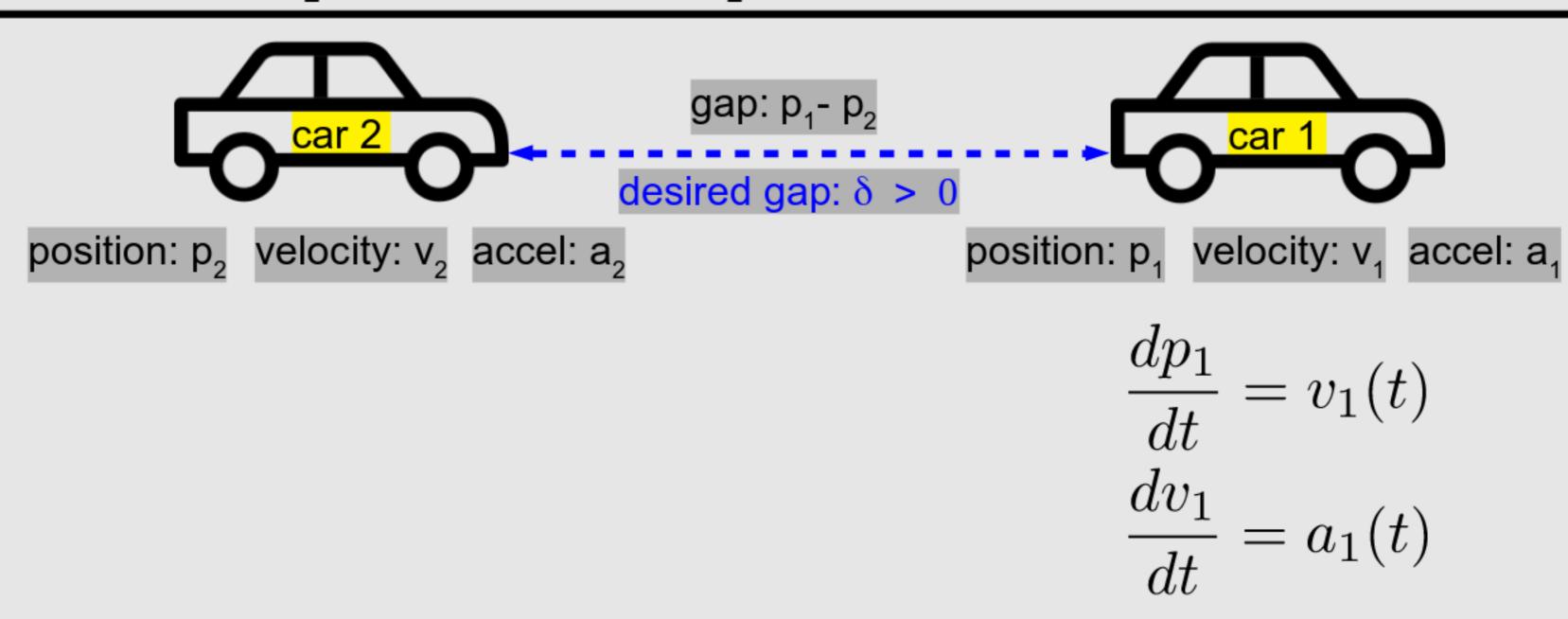


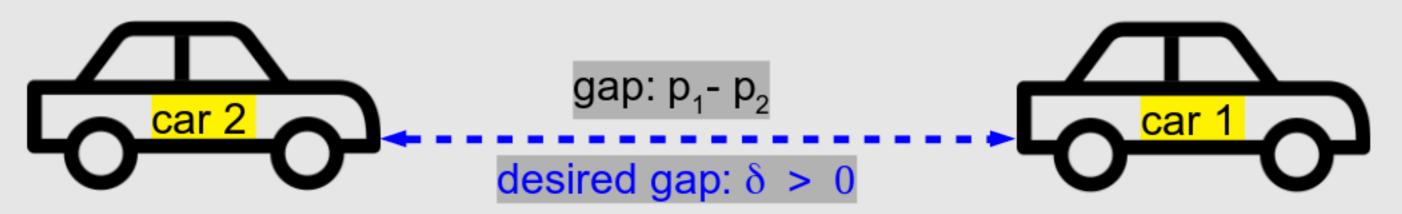












position: p, velocity: v, accel: a,

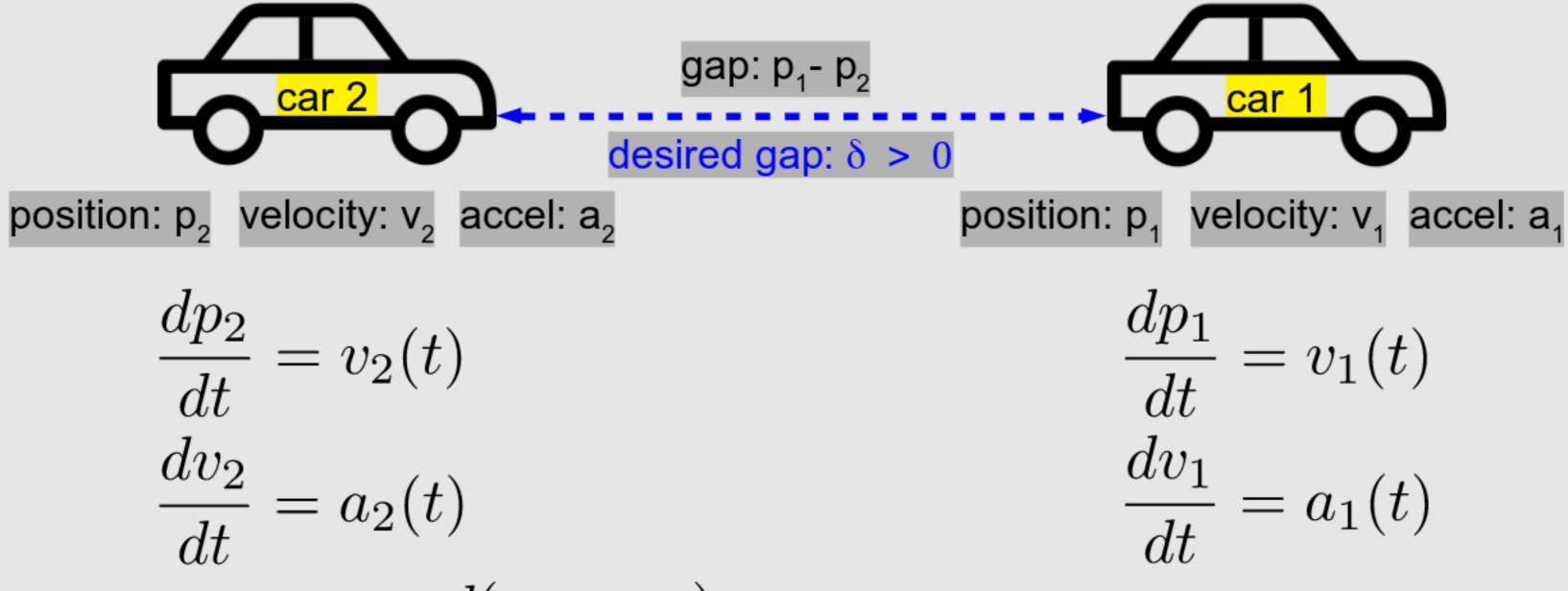
$$\frac{dp_2}{dt} = v_2(t)$$

$$\frac{dv_2}{dt} = a_2(t)$$

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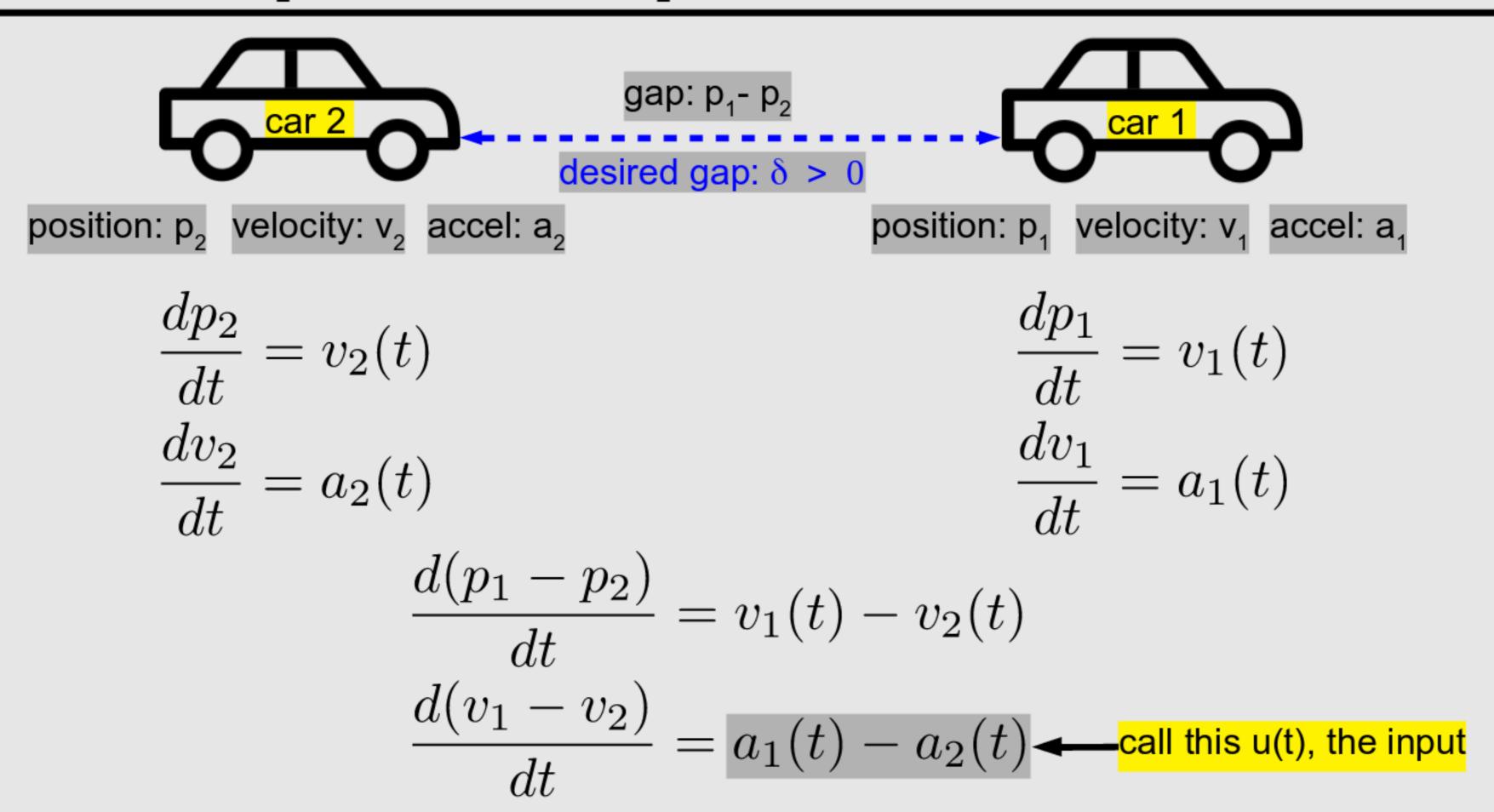
$$\frac{dp_1}{dt} = v_1(t)$$

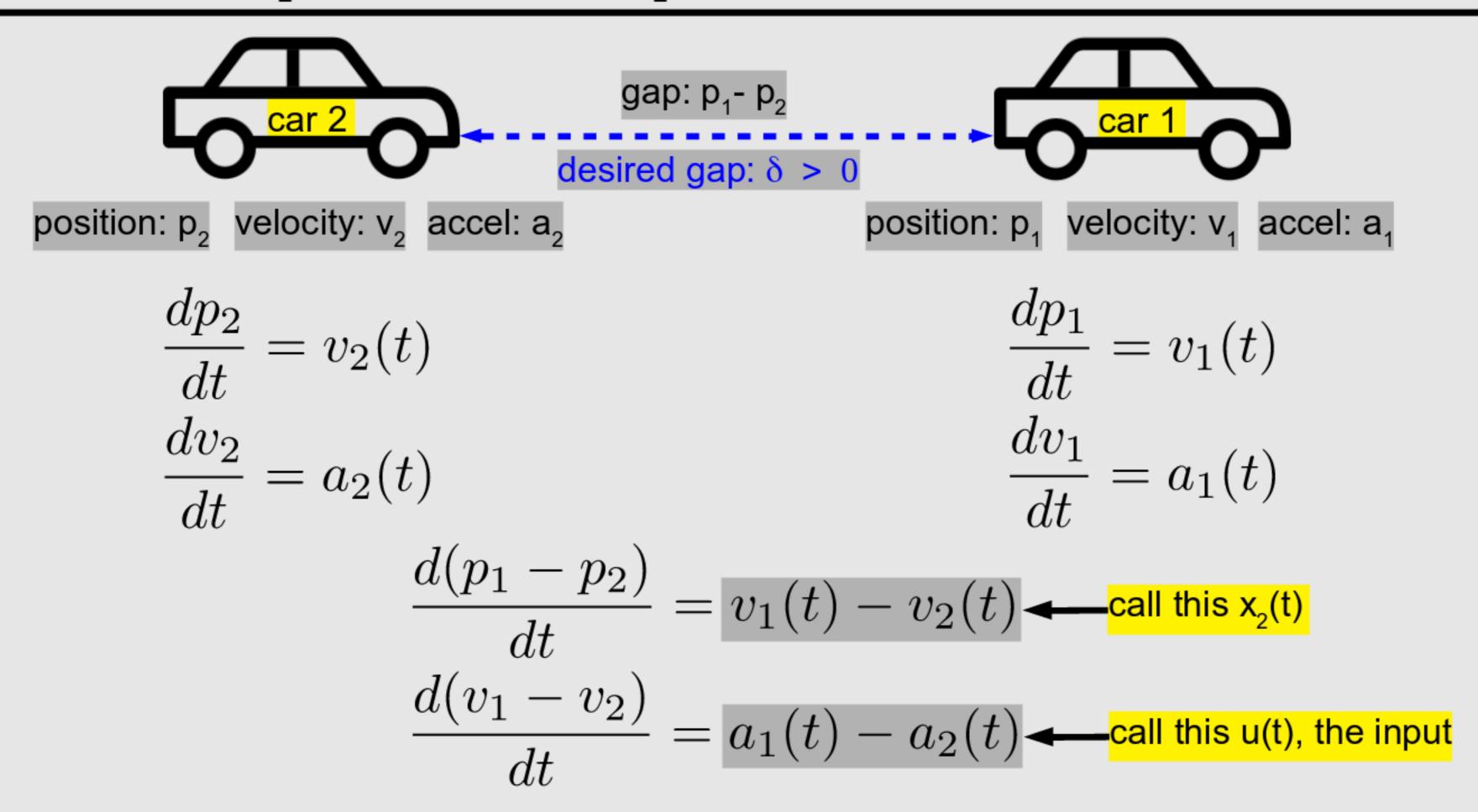
$$\frac{dv_1}{dt} = a_1(t)$$

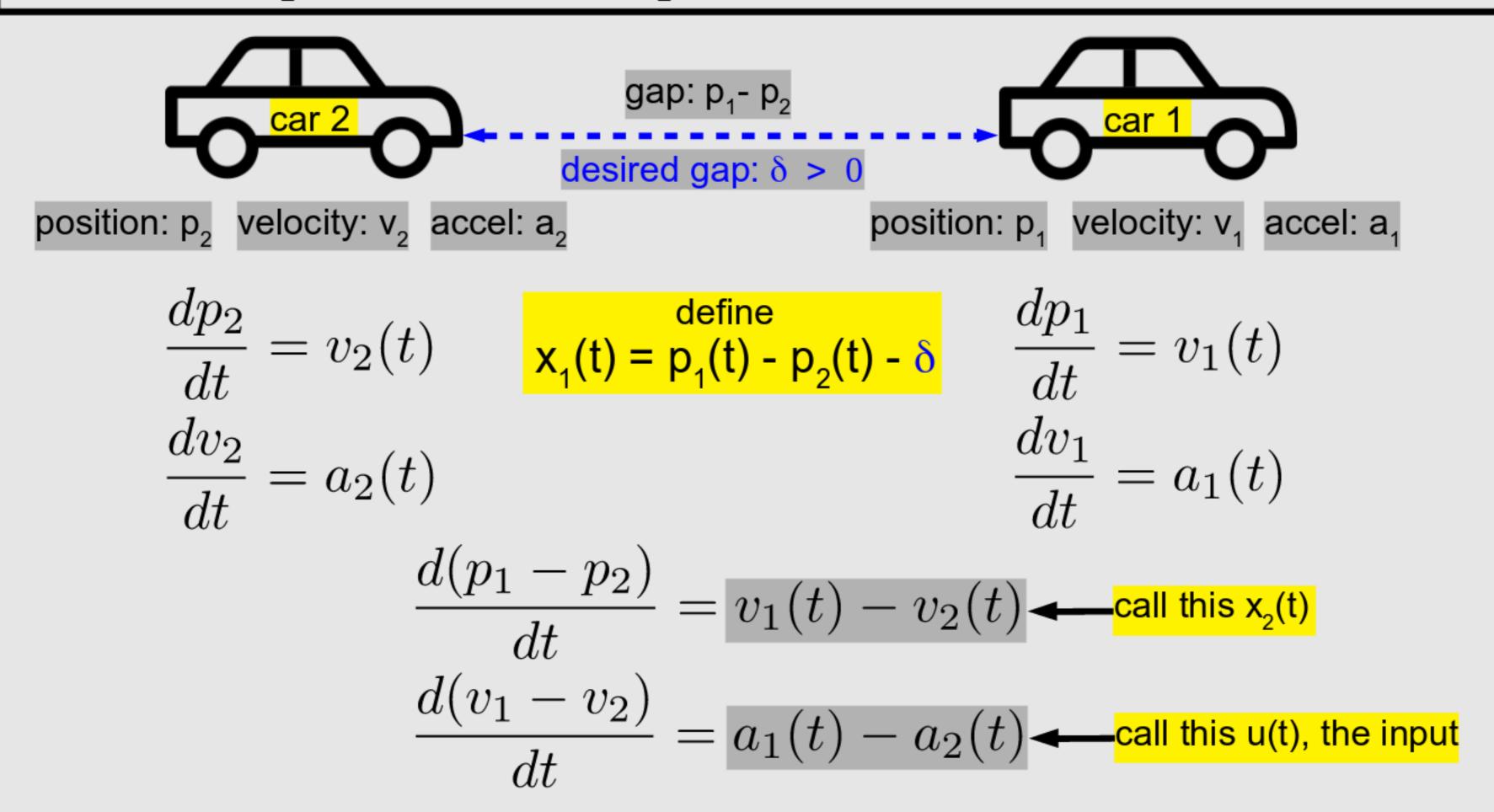


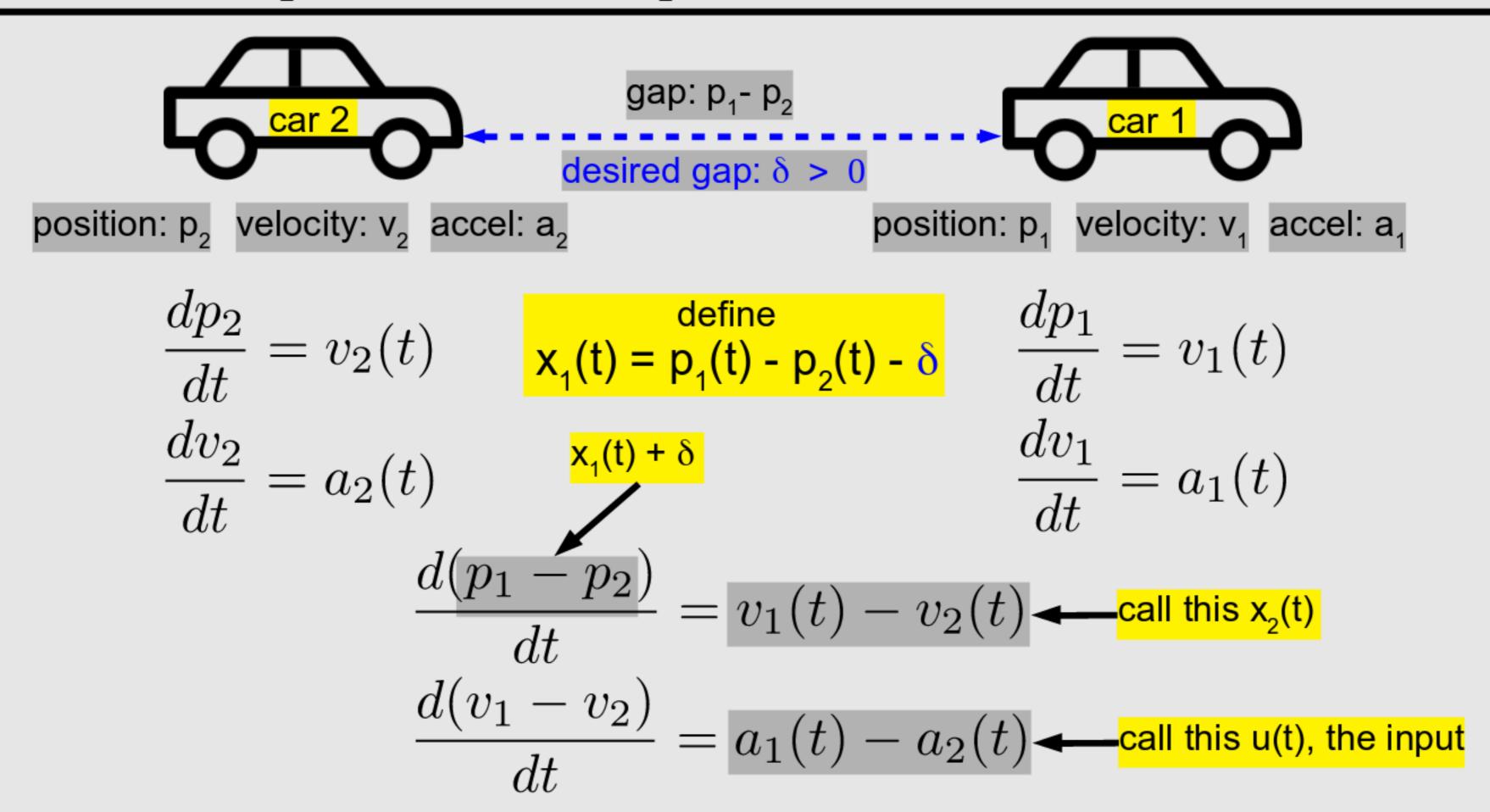
$$\frac{d(p_1 - p_2)}{dt} = v_1(t) - v_2(t)$$

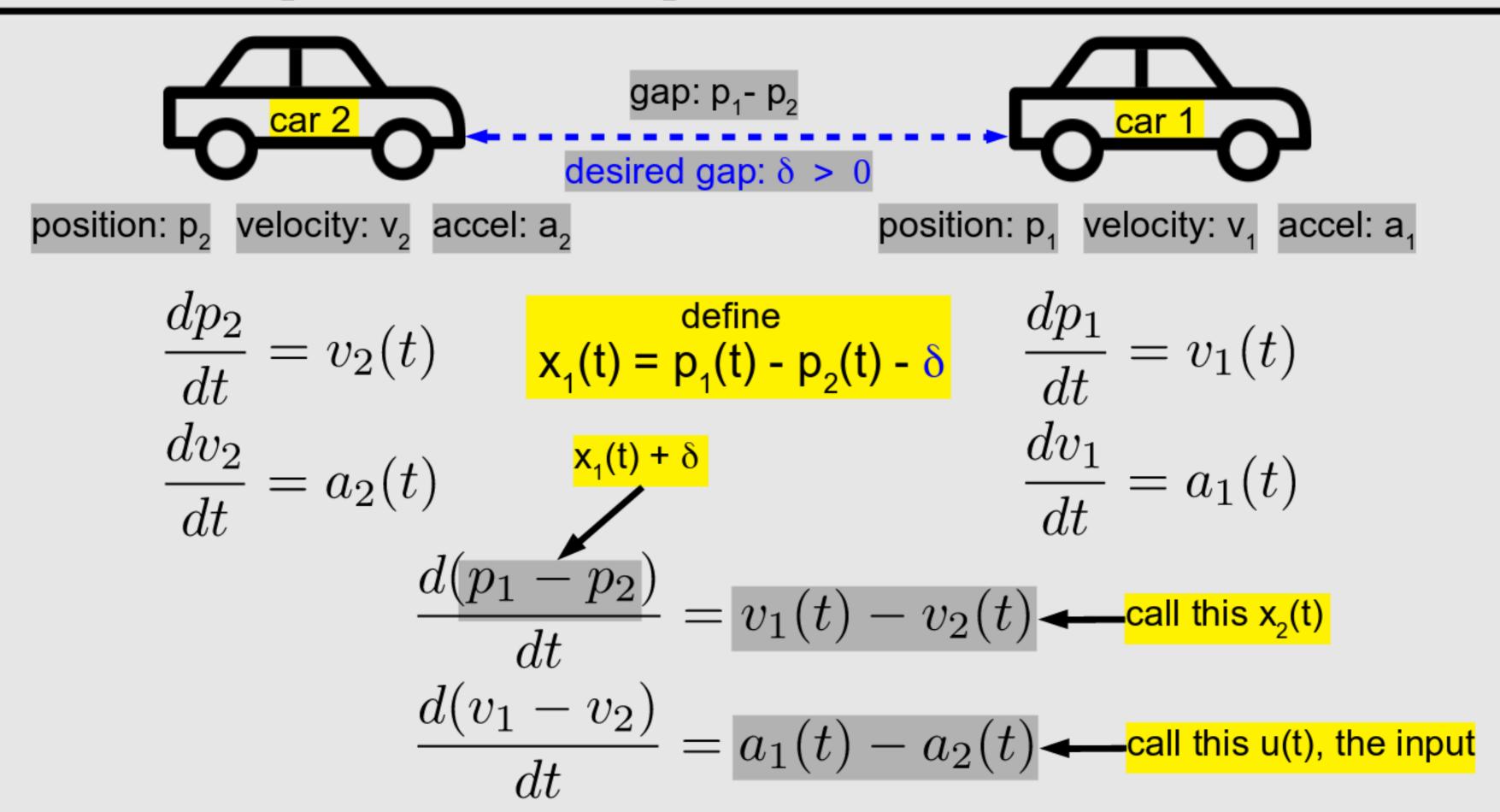
$$\frac{d(v_1 - v_2)}{dt} = a_1(t) - a_2(t)$$





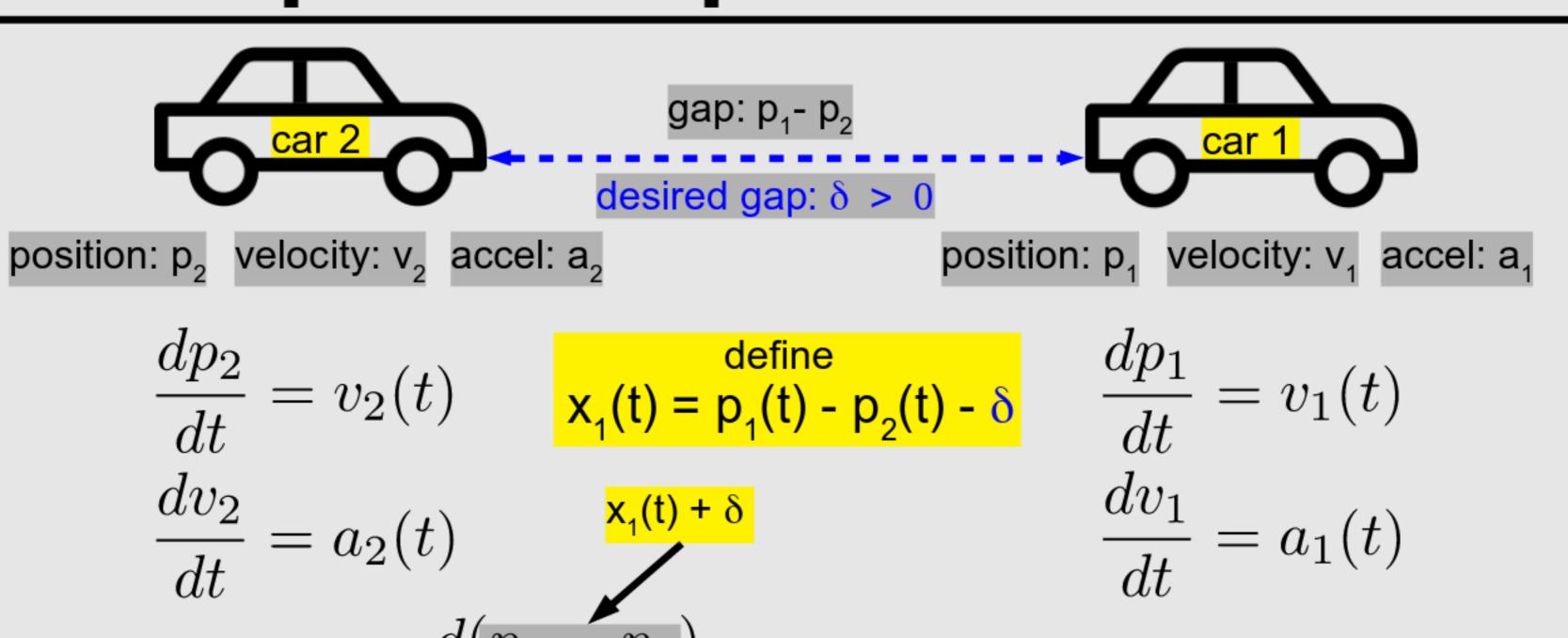






$$\frac{dx_1}{dt} = x_2(t)$$

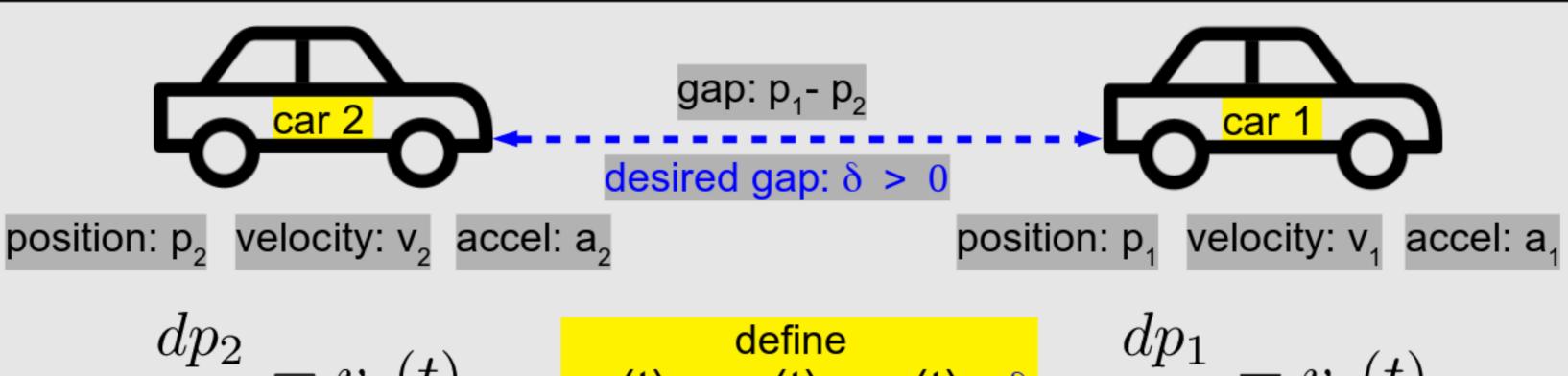
$$\frac{dx_2}{dt} = u(t)$$



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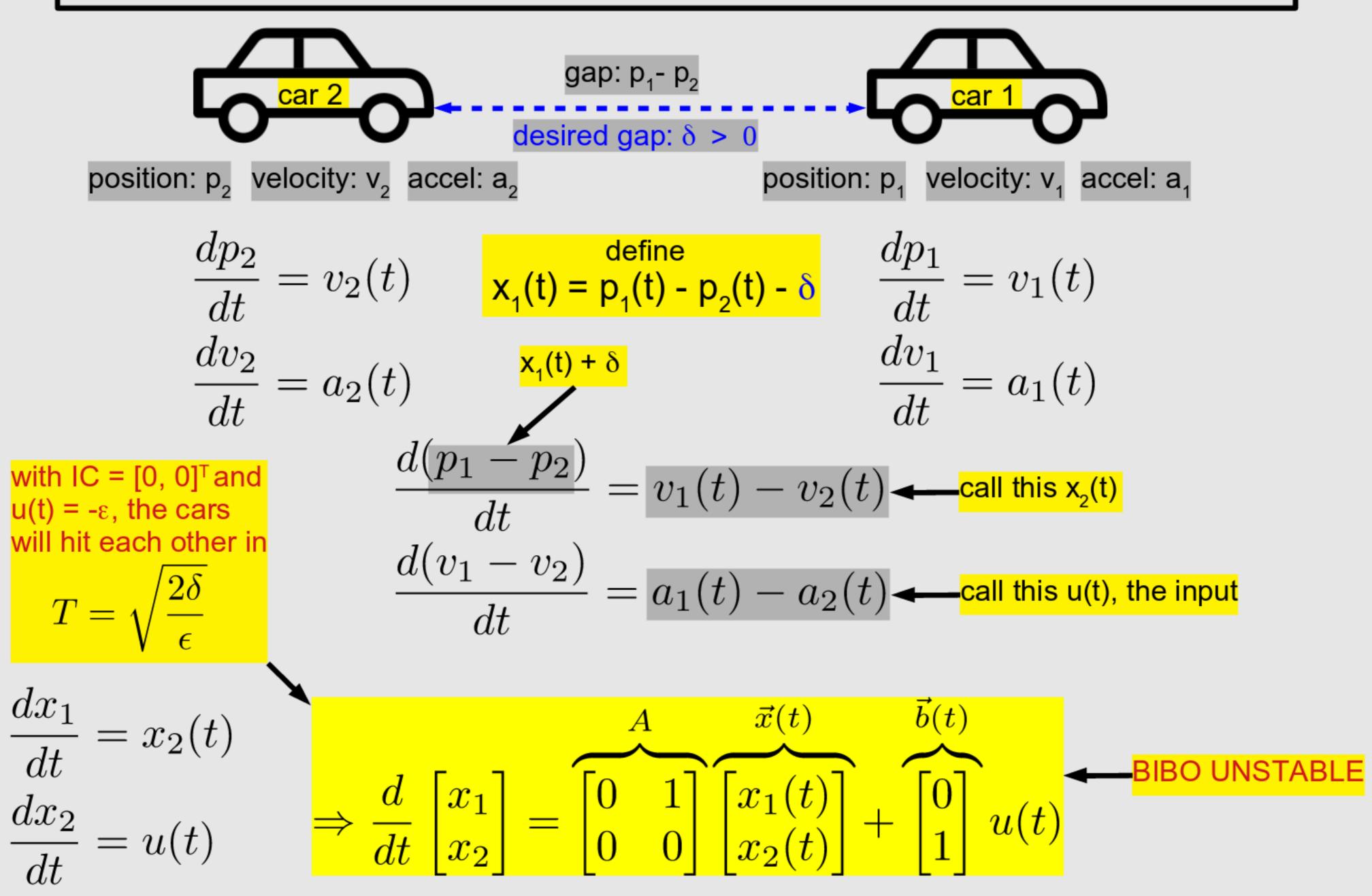
$$\Rightarrow \frac{d}{dt} \begin{bmatrix} x_1 \\ x_2 \end{bmatrix} = \begin{bmatrix} 0 & 1 \\ 0 & 0 \end{bmatrix} \begin{bmatrix} x_1(t) \\ x_2(t) \end{bmatrix} + \begin{bmatrix} 0 \\ 1 \end{bmatrix} u(t)$$



$$\begin{split} \frac{dp_2}{dt} &= v_2(t) & \mathbf{x_1(t)} = \mathbf{p_1(t)} - \mathbf{p_2(t)} - \delta & \frac{dp_1}{dt} = v_1(t) \\ \frac{dv_2}{dt} &= a_2(t) & \frac{\mathbf{x_1(t)} + \delta}{dt} & \frac{dv_1}{dt} = a_1(t) \\ & \frac{d(p_1 - p_2)}{dt} = v_1(t) - v_2(t) & \text{call this } \mathbf{x_2(t)} \\ & \frac{d(v_1 - v_2)}{dt} = a_1(t) - a_2(t) & \text{call this u(t), the input} \end{split}$$

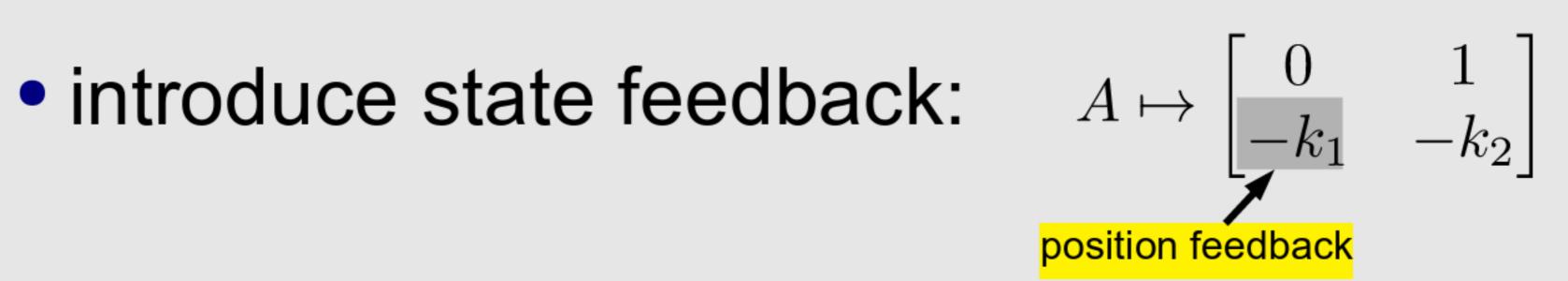
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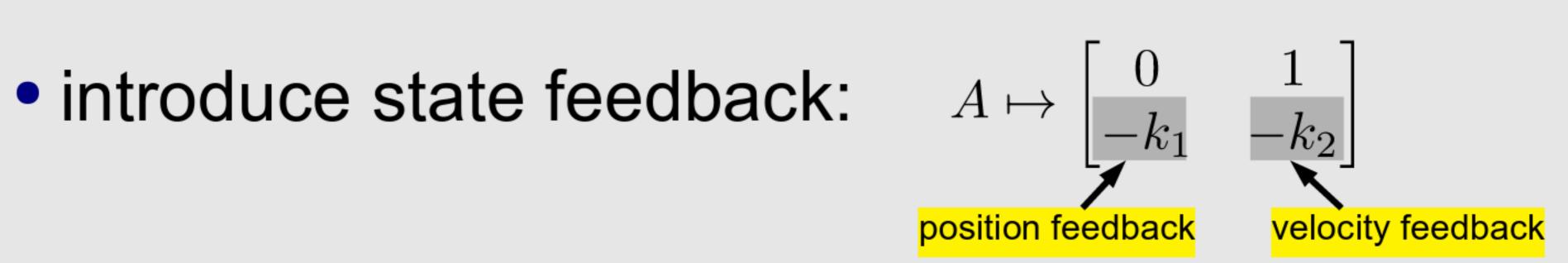
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 BIBO UNSTABLE (a)



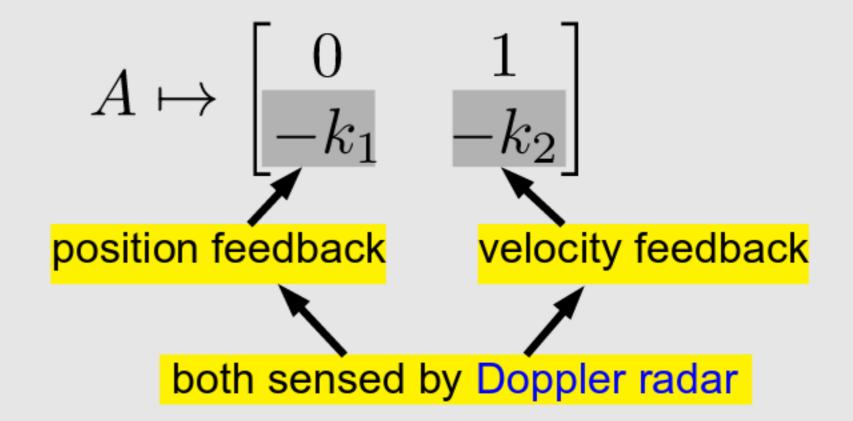
• introduce state feedback: $A \mapsto \begin{bmatrix} 0 & 1 \\ -k_1 & -k_2 \end{bmatrix}$

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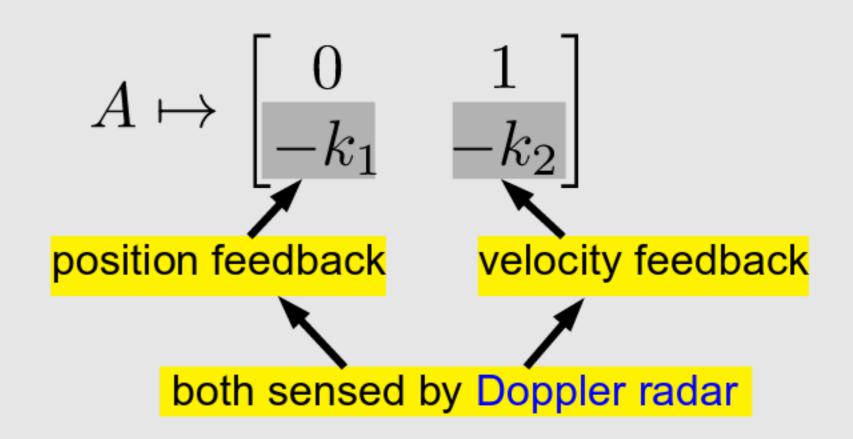




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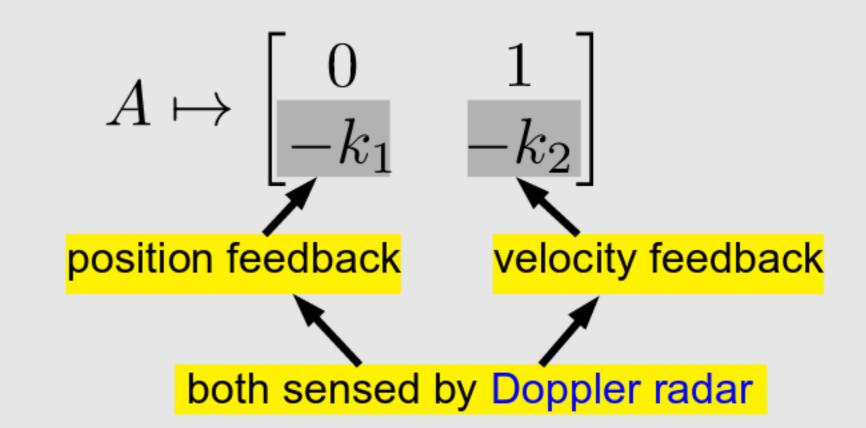


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 - $\lambda_{1,2} = -\frac{k_2}{2} \pm \frac{1}{2} \sqrt{k_2^2 4k_1}$



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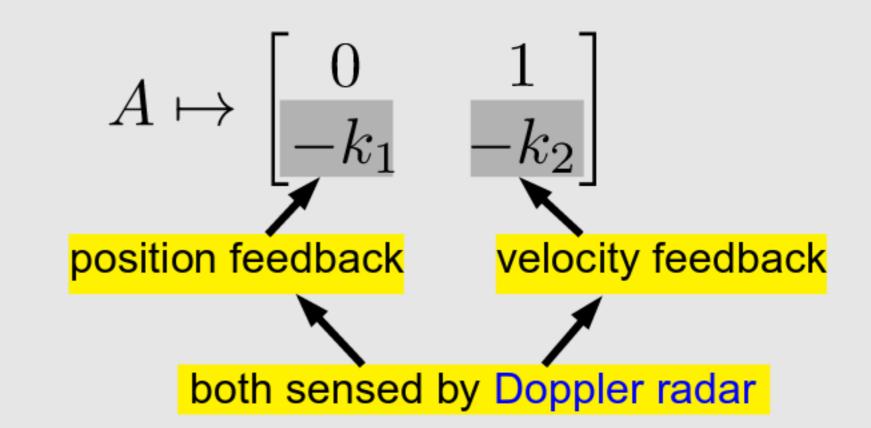
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- stabilization
 - $k_2 > 0$, $k_1 > 0$ ensures eigenvalues have -ve real parts

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- stabilization
 - $k_2 > 0$, $k_1 > 0$ ensures eigenvalues have -ve real parts
- small errors in the acceleration u(t) → only small changes to the desired distance δ
 - see handwritten notes for details

Controllable Systems can be Stabilized

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 - CCF systems can be stabilized by feedback
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 - will be a linear expression in k₁, k₂, ..., k_n

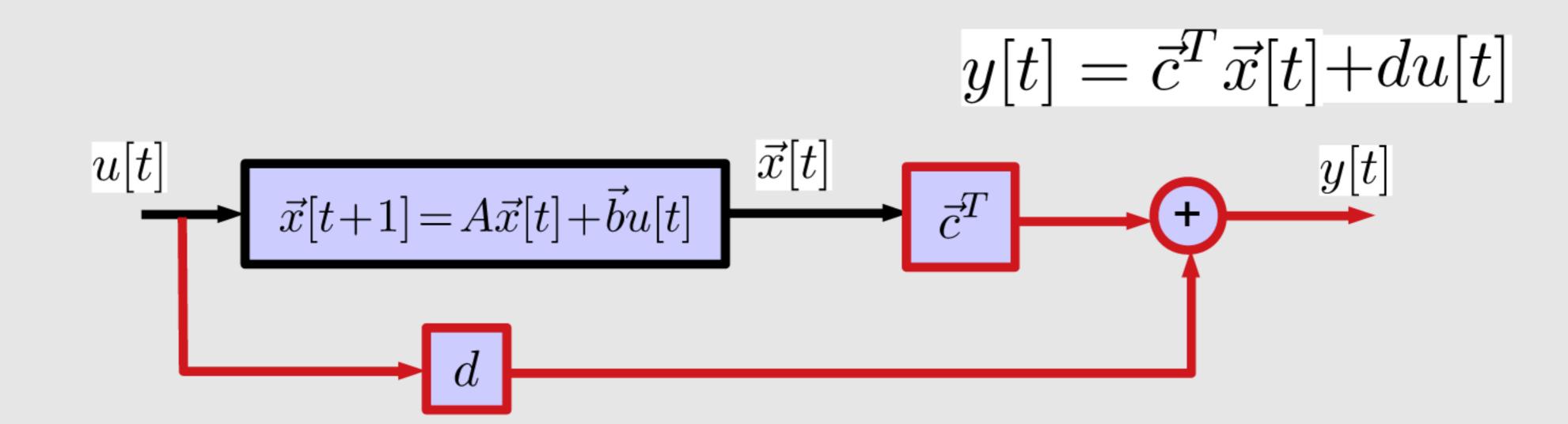
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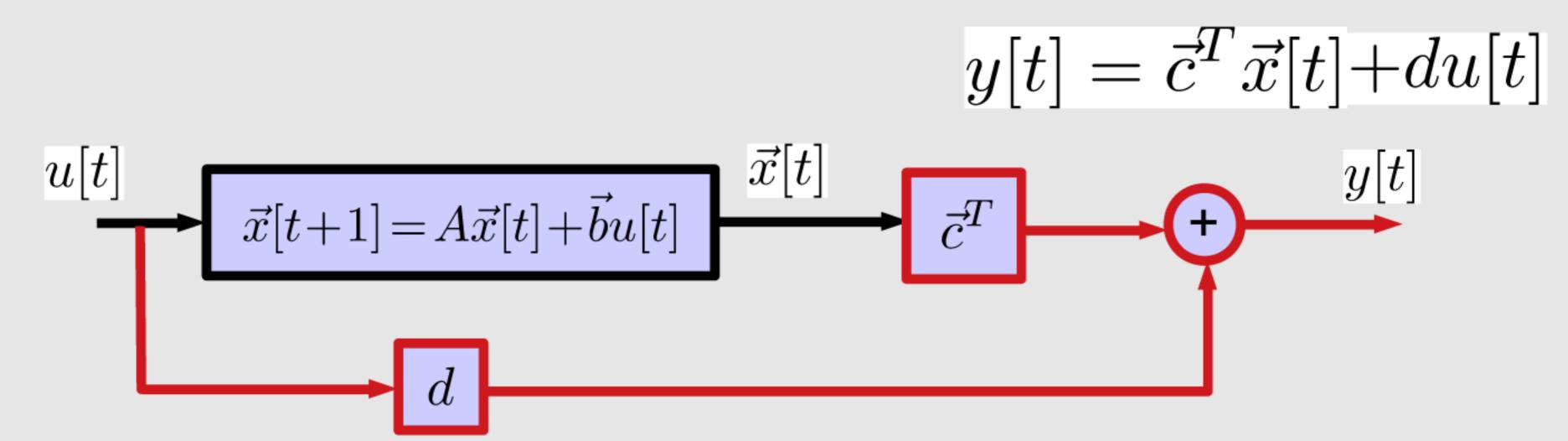
determined by the entries of A, b, and by $\lambda_1, \ldots, \lambda_n$

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 - ⇒ solve $M\vec{k} = \vec{r}$ for \vec{k} (usually numerically) determined by the entries of A, b, and by $\lambda_1, ..., \lambda_n$

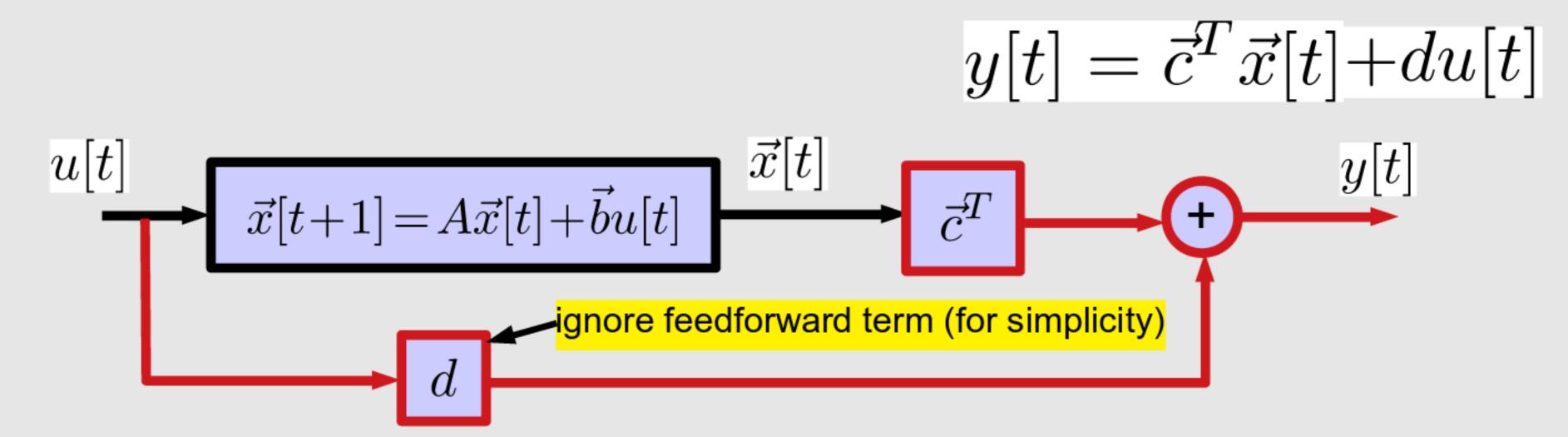
- suppose we have just a SCALAR output y[t]
 - i.e., don't have access to all of $\vec{x}[t]$ for feedback



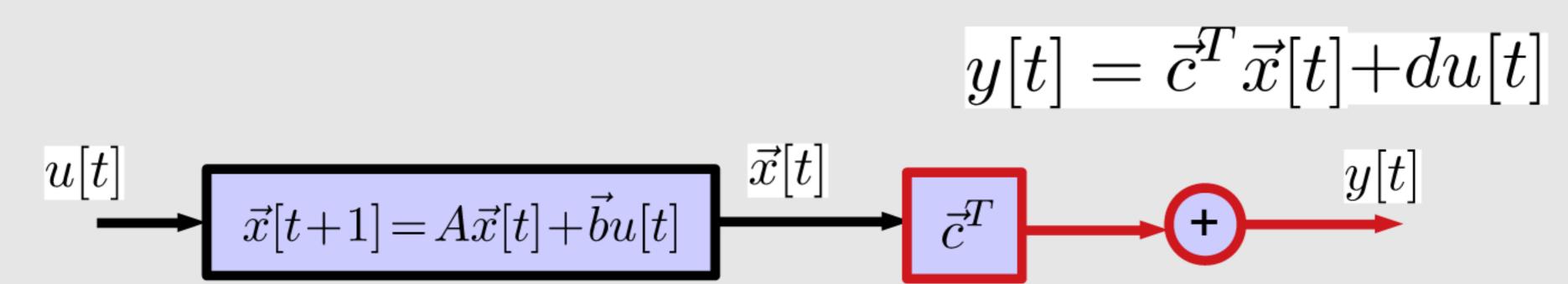
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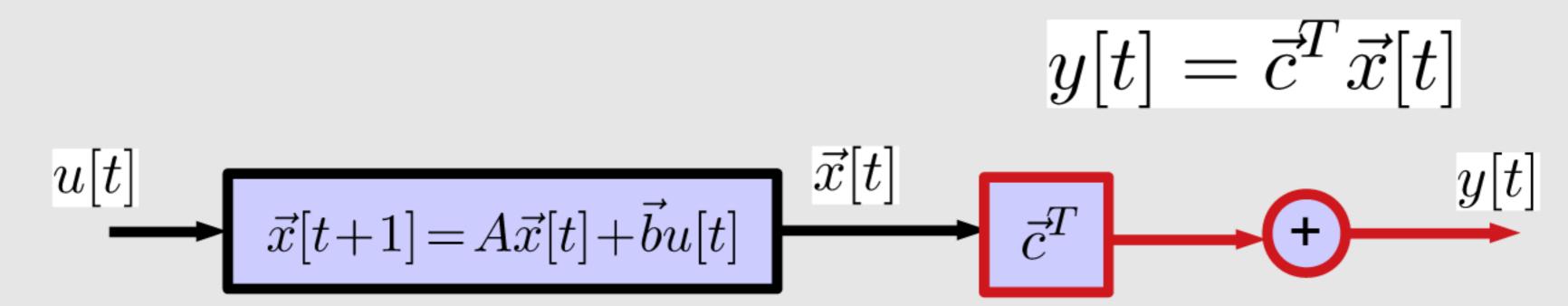
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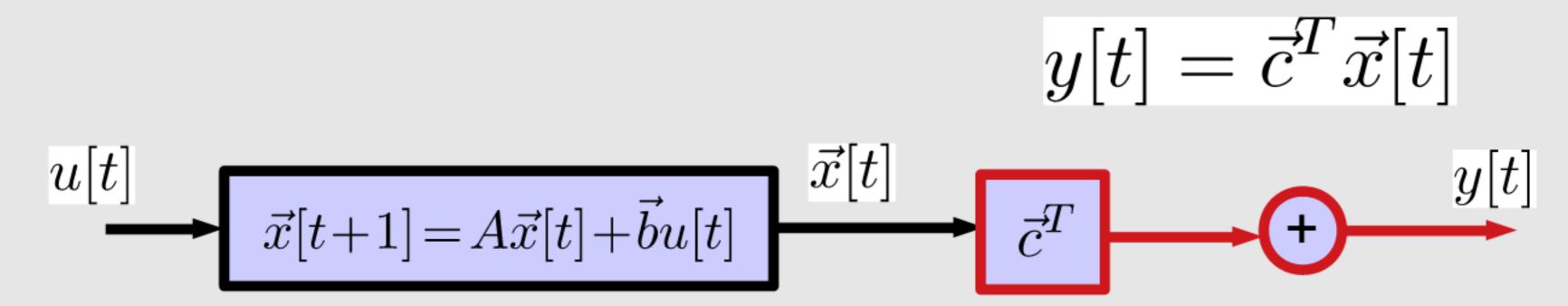
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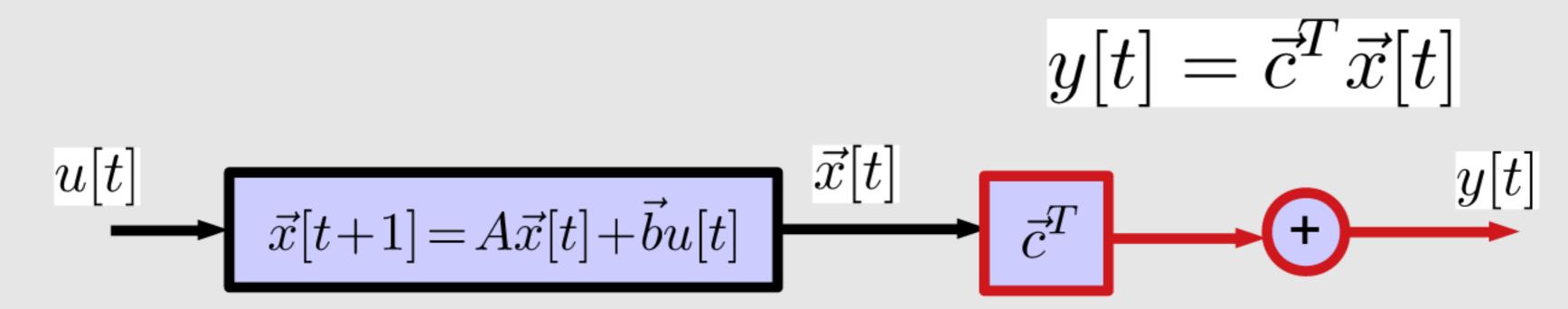


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- More precisely:
 - suppose we know: A, \vec{b} , \vec{c}^T and u[t]
 - → and can measure y(t)
 - can we recover $\vec{x}[t]$?

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__If yes: the system is called OBSERVABLE

• We know that
$$\vec{x}[t] = A^{t-1} \vec{x}[0] + \sum_{i=1}^{s} A^{t-i} \vec{b}u[i-1]$$

we know (or can calculate) these

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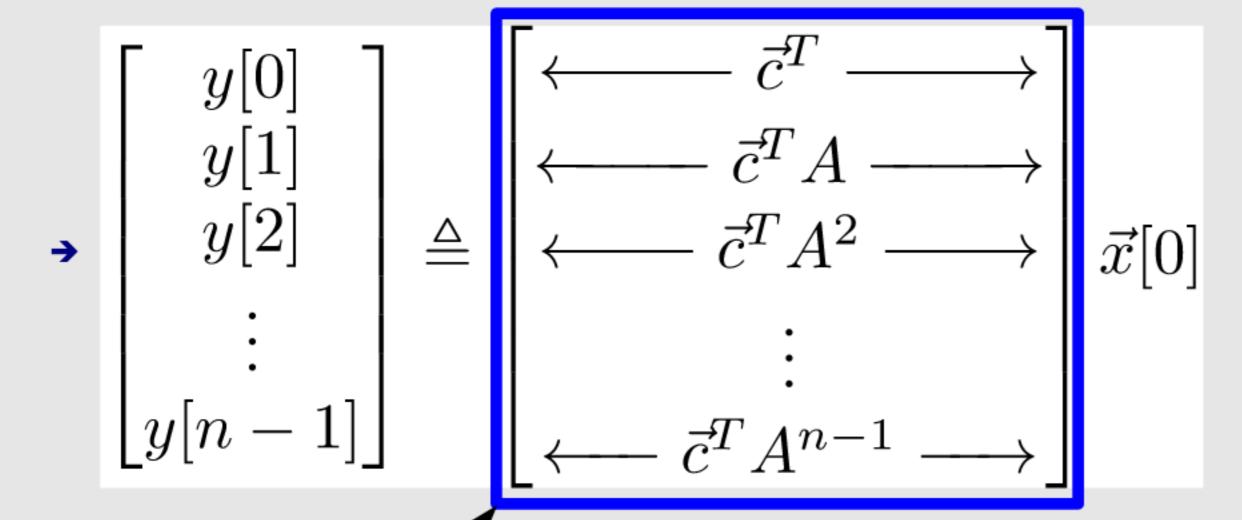
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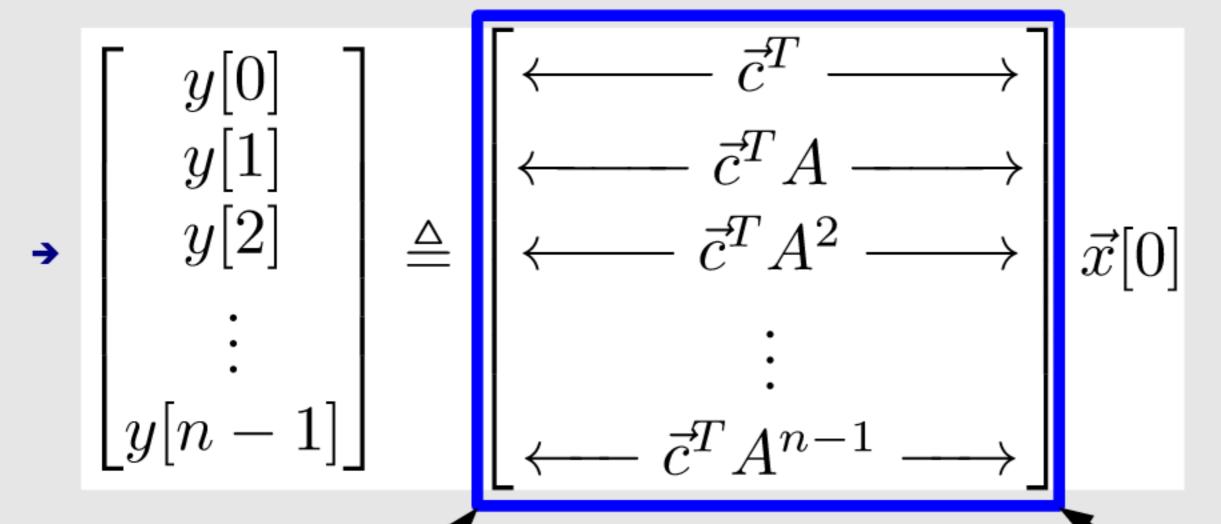
observability matrix (nxn)

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observability matrix (nxn)

must be full-rank/non-singular/invertible to recover $\vec{x}(t)$ uniquely from measurements of y(t)

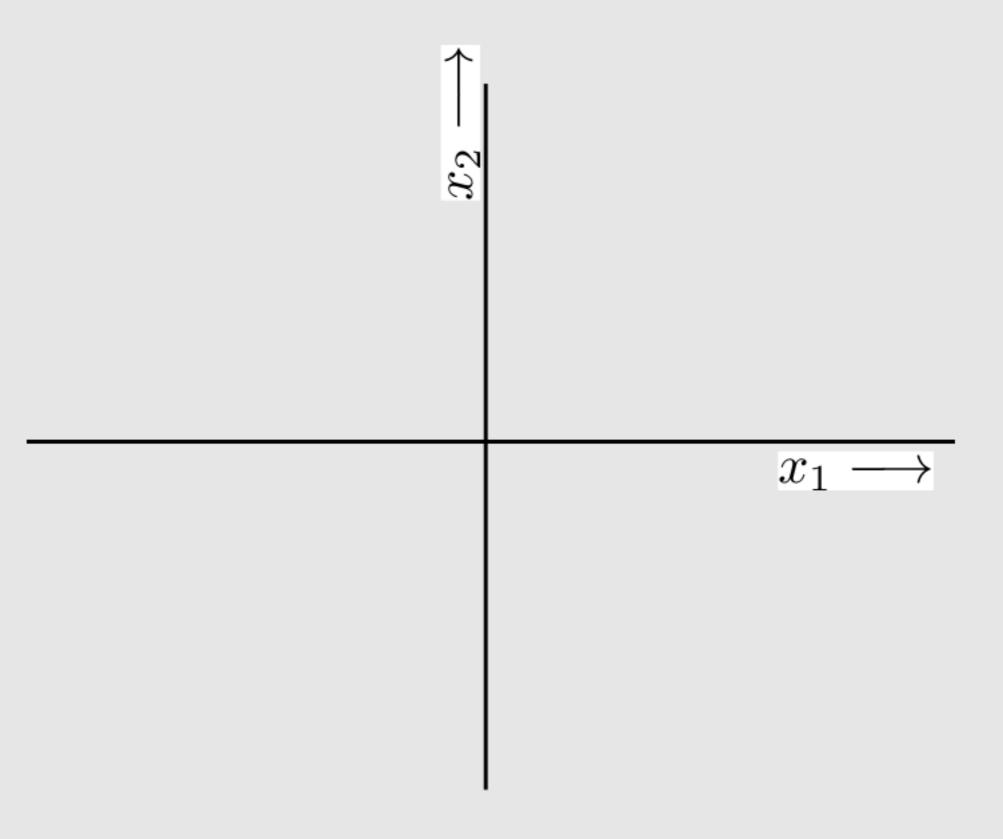
 $\begin{bmatrix} x_1[t+1] \\ x_2[t+1] \end{bmatrix} = \begin{bmatrix} \cos(\theta) & -\sin(\theta) \\ \sin(\theta) & \cos(\theta) \end{bmatrix} \begin{bmatrix} x_1[t] \\ x_2[t] \end{bmatrix}$

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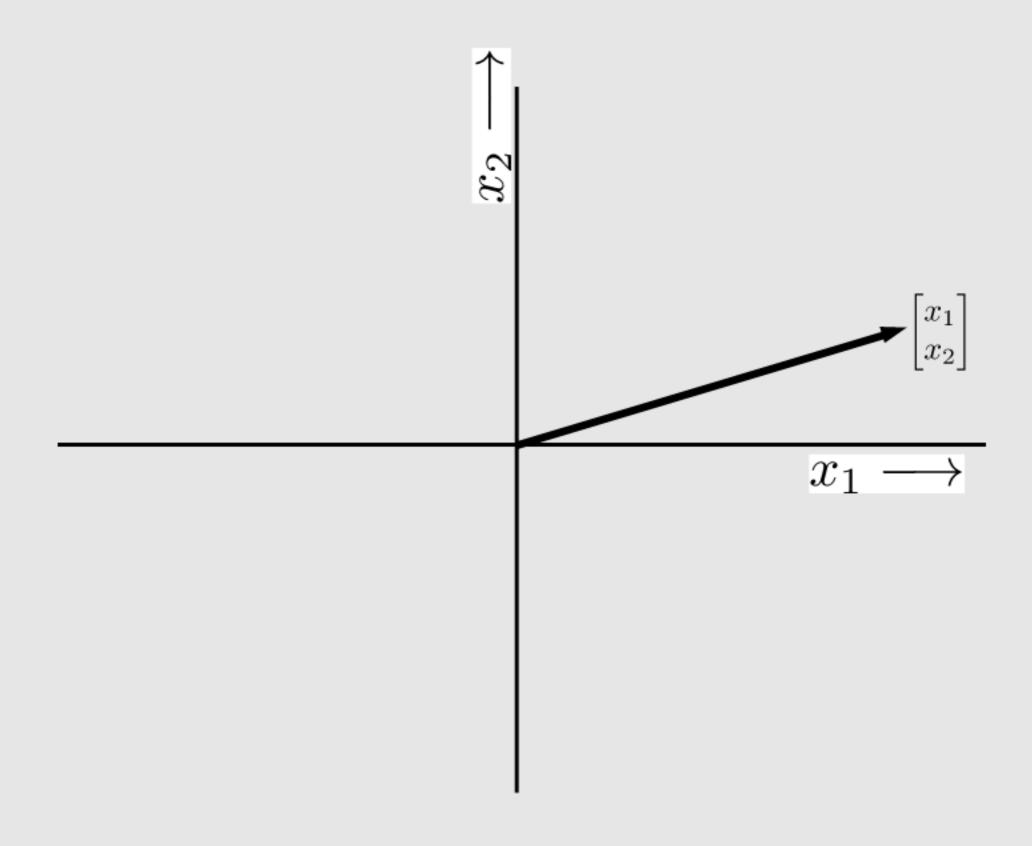
this is a "rotation matrix" - call it A

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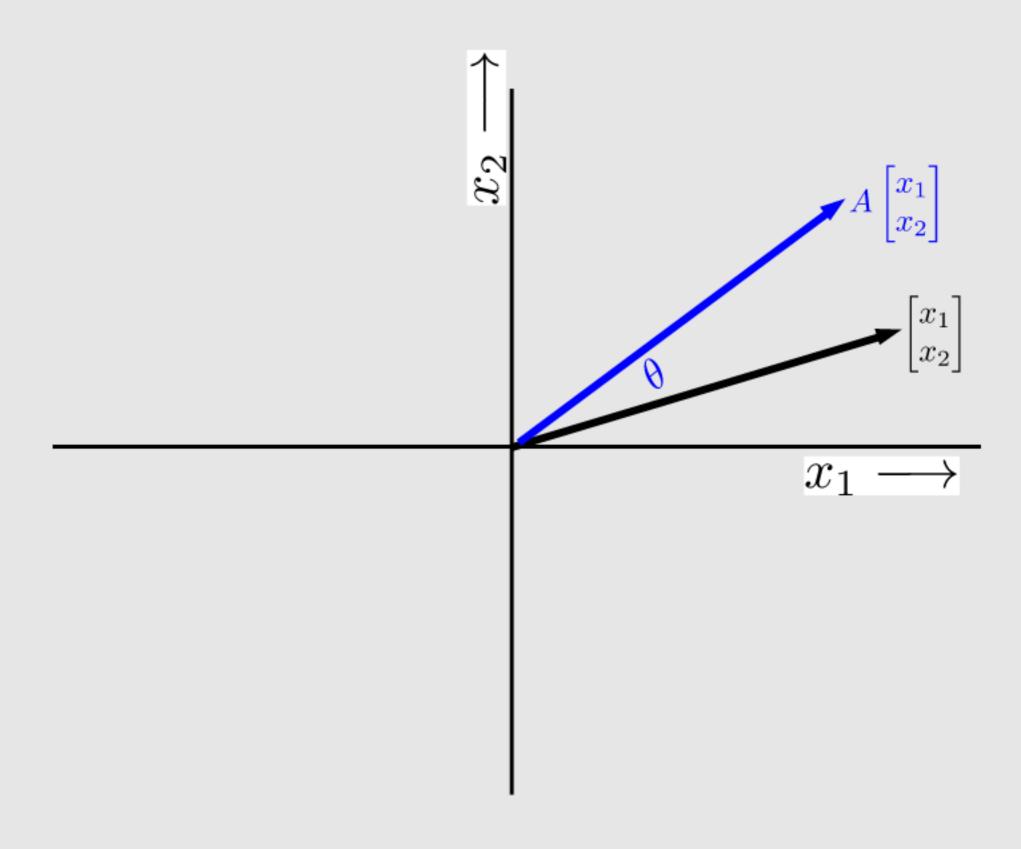
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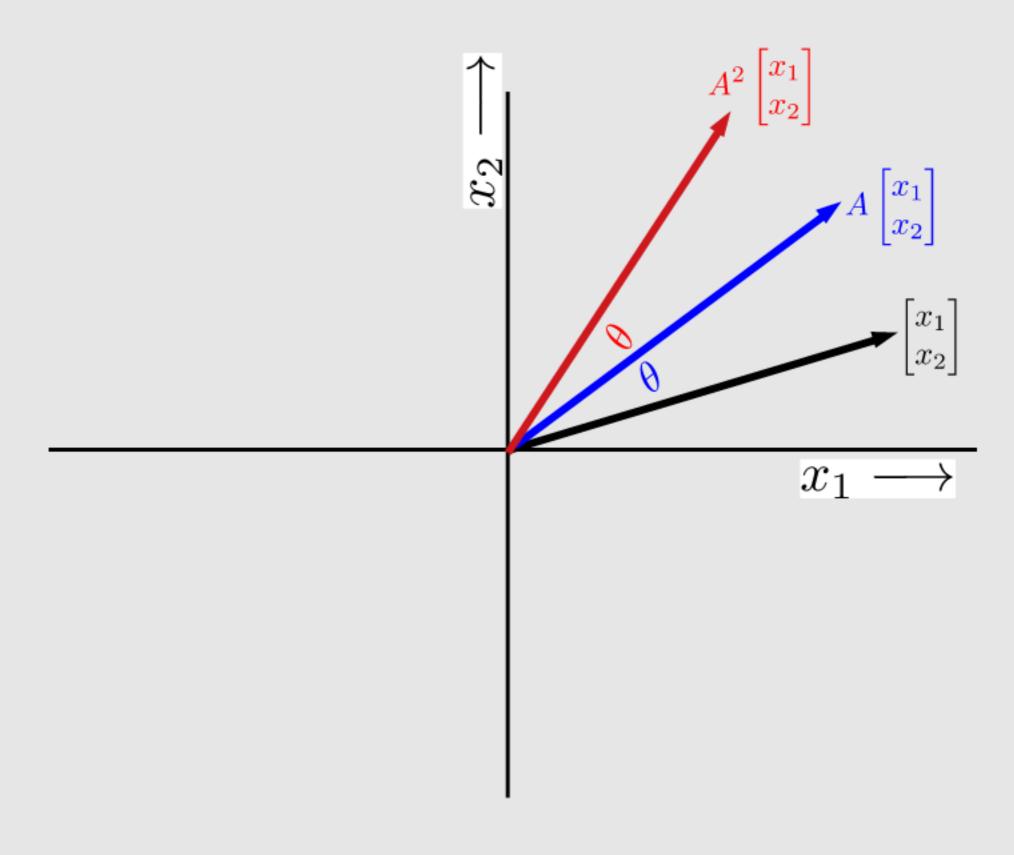
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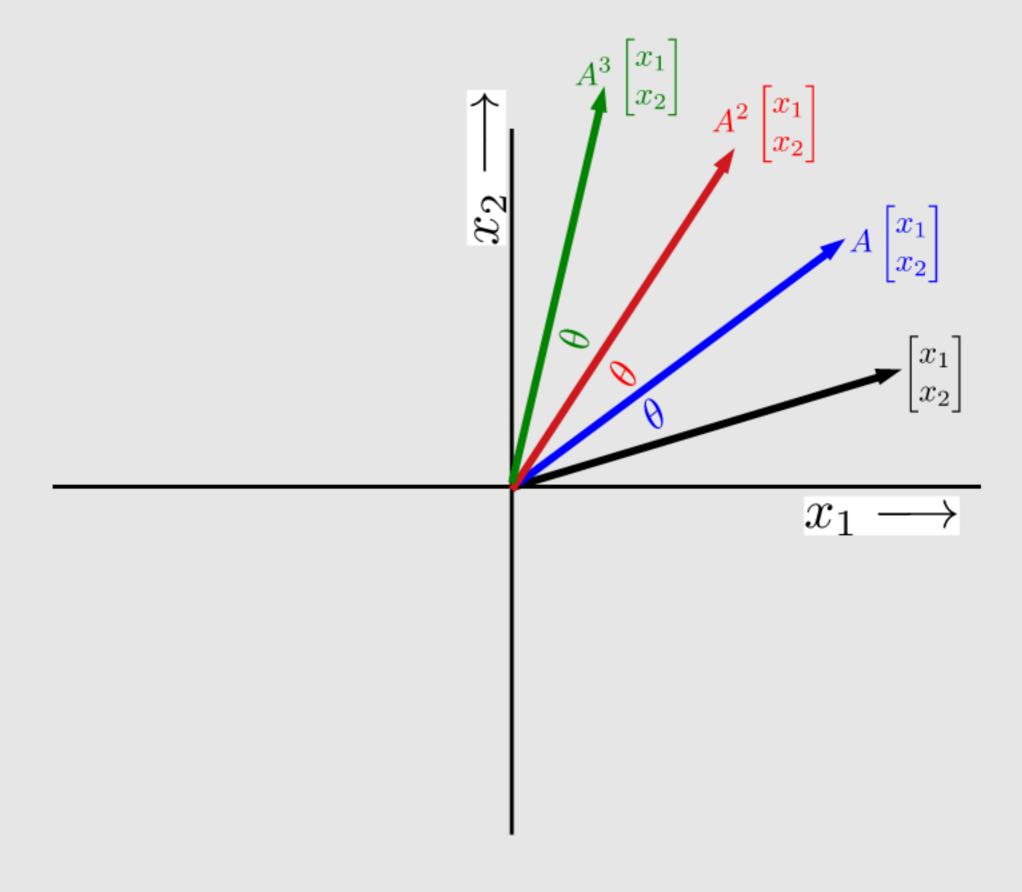
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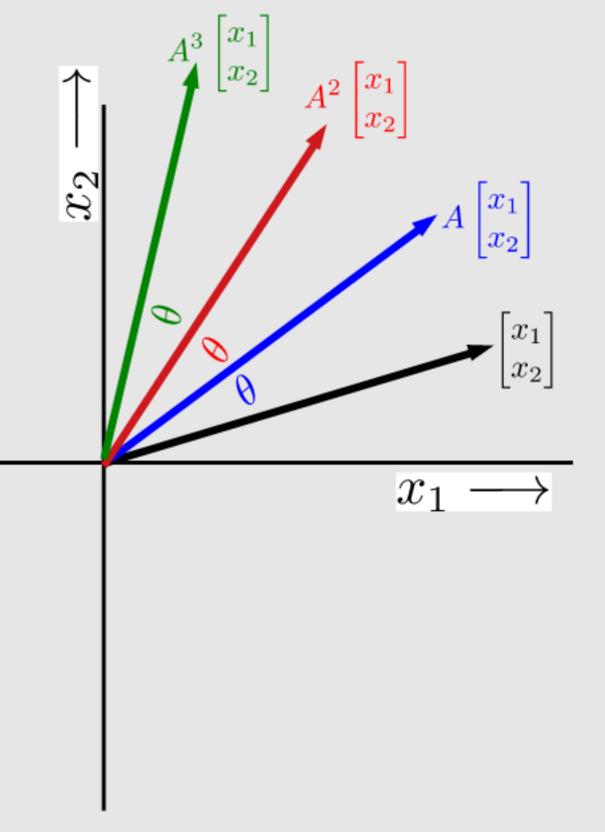
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Each application of A rotates

by θ



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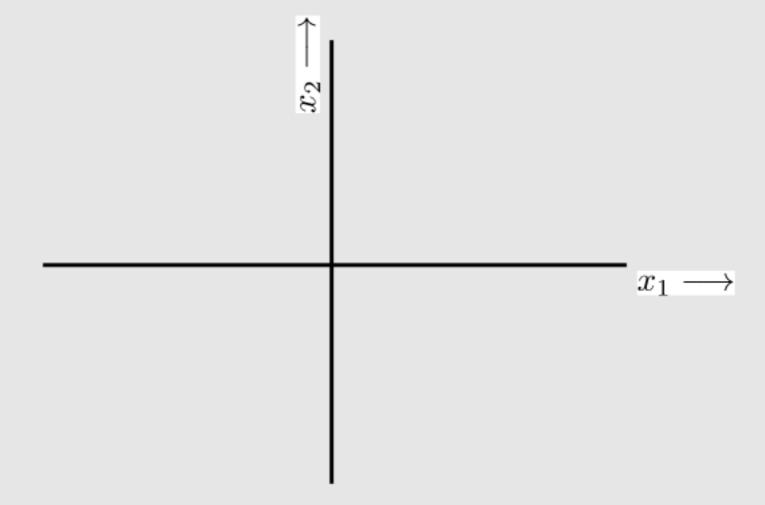
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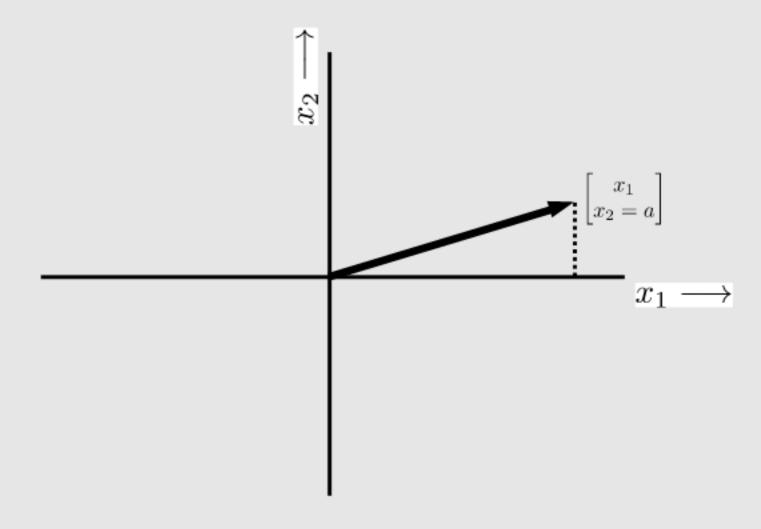
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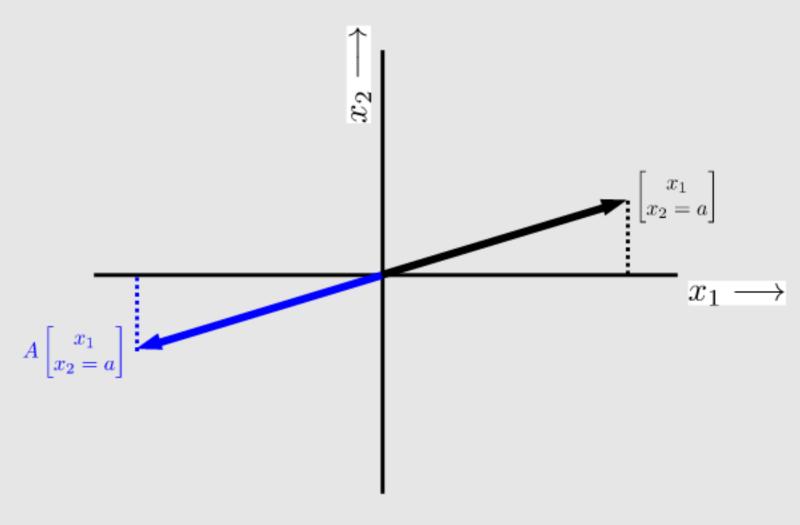
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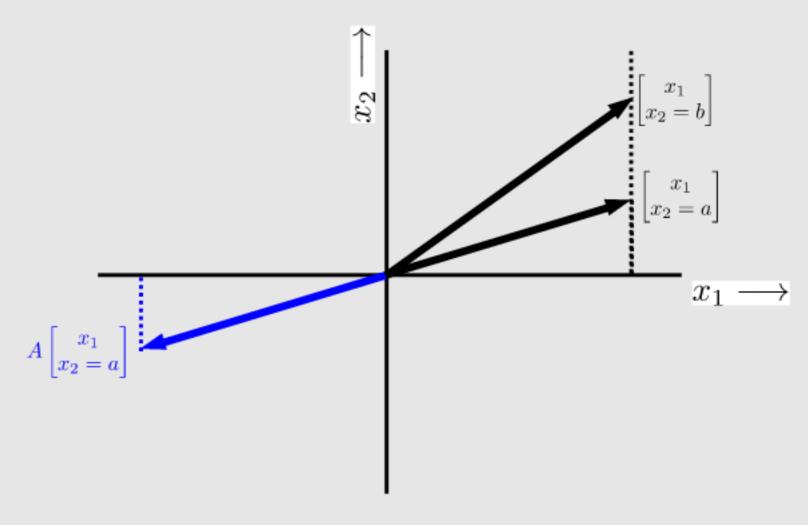
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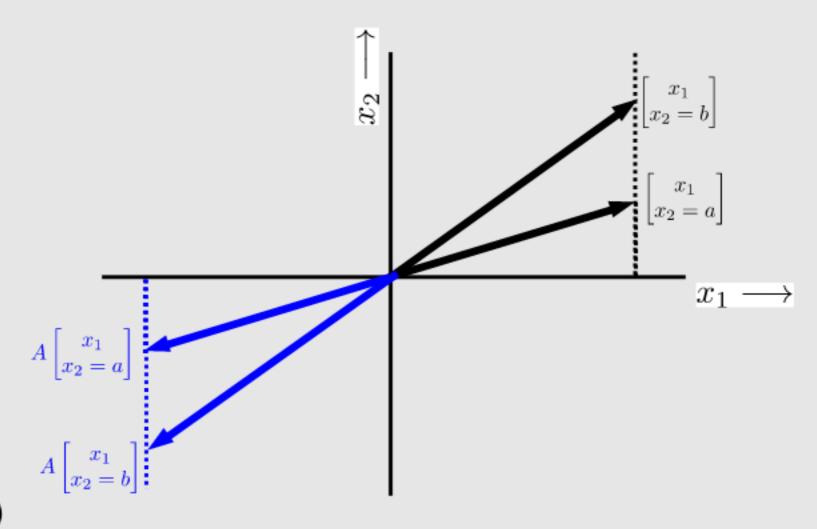
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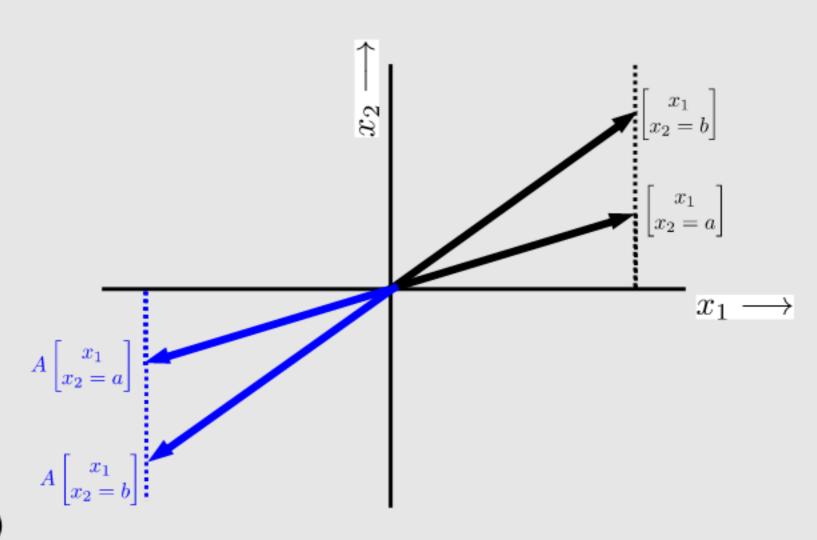
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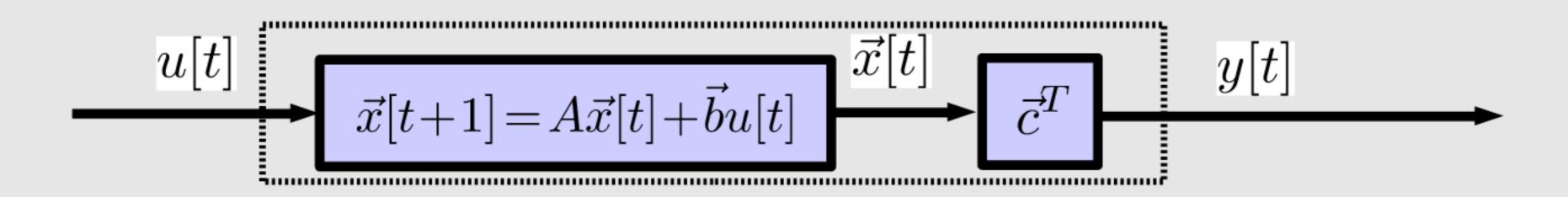


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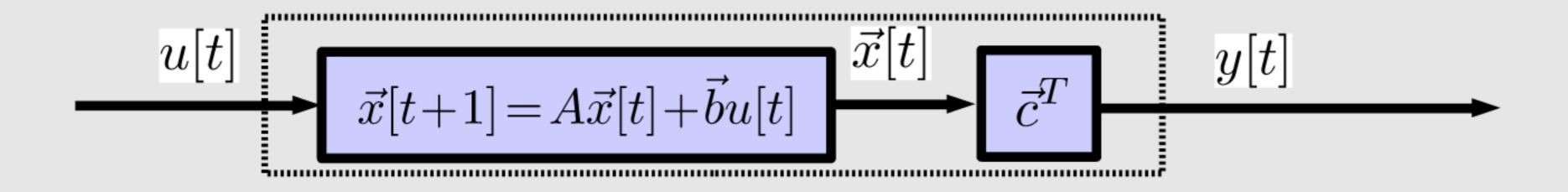
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 - cannot recover x₂ uniquely



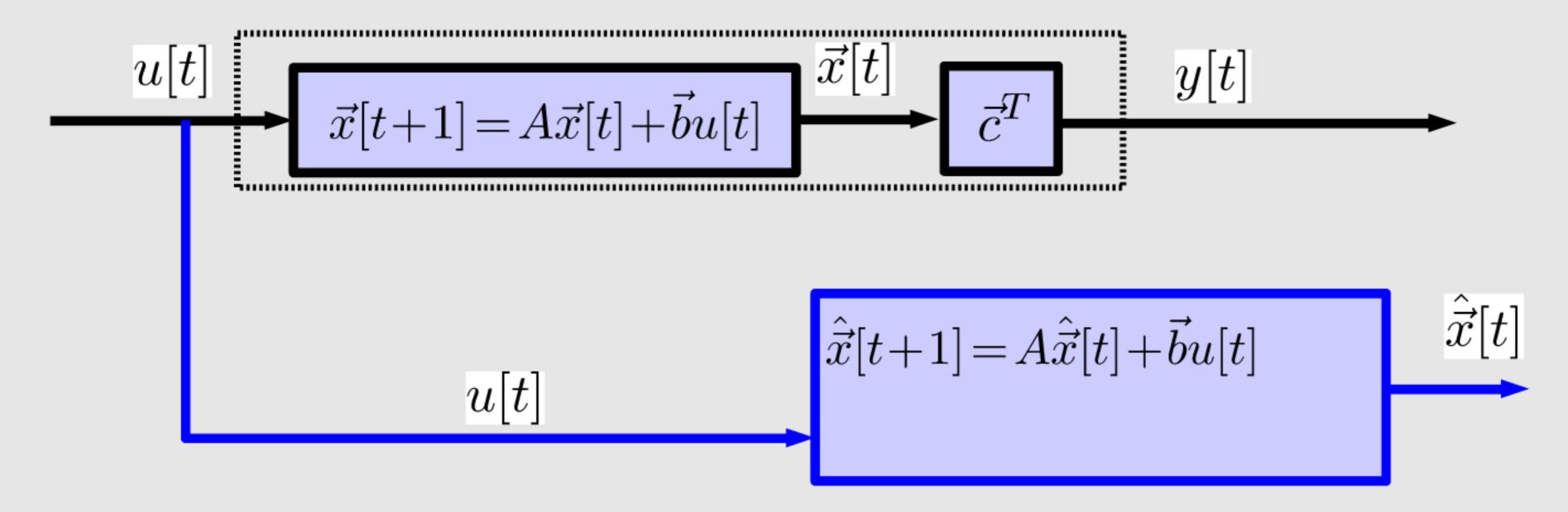
- Can we make a system that recovers $\vec{x}[t]$ from y[t] in real time?
 - (we can use our knowledge of A, \vec{b} , u[t] and y[t])



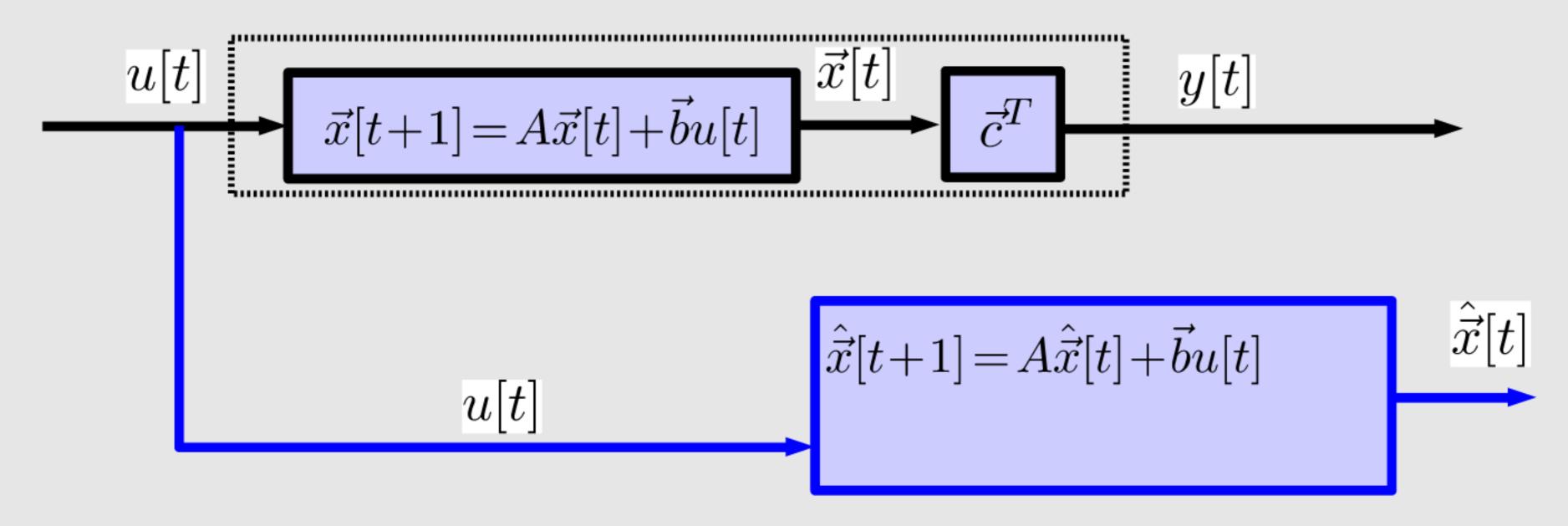
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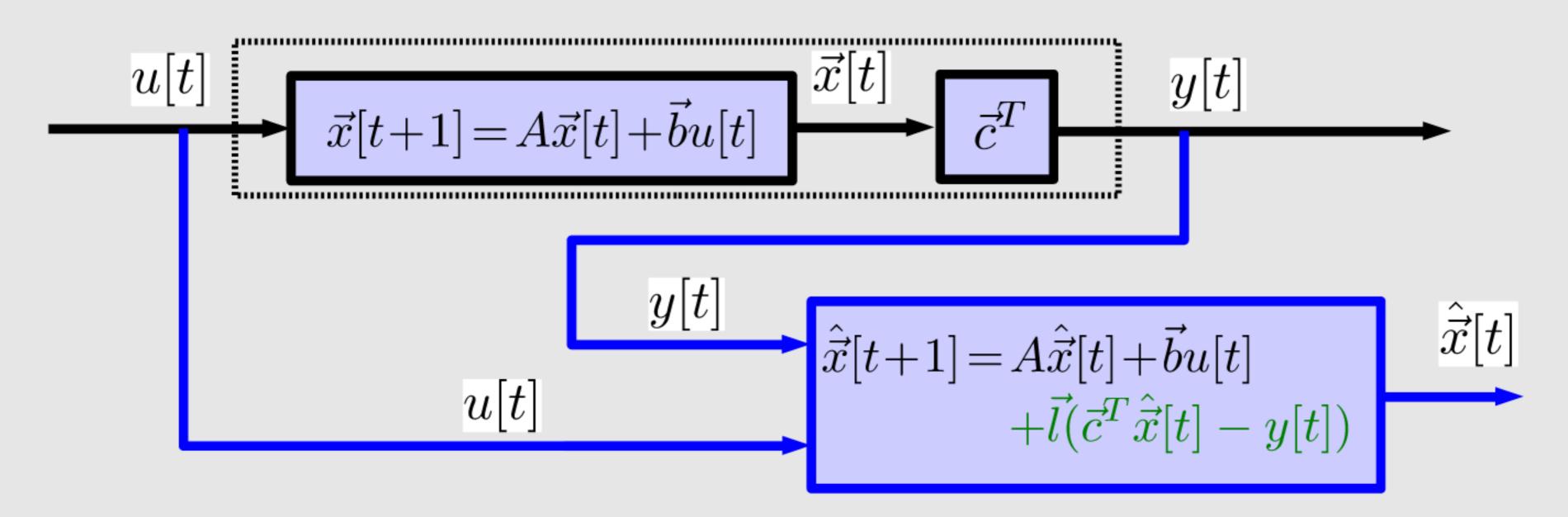
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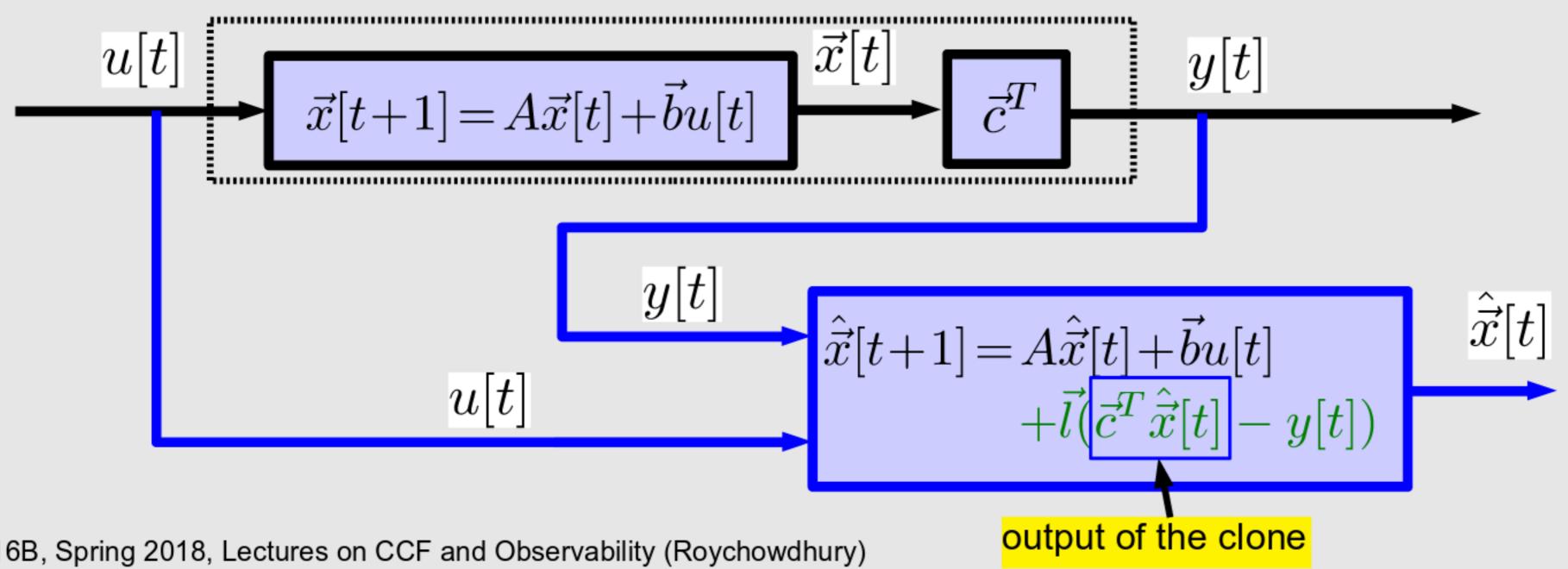
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 - next: incorporate the difference between the outputs of the actual system and the clone



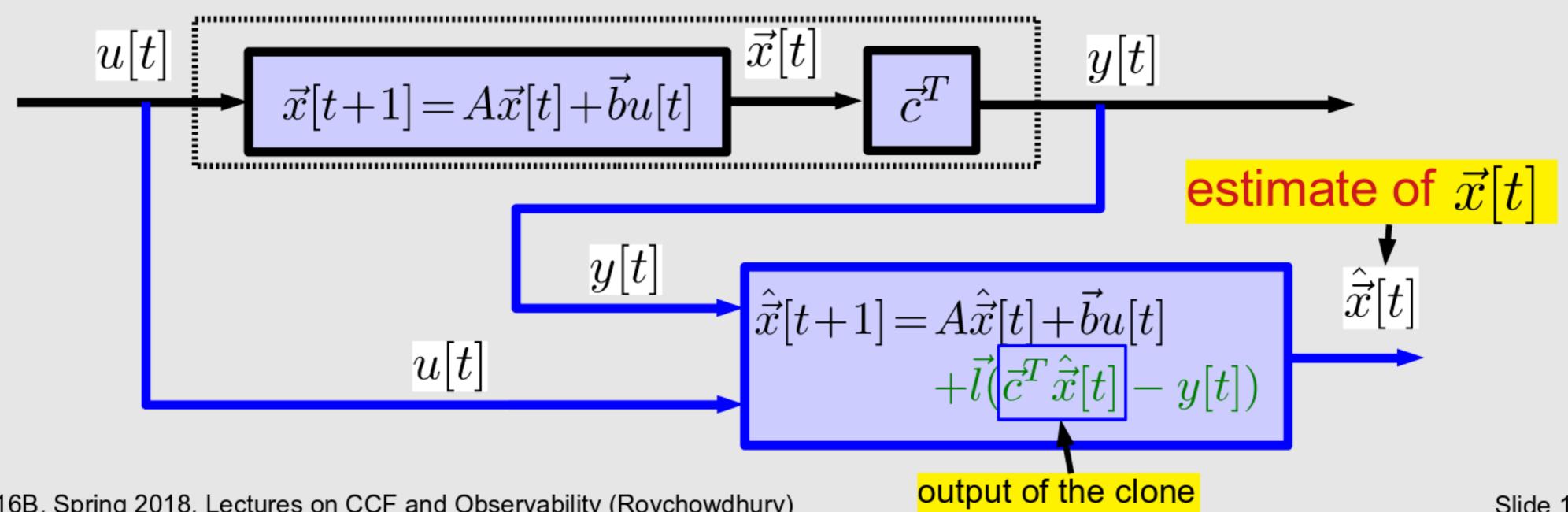
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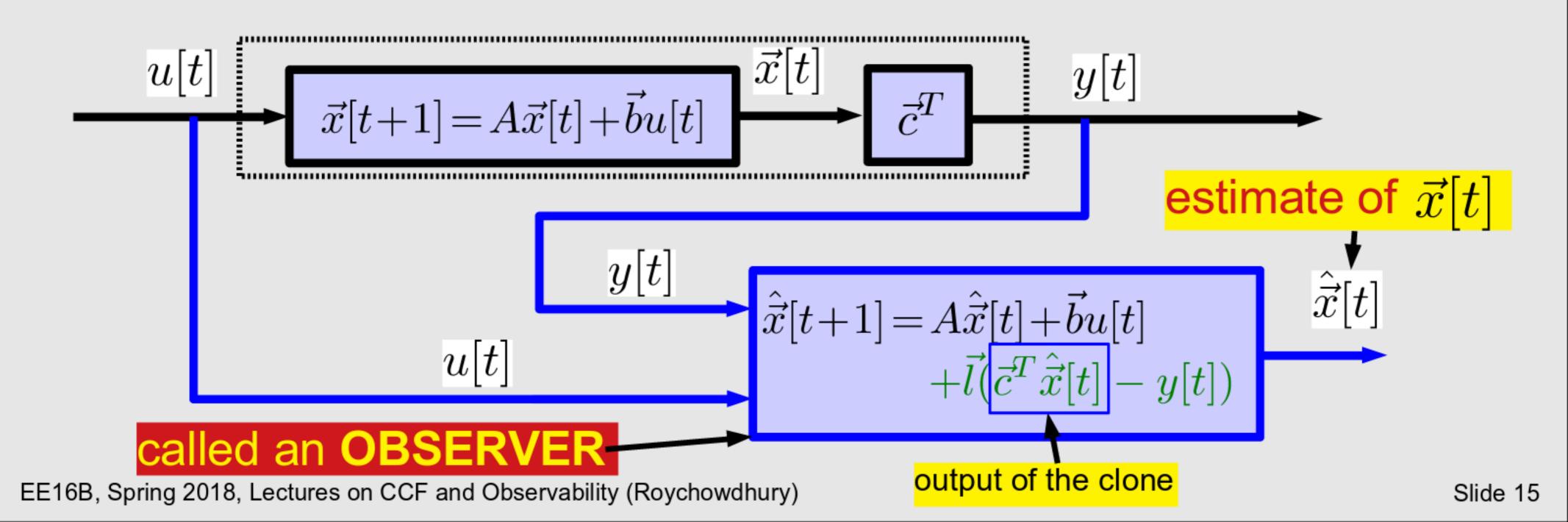
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- i.e., can always make $A + \vec{l} \vec{c}^T$ stable if $(A^T, -\vec{c})$ is controllable (using previous controllability + feedback result)

• $(A^T, -\vec{c})$ controllable $\rightarrow -\left[\vec{c} | A^T\vec{c} | \cdots | (A^T)^{n-2}\vec{c} | (A^T)^{n-1}\vec{c}\right]$ must be full rank

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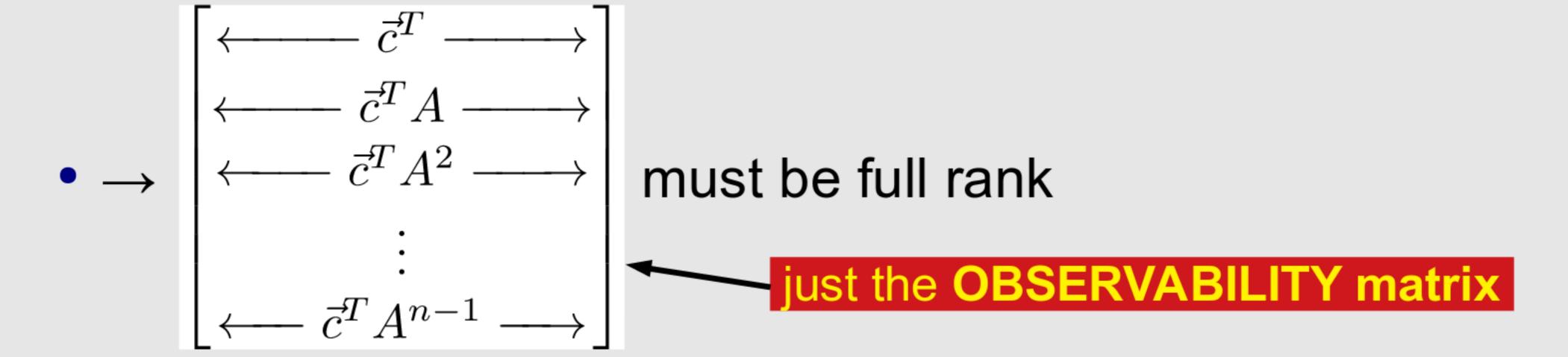
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• Conclusion: if a system is observable, we can build an observer for it whose estimate $\hat{\vec{x}}[t]$ will approximate $\hat{\vec{x}}[t]$ more and more closely with t

• example:
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Observer: Rotation Matrix Example

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 - warning: if complex, ensure evs are complex conjugates
 - > what will happen if you don't?

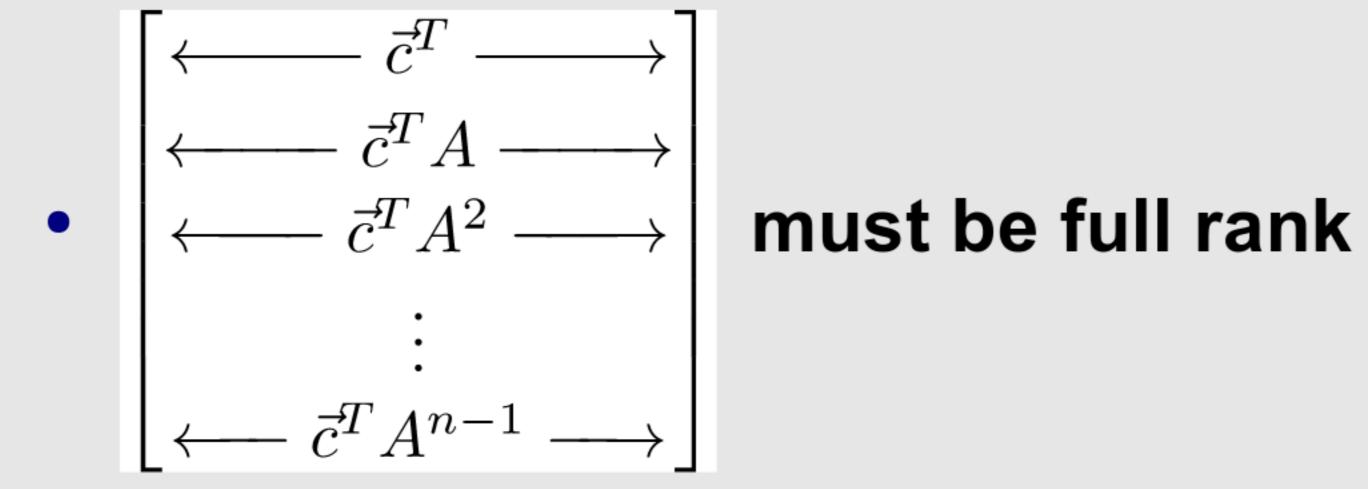
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Observability: The Continuous Case

- Observability for C.T. state-space systems
 - and implications for placing observer eigenvalues
- EXACTLY THE SAME CRITERIA



Stability for C.T. means Re(eigenvalues) < 0

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 - due to the relationship between position, velocity and acceleration
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 - NOT A VERY PRACTICALLY USEFUL WAY TO LOCATE YOURSELF

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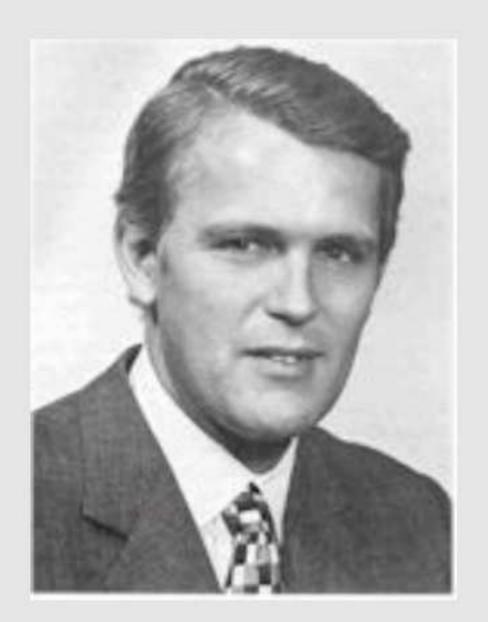
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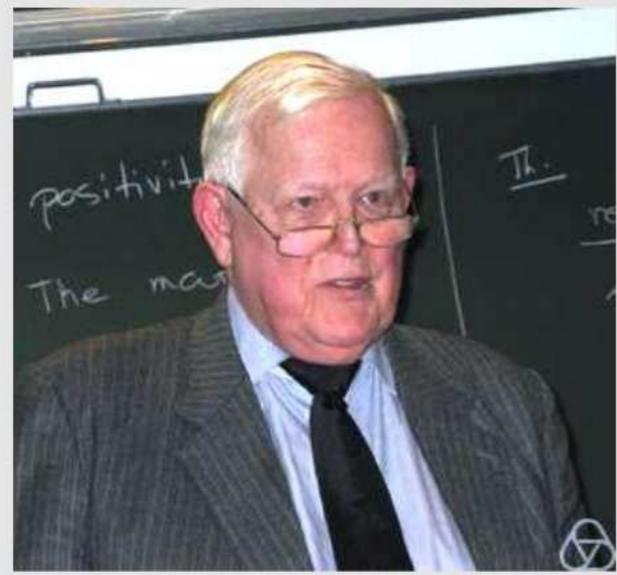
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 - this is the famous KALMAN FILTER
 - used in all rockets, drones, autonomous cars, ships, ...

Rudolf Kálmán "inventor" of control theory: 1950s/60s

- state-space representations
- stability, controllability, observability and implications
- Kalman filter



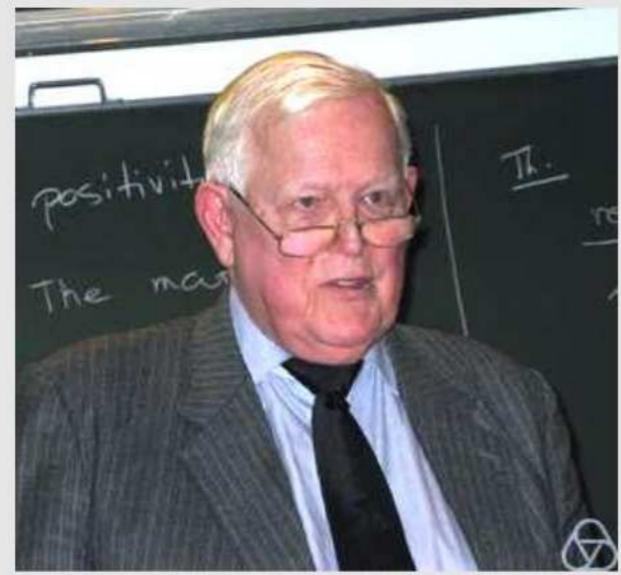




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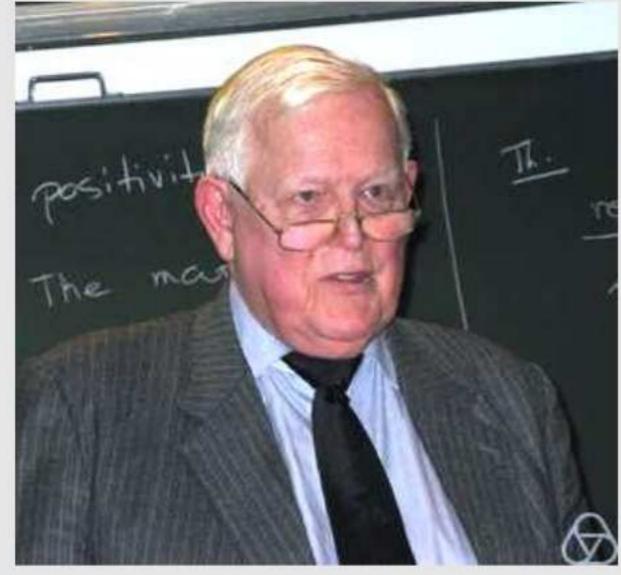




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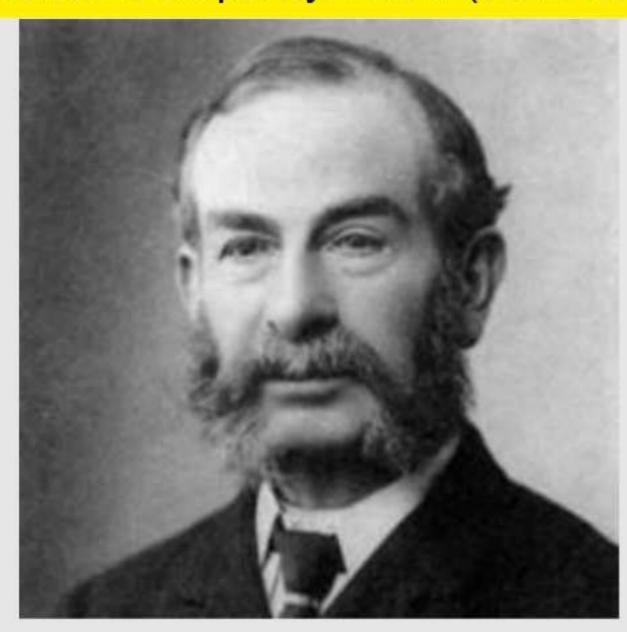




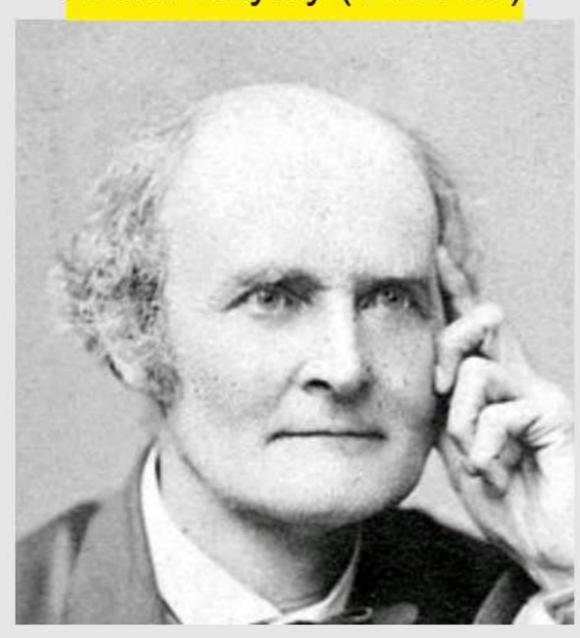
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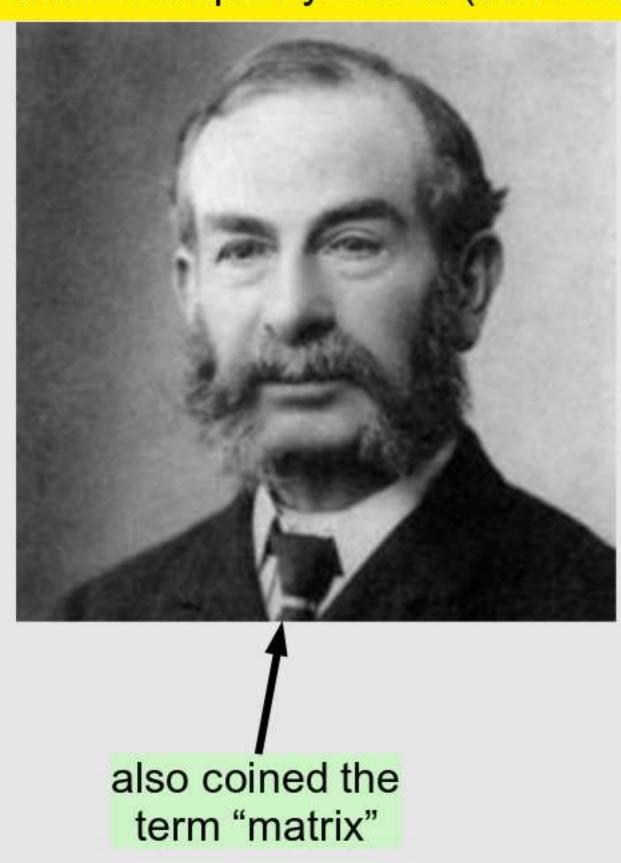
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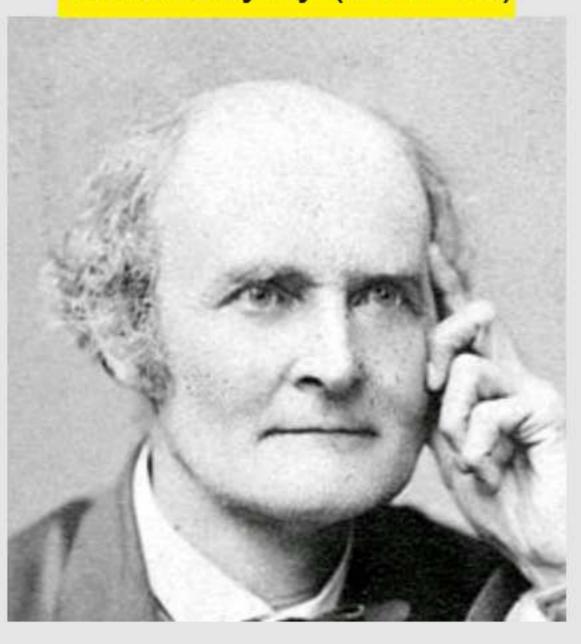
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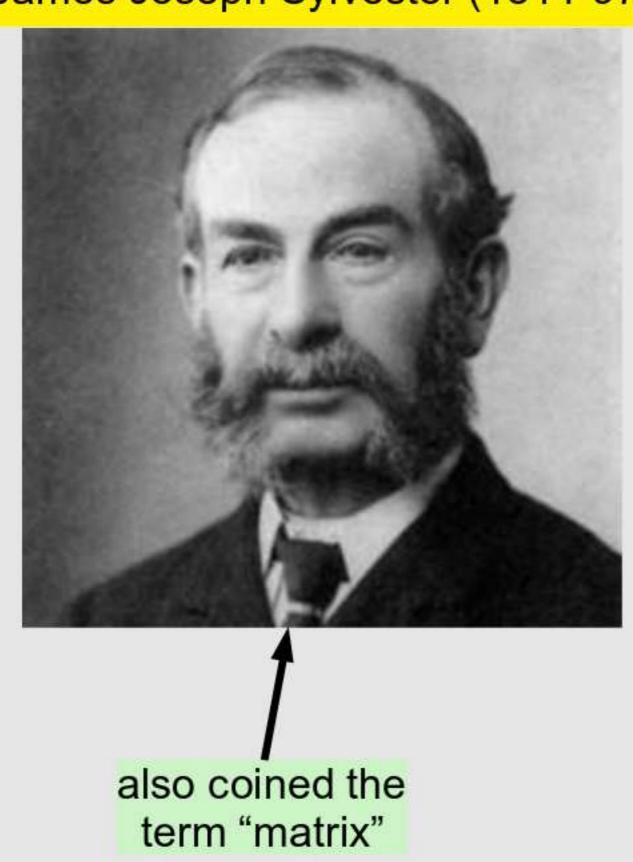
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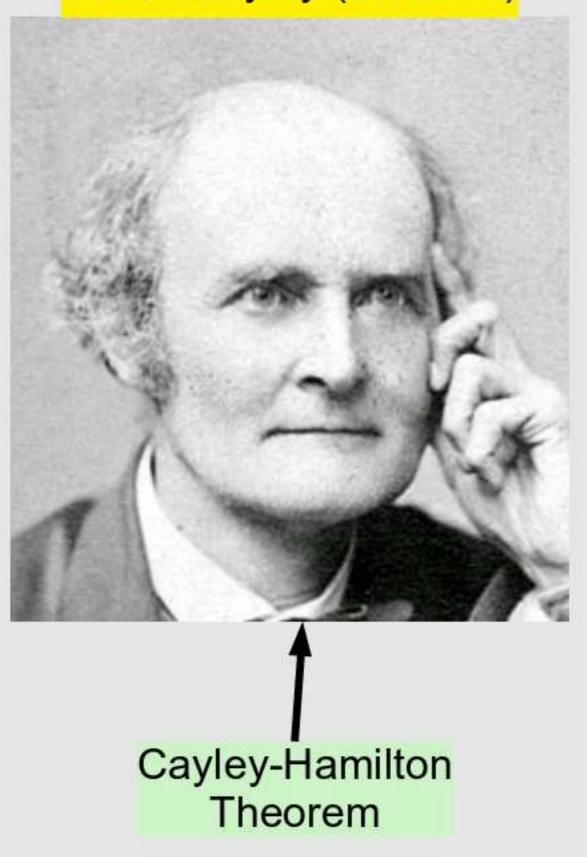
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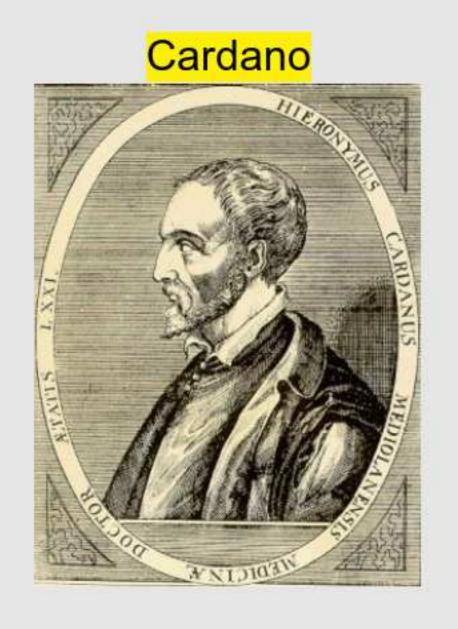


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- known and used in <u>China</u> before 100BC (!)
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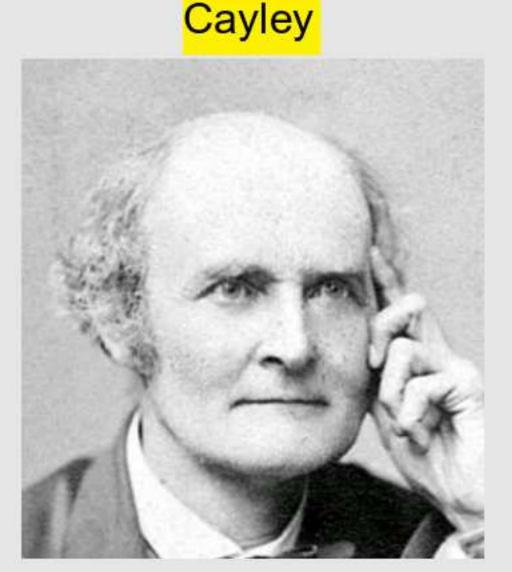
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- 1683: Seki ("Japan's Newton") used matrices
- developed in Europe by Gauss and many others
 - finally, into its modern form by Cayley (mid 1800s)









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