

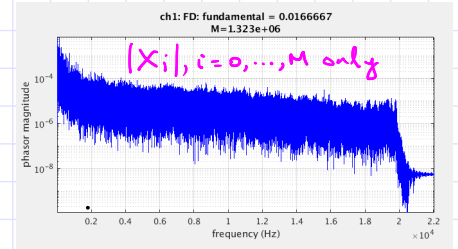
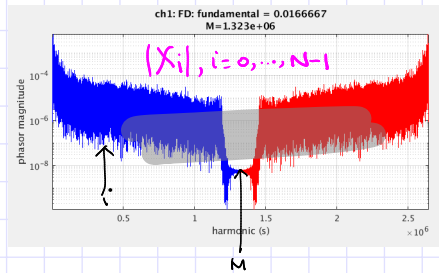
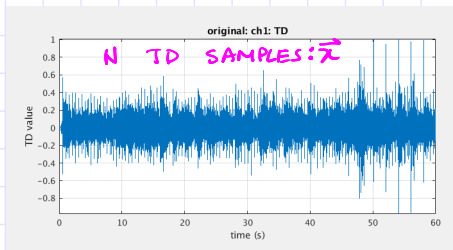
## THE DISCRETE FOURIER TRANSFORM (DFT) AND ITS USES

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## 8. FILTERING VIA THE DFT

→ RECALL: (60+ε)s OF CALLING-ELVIS:  $T \approx 60s$ ,  $f_0 = \frac{1}{T} \approx \frac{1}{60} \text{ Hz}$ ,  $M = 1,323,000$ ,  $N = 2M + 1 = 2,626,001$

$$\vec{X} = F_N \vec{x}$$



→ RECALL: FREQ. DOMAIN INTERPRETATION:

→  $\vec{x}$  are samples of  $x(t) = \underbrace{\frac{X_0}{N}}_{\text{DC}} + \sum_{i=1}^M \underbrace{\left( \frac{2}{N} |X_i| \right)}_{\text{magnitude of } X_i} \cos(2\pi f_0 i t + \underbrace{\angle X_i}_{\text{angle of } X_i = \theta_i})$

$A_i \cos(2\pi f_0 i t + \theta_i)$

$A_i e^{j\theta_i} = \frac{2}{N} X_i$  ( $i=1, \dots, M$ )

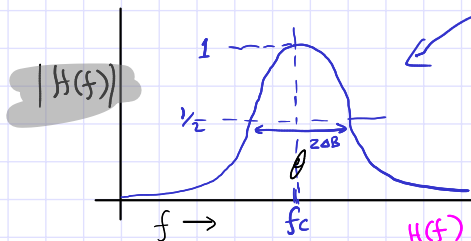
$\sum X_i$ : PHASOR at freq =  $f_0$

$$X_i = |X_i| e^{j\theta_i}$$

→ FILTERING:

→ MODIFY PHASOR AT any freq.  $f$  by  $H(f)$  (MULTIPLYING WITH TRANSFER FUNCTION OF FILTER)

→ EXAMPLE: BAND-PASS FILTER



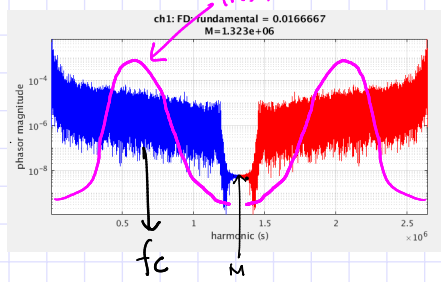
→ Eg:

$$H(f) = \frac{1}{\left(\frac{f-f_c}{\Delta B}\right)^2 + 1}$$

$$f = f_c \pm \Delta B \Rightarrow H(f) = \frac{1}{2}$$

SIMILAR TO  $H(f)$  OF AN RLC CIRCUIT

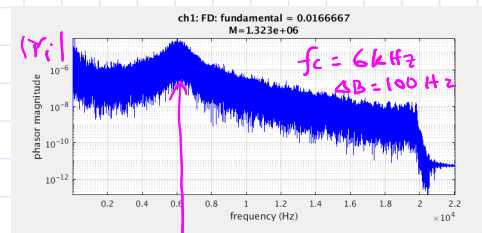
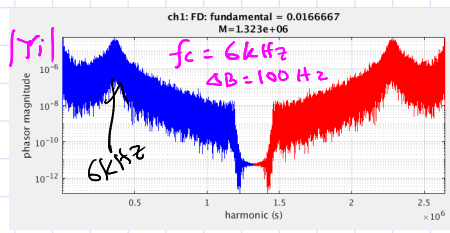
→ APPLY →



$$\rightarrow Y_i = X_i H(i f_0), \quad i=0, \dots, M$$

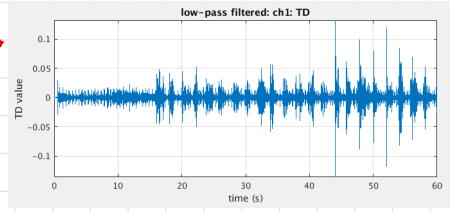
$$\rightarrow Y_{N-i} = \overline{Y_i}, \quad i=0, \dots, M \leftarrow \text{COMPLEX CONJUGATES}$$

→ FILTERED: FD.



DEMO: PLAY IN MATLAB →

TD



## 9. CONVOLUTION AND THE DFT

9A: → RECALL CONVOLUTION:

$y[t]$ : CAUSAL  $\Rightarrow y[t] = 0 \quad \forall t < 0$

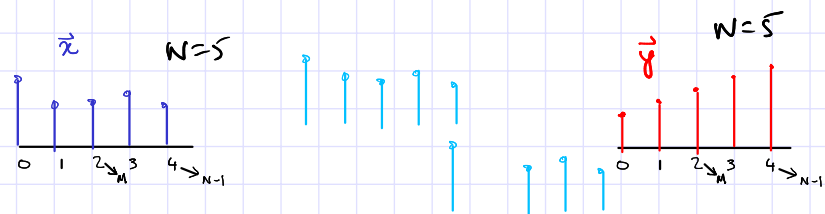
$$\rightarrow z[t] \triangleq u[t] \otimes h[t] \triangleq \sum_{i=-\infty}^{\infty} u[i] h[t-i]$$

$$= \sum_{i=-\infty}^t u[i] h[t-i] \quad (\text{IF } h[t] \text{ CAUSAL})$$

$$= \sum_{i=0}^t u[i] h[t-i] \quad (\text{IF } h[t] \text{ and } u[t] \text{ BOTH CAUSAL})$$

9B: CIRCULAR CONVOLUTION

→ GIVEN  $\vec{x}$  &  $\vec{y}$ , both  $\in \mathbb{R}^N$



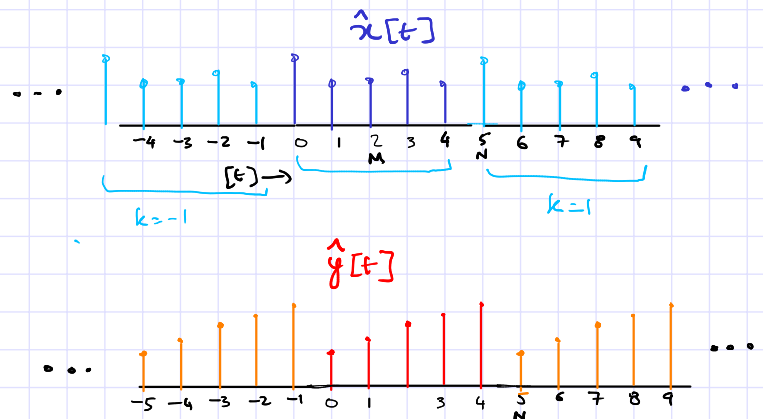
→ EXTEND BOTH PERIODICALLY (CIRCULARLY)

$$\rightarrow \hat{x}[t] = x[t], \quad t = 0, \dots, N-1$$

$$\rightarrow \hat{x}[t] = \hat{x}[t + kN], \quad \forall t, \forall k \in \mathbb{Z}$$

$$\rightarrow \hat{y}[t] \triangleq y[t], \quad t = 0, \dots, N-1$$

$$\rightarrow \hat{y}[t] = \hat{y}[t + kN] \quad \forall t, \forall k \in \mathbb{Z}$$



REGULAR  
 $\rightarrow \hat{z}[t] = \hat{x}[t] \circledast \hat{y}[t] = \sum_{i=0}^{\infty} \hat{x}[i] \hat{y}[t-i]$

$\rightarrow$  NOW DEFINE CIRCULAR

CONVOLUTION TO BE:

CIRCULAR  
 $\rightarrow \hat{z}[t] \triangleq \sum_{i=0}^{N-1} \hat{x}[i] \hat{y}[t-i], \quad \forall t.$

$\rightarrow$  WE'LL WRITE IT AS:

$$\hat{z}[t] = \hat{x}[t] \circledast_c \hat{y}[t]$$

CIRCULAR CONVOLUTION

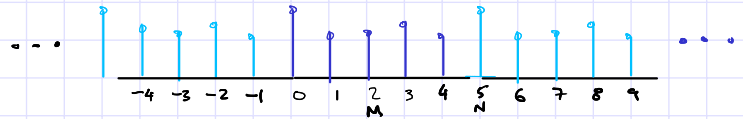
$\rightarrow$  CLAIM:  $\hat{z}[t]$  is periodic/circular

$\rightarrow$  Proof:

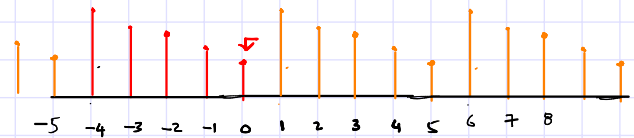
$$\begin{aligned} \hat{z}[t+kN] &= \sum_{i=0}^{N-1} \hat{x}[i] \hat{y}[t-i+kN] \\ &= \sum_{i=0}^{N-1} \hat{x}[i] \hat{y}[t-i] = \hat{z}[t] \end{aligned}$$

$\hat{z}[k] = \hat{z}[0]$

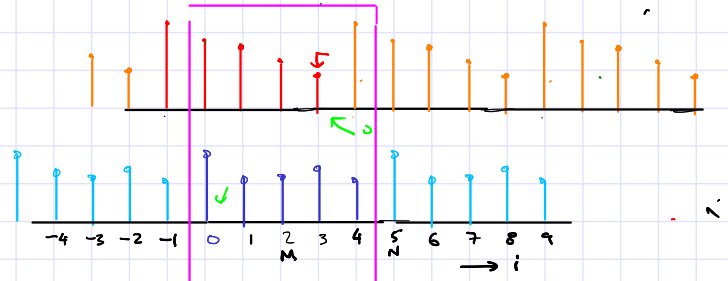
$\hat{x}[i]$



$\hat{y}[t-i]$

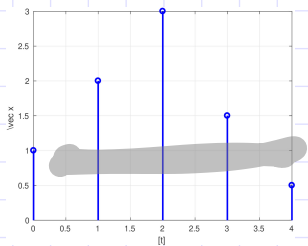


$\hat{x}[i] \hat{y}[t-i]$   
 $\leftarrow t=3 \rightarrow$

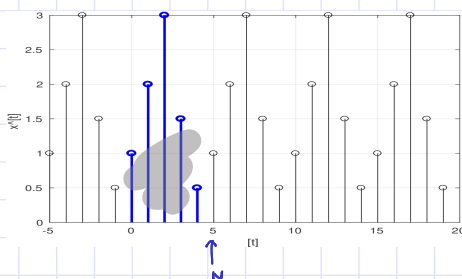


$\rightarrow$  MULTIPLY PAIRWISE AND ADD  $= \hat{z}[t=3]$

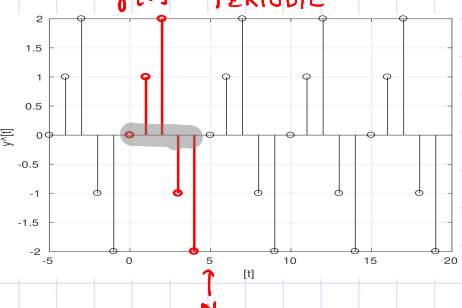
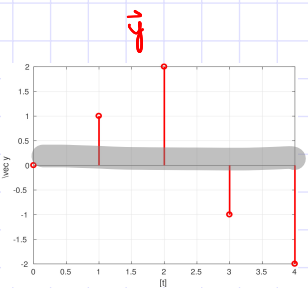
$\hat{x}$



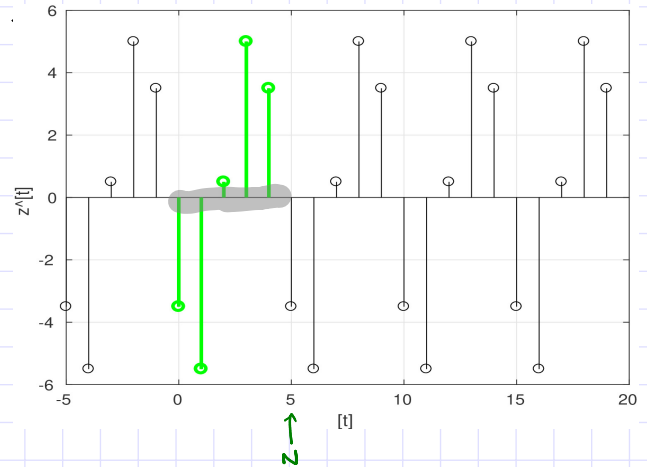
$\hat{x}[t] - \text{PERIODIC}$



$\hat{y}[t] - \text{PERIODIC}$



$$\hat{z}[t] = \hat{x}[t] \circledast_c \hat{y}[t]$$



## 9C: DFT and CIRCULAR CONVOLUTION

→ KEY PROPERTY OF DFTs:

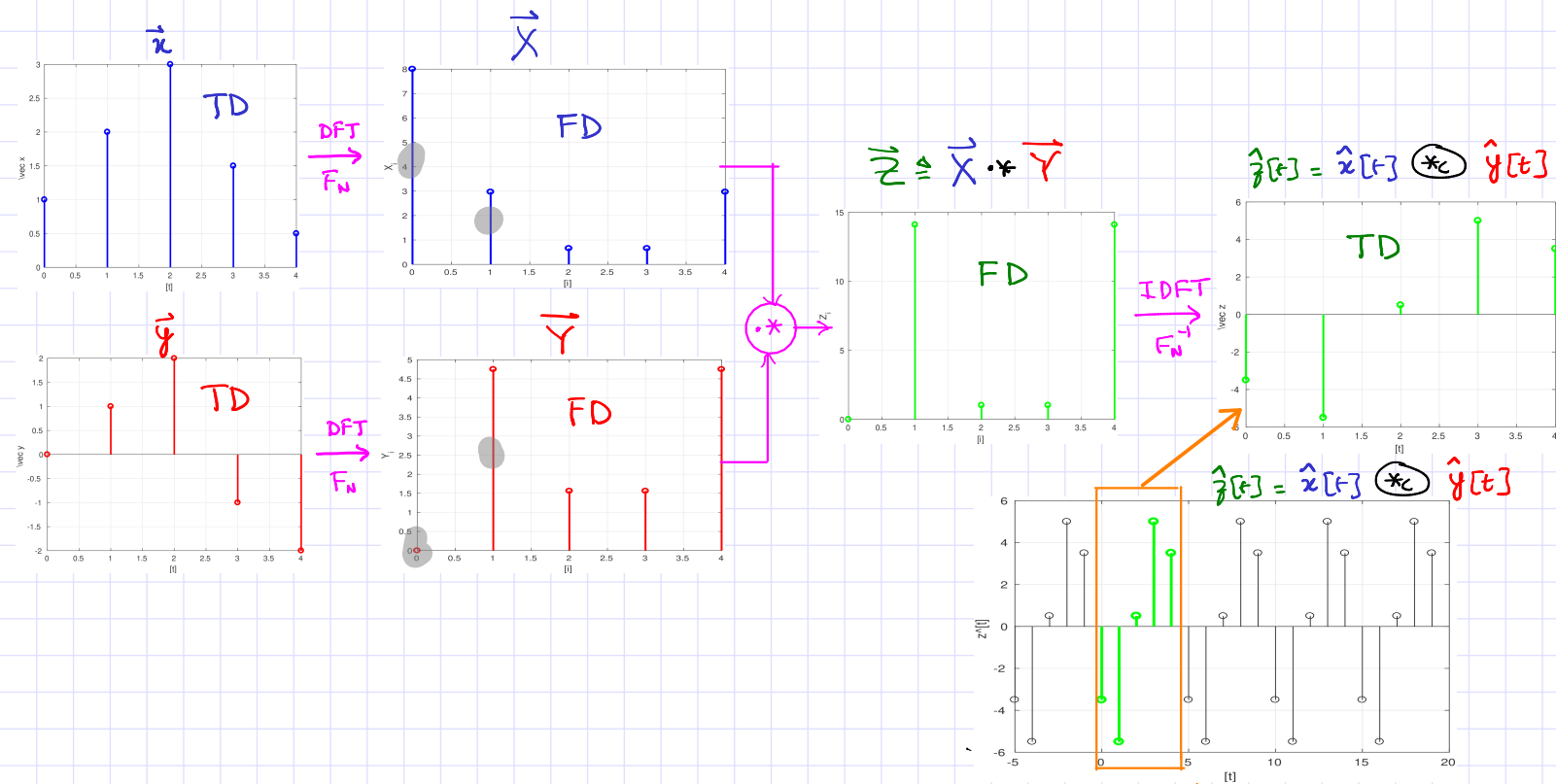
$$\left[ \begin{array}{l} \rightarrow \text{EXTEND } \hat{x} \text{ and } \hat{y} \text{ periodically to } \hat{x}[k] \text{ \& } \hat{y}[k], \text{ resp.} \\ \rightarrow \text{Define } \hat{z}[k] = \hat{x}[k] \circledast \hat{y}[k] \quad \leftarrow \text{CIRCULAR CONVOLUTION} \\ \rightarrow \text{Define } \vec{z} = [\hat{z}[0], \hat{z}[1], \dots, \hat{z}[N-1]]^T \end{array} \right]$$

→ LET  $\vec{X} = F_N \vec{x}$ ,  $\vec{Y} = F_N \vec{y}$ ,  $\vec{Z} = F_N \vec{z}$  (i.e., take the DFTs of  $\vec{x}$ ,  $\vec{y}$ , and  $\vec{z}$ )

$\vec{x} = \begin{bmatrix} x_0 \\ x_1 \\ \vdots \\ x_{N-1} \end{bmatrix}$ 
 $\vec{y} = \begin{bmatrix} y_0 \\ y_1 \\ \vdots \\ y_{N-1} \end{bmatrix}$ 
 $\vec{z} = \begin{bmatrix} z_0 \\ z_1 \\ \vdots \\ z_{N-1} \end{bmatrix}$

→ then:  $\vec{Z} = \vec{X} \cdot \vec{Y} = \begin{bmatrix} x_0 y_0 \\ x_1 y_1 \\ \vdots \\ x_{N-1} y_{N-1} \end{bmatrix}$

ELEMENT BY ELEMENT MULTIPLICATION



→ DFT TURNS CIRCULAR CONVOLUTION IN THE TIME-DOMAIN INTO MULTIPLICATION IN THE FREQUENCY DOMAIN

## THE DFT AND LTI SYSTEMS

→ RECALL: LTI SYSTEMS CHARACTERIZED BY IMPULSE RESPONSE  $h[t]$

$$\rightarrow y[t] = u[t] \circledast h[t] = \sum_{i=0}^t u[i] h[t-i] \quad (\text{assuming } h[t] \text{ \& } u[t] \text{ both causal})$$

↑ NOT CIRCULAR CONV.

→ Q: SUPPOSE I WANT TO CALCULATE  $y[0], \dots, y[n]$

→ HOW MANY MULTIPLICATION OPERATIONS DO I NEED?

→ FOR  $y[0]: 1; y[1]: 2; y[2]: 3, \dots, y[n]: n+1$

$$\Rightarrow 1+2+\dots+(n+1) = \frac{(n+1)(n+2)}{2} \approx \frac{n^2}{2} \text{ if } n \text{ is large.}$$

→ TRY  $n \sim 2 \times 10^6$  (1 minute of CD audio)  $\Rightarrow \frac{n^2}{2} \sim 2 \times 10^{12} = 2 \text{ TRILLION OPS}$

↳ ~10m COMPUTER TIME

→ CAN WE DO IT (MUCH) FASTER?

→ YES: USING THE DFT AND CIRCULAR CONVOLUTION

→ STRATEGY

→ FIRST, FIX THE NUMBER OF  $y[t]$  VALUES YOU WANT:  $y[0], \dots, y[N-1]$  DECIDE N

→ A: EXPRESS  $y[t]$  AS A CIRCULAR CONVOLUTION, NOT JUST A REGULAR CONVOLUTION

$$\rightarrow y[t] = u[t] \circledast h[t]$$

$$\stackrel{?}{=} \hat{u}_p[t] \circledast \hat{h}_p[t], \quad t = 0, \dots, N-1$$

→ CAN WE FIND SUITABLE  $\hat{u}[t]$  AND  $\hat{h}_p[t]$ ?

→ B: IF WE CAN, THEN THE DFT CAN PERFORM CIRCULAR CONVOLUTION:

$$\rightarrow \vec{\hat{y}}_p \triangleq (\text{DFT of } \vec{\hat{u}}_p) * (\text{DFT of } \vec{\hat{h}}_p)$$

$$\rightarrow \vec{\hat{y}}_p = \text{IDFT of } \vec{\hat{y}}_p$$

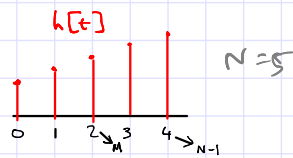
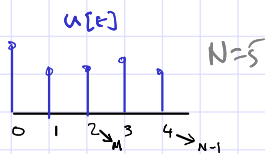
→ THE FIRST  $N$  ENTRIES SHOULD BE  $y[0], \dots, y[N-1]$

# 9D: EXPRESSING CONVOLUTION AS CIRCULAR CONVOLUTION

→ WE HAVE:  $y[t] = \sum_{i=0}^t u[i] h[t-i]$  (u, h, BOTH CAUSAL)

→ WE ARE INTERESTED ONLY IN  $t = 0, \dots, N-1$

→ WILL NEED ONLY  $u[0], \dots, u[N-1]$  AND  $h[0], \dots, h[N-1]$



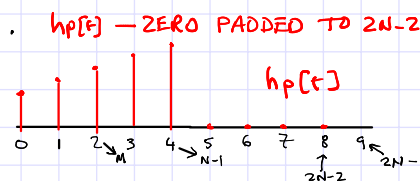
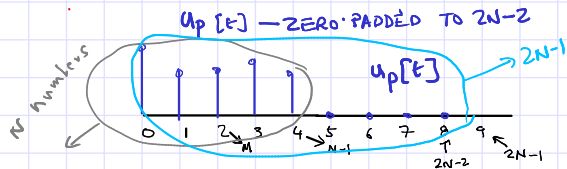
→ DEFINE  $u_p[t]$  and  $h_p[t]$ : EXTENDED AND ZERO PADDED VERSIONS OF  $u[t]$  /  $h[t]$

→  $u_p[t] = u[t]$ ,  $t = 0, \dots, N-1$  (N original TD values)

→  $u_p[N], u_p[N+1], \dots, u_p[2N-2] = 0$  ← (ADDITIONAL N-1 ZEROS)

→  $h_p[t] = h[t]$ ;  $t = 0, \dots, N-1$  (N original TD values)

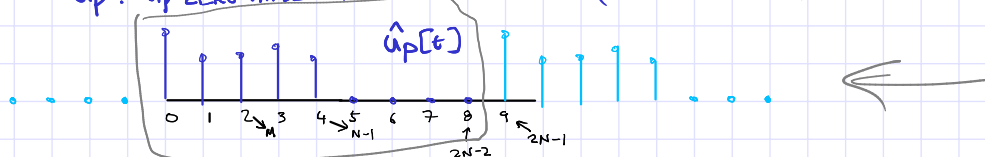
→  $h_p[N], h_p[N+1], \dots, h_p[2N-2] = 0$  ← (ADDITIONAL N-1 ZEROS)



→ NOW EXTEND BOTH PERIODICALLY (MAKE THEM CIRCULAR)

$N \rightarrow 2N-1$   
↓

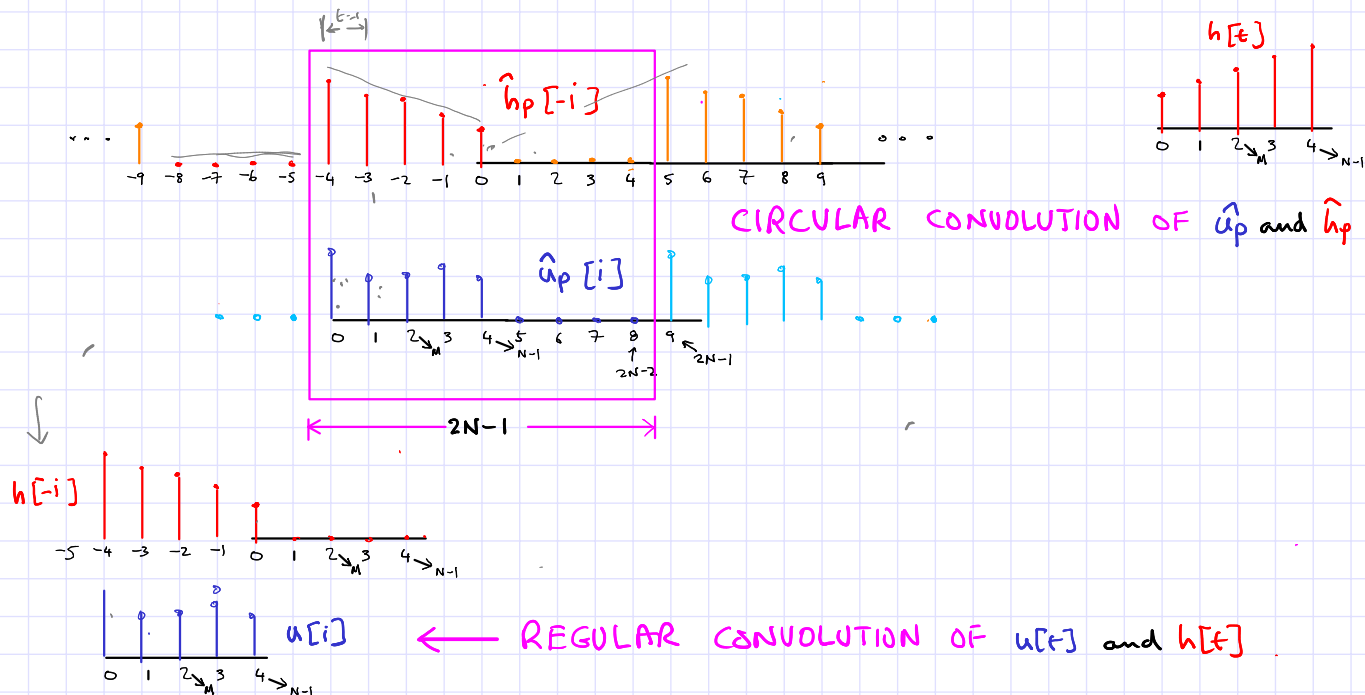
$\hat{u}_p$ :  $u_p$  ZERO PADDED TO  $2N-2$  AND REPEATED (WITH PERIOD  $2N-1$ )



$\hat{h}_p$ :  $h_p$  ZERO PADDED TO  $2N-2$  AND REPEATED (W PERIOD  $2N-1$ )



→ WHAT DOES THE CIRCULAR CONVOLUTION OF  $\hat{u}_p$  and  $\hat{h}_p$  LOOK LIKE?  
 → VS REGULAR CONVOLUTION?



→ IDENTICAL FOR  $t = 0, \dots, N-1$ !  
 → HENCE, CAN USE DFTs FOR REGULAR CONVOLUTION

### 9E. STRATEGY FOR FAST CONVOLUTION USING DFTs

→ CHOOSE  $N = 2M+1$  (odd)

→ set  $\vec{u}_p = \begin{cases} u[t], & t=0, \dots, N-1 \\ 0, & t=N, \dots, 2N-2 \end{cases} \vec{u}_p \in \mathbb{R}^{2N-1}$

→ set  $\vec{h}_p = \begin{cases} h[t], & t=0, \dots, N-1 \\ 0, & t=N, \dots, 2N-2 \end{cases} \vec{h}_p \in \mathbb{R}^{2N-1}$

→ RUN 2 DFTs:  $\vec{U}_p \triangleq F_{2N-1} \vec{x}_p$ ,  $\vec{H}_p \triangleq F_{2N-1} \vec{x}_p$

→ SET  $\vec{Y}_p = \vec{X}_p \circledast \vec{H}_p$   
 ↳ ELEM. BY ELEM. MULT.

→ RUN an INVERSE DFT:  $\vec{y}_p = F_{2N-1}^{-1} \vec{Y}_p$

→  $y[0] = y_p[0] \dots, y[N-1] = y_p[N-1] (= u[t] \circledast h[t])$

→ IGNORE  $y_p[N], \dots, y_p[2N-2]$

$N \log(N)$

#### COMPUTATION

FFT:  $\sim 2N \log(2N) \times 2$

$\circledast$ :  $\sim 2N$  MULTIPLICATIONS

IFFT:  $\sim 2N \log(2N)$

→  $6N \log(2N) + 2N$

→ vs.  $\frac{N^2}{2}$

→ TRY  $N = 2 \times 10^6$

→ FFT BASED:  $2.67 \times 10^8$

→ DIRECT:  $2 \times 10^{12}$

## → APPENDIX:

— DFT TURNS CIRCULAR CONVOLUTION INTO MULTIPLICATION: PROOF

$$\vec{z} = F_N \vec{y}$$

$$z_i = \sum_{j=0}^{N-1} y_j \bar{\omega}_N^{ij}$$

→ GIVEN  $z[t] = x[t] \circledast y[t] = \sum_{i=0}^{N-1} x[i] y[t-i]$

$$\begin{aligned} \rightarrow z_i &= \sum_{j=0}^{N-1} \bar{\omega}_N^{-ij} z[j] = \sum_{j=0}^{N-1} \bar{\omega}_N^{-ij} \left( \sum_{k=0}^{N-1} x[k] y[j-k] \right) = \sum_{k=0}^{N-1} x[k] \left( \sum_{j=0}^{N-1} \bar{\omega}_N^{-ij} y[j-k] \right) \\ &= \sum_{k=0}^{N-1} x[k] \left[ \underbrace{\sum_{j=0}^{k-1} \bar{\omega}_N^{-ij} y[j-k]}_{\text{CALL THIS } A_k} + \underbrace{\sum_{j=k}^{N-1} \bar{\omega}_N^{-ij} y[j-k]}_{\text{CALL THIS } B_k} \right] = \sum_{k=0}^{N-1} x_k (A_k + B_k) \end{aligned}$$

$$\rightarrow A_k = \sum_{j=0}^{k-1} \bar{\omega}_N^{-ij} y[j-k]$$

→ From circularity:  $y[j-k] = y[j-k+N]$

$$\Rightarrow A_k = \sum_{j=0}^{k-1} \bar{\omega}_N^{-ij} y[j-k] = \sum_{j=0}^{k-1} \bar{\omega}_N^{-ij} y[j-k+N]$$

→ Define  $l = j-k+N \Rightarrow j = l+k-N$

$$\Rightarrow A_k = \sum_{j=0}^{k-1} \bar{\omega}_N^{-ij} y[j-k+N] = \sum_{l=N-k}^{N-1} \bar{\omega}_N^{-i(l+k-N)} y[l] = \frac{\bar{\omega}_N^{-ik}}{\omega_N} \left( \sum_{l=N-k}^{N-1} \bar{\omega}_N^{-il} y[l] \right)$$

$$\rightarrow B_k = \sum_{j=k}^{N-1} \bar{\omega}_N^{-ij} y[j-k]$$

→ Let  $l = j-k \Rightarrow j = l+k$

$$\Rightarrow B_k = \sum_{j=k}^{N-1} \bar{\omega}_N^{-ij} y[j-k] = \sum_{l=0}^{N-k-1} \bar{\omega}_N^{-i(l+k)} y[l] = \bar{\omega}_N^{-ik} \sum_{l=0}^{N-k-1} \bar{\omega}_N^{-il} y[l]$$

$$\Rightarrow A_k + B_k = \bar{\omega}_N^{-ik} \sum_{l=N-k}^{N-1} \bar{\omega}_N^{-il} y[l] + \bar{\omega}_N^{-ik} \sum_{l=0}^{N-k-1} \bar{\omega}_N^{-il} y[l] = \bar{\omega}_N^{-ik} \underbrace{\sum_{l=0}^{N-1} \bar{\omega}_N^{-il} y[l]}_{Y_i} = \bar{\omega}_N^{-ik} Y_i$$

$$\Rightarrow z_i = \sum_{k=0}^{N-1} (A_k + B_k) x_k = \sum_{k=0}^{N-1} \bar{\omega}_N^{-ik} x_k Y_i = \underbrace{Y_i}_{X_i} \sum_{k=0}^{N-1} \bar{\omega}_N^{-ik} x_k = X_i Y_i \leftarrow \text{PROVED.}$$

→ ALGEBRAIC PROOF THAT  $\hat{u}[t] \circledast \hat{h}[t] = u[t] \circledast h[t]$  for  $t=0, \dots, N-1$

$$\rightarrow \hat{u}[t] \circledast \hat{h}[t] = \sum_{i=0}^{2N-2} \hat{u}[i] \hat{h}[t-i]$$

$$\Rightarrow \hat{u}[t] \circledast \hat{h}[t] = \sum_{i=0}^{N-1} \hat{u}[i] \hat{h}[t-i] + \sum_{i=N}^{2N-2} \hat{u}[i] \hat{h}[t-i] \quad \begin{matrix} 0 \rightarrow \text{by padding} \\ \uparrow \end{matrix}$$

$$= \sum_{i=0}^{N-1} \hat{u}[i] \hat{h}[t-i]$$

→ WE ARE INTERESTED ONLY IN  $t=0, \dots, N-1$

NOTE:  $t-i$  ranges from  $(-N+1)$  to  $(N-1)$

NOTE2:  $\hat{h}[-N+1], \hat{h}[-N+2], \dots, \hat{h}[-1] = 0$

→ BECAUSE OF CIRCULARITY and ZERO PADDING:

$$\rightarrow \text{eg: } \hat{h}[-N+1] = \hat{h}[-N+1+2N-1] = \hat{h}[N] = 0$$

CIRCULARITY      FIRST ZERO PADDING LOCATION

$$\rightarrow \text{eg: } \hat{h}[-1] = \hat{h}[-1+2N-1] = \hat{h}[2N-2] = 0$$

LAST ZERO PADDING LOCATION

$$\Rightarrow \hat{u}[t] \circledast \hat{h}[t] = \sum_{i=0}^{N-1} \hat{u}[i] \hat{h}[t-i] = \sum_{i=0}^t \hat{u}[i] \hat{h}[t-i], \quad t=0, \dots, N-1$$

↓  
0 if  $i > t$

→  $t$  ranges from 0 to  $N-1$

→  $i$  ranges from 0 to  $t \Rightarrow t-i$  ranges from 0 to  $t$

→  $i$  and  $t-i$  both range from 0 to  $N-1$

$$\Rightarrow \hat{u}[i] = u[i] \text{ and } \hat{h}[t-i] = h[t-i]$$

$$\Rightarrow \hat{u}[t] \circledast \hat{h}[t] = \sum_{i=0}^t u[i] h[t-i] = u[t] \circledast h[t], \quad t=0, \dots, N-1$$

→ SUMMARY:

FROM  $t=N$  to  $t=2N-2$  WITH PERIOD  $2N-1$   
 → ZERO-PAD  $u[t]$  and  $h[t]$ , THEN MAKE CIRCULAR/PERIODIC, CALL THESE  $\hat{u}[t]$  and  $\hat{h}[t]$

$$\Rightarrow \text{then } \hat{u}[t] \circledast \hat{h}[t] = u[t] \circledast h[t] \text{ for } t=0, \dots, N$$